Experiments on 2-DOF Helicopter Using Approximate Dynamic Programming

by

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Abstract

In recent years, the use of unmanned aerial vehicles has seen significant growth in commercial and military sectors. Motion control of such vehicles remains difficult due to precessional torques causing erratic movements. Conventional linear control techniques are widely used. This project aims to experiment with the use of an adaptive motion control strategy. Specifically, a model-based reinforcement learning strategy, approximate dynamic programming, was examined for use on a 2 degree-of-freedom (2-DOF) helicopter, the Quanser AERO.

The adaptive motion control strategy for a 2-DOF helicopter was implemented following electrical engineering methodologies. The proposed strategy was modified for the specific system which required modeling and mathematical analysis. Simulations and experiments were conducted in certain operating conditions. The motion control strategy was implemented to the Quanser AERO using Simulink, a Raspberry Pi, and a Raspberry Pi/smart phone. With successful implementation, a proof of concept can be attained for use in other applications using embedded systems.

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Abbreviations

- **2-DOF** 2 Degrees-Of-Freedom
- \mathbf{ADP} Approximate Dynamic Programming
- \mathbf{LQR} Linear Quadratic Regulator
- \mathbf{PID} Proportional-Integral-Derivative
- ${\bf RMSE}$ Root Mean Squared Error
- ${\bf SPI}$ Serial Peripheral Interface

 $\mathbf{V}\textbf{-}\mathbf{R}\mathbf{E}\mathbf{P}$ - Virtual Robot Experimentation Platform

Mathematical Symbols

- ${\cal J}_p({\cal J}_y)$ Total moment of inertia about pitch (yaw) axis
- $J_{\theta}(J_{\psi})$ Total moment of inertia about pitch (yaw) axis
- ${\cal D}_p({\cal D}_y)$ Damping constant about pitch (yaw) axis
- $K_{\mathbf{sp}}$ Stiffness about pitch axis
- $K_{\mathbf{pp}}(K_{\mathbf{py}})$ Torque thrust gain acting on pitch from pitch (yaw) rotor
- $K_{\mathbf{y}\mathbf{y}}(K_{\mathbf{y}\mathbf{p}})$ Torque thrust gain acting on yaw from yaw (pitch) rotor
- $V_p(V_y)$ Applied voltage to pitch (yaw) motor
- $V_0(V_1)$ Applied voltage to motor #0~(#1)
- $\theta(t)[\psi(t)]$ Pitch [yaw] angle at time $t\geq 0$

Chapter 1

Introduction

The area of unmanned aerial vehicles has seen significant growth over the past few years. Popular configurations of unmanned aerial vehicles include quadcopters and helicopters. The difficulty with such unmanned aerial vehicles is the nonlinear nature of the system due to coupling between the various rotors in addition to the stochastic nature of the environment that these vehicles operate in.

One such platform for testing control techniques of such unmanned aerial vehicles is the Quanser AERO. Essentially, the Quanser AERO is a testbed developed by Quanser Inc. The Quanser AERO offers accurate measurements of the system's pitch angle, yaw angle, as well as a variety of other velocity and acceleration data.

Frequently, the Quanser AERO is used by researchers to implement test algorithms for validation on a physical system that bear resemblance to a quadcopter or helicopter. Those techniques often involve expansive linearization; however, most techniques are not adaptive. The proposed model-based reinforcement learning technique known as approximate dynamic programming, ADP, uses an approximate system model and measured state error data to update the state-feedback gain of the system.

1.1 Literature Review

The Quanser AERO has been studied extensively for a variety of control methods, but Quanser also provides instructional workbooks and lab guides. Quanser documentation uses the linearized system model to calculate LQR state-feedback gain for teaching purposes. Quanser documentation also implements the use of linear-quadratic Gaussian (LQG) control, but more advanced control techniques have been proposed as well.

More extensive control techniques have been studied outside of Quanser Inc. LQR has been used by [17], but an additional adaptive controller was implemented. Sliding mode control has been studied by [1], but sliding mode control can become difficult to implement in comparison to other control techniques.

Fuzzy model-based control was studied by [5, 12]. [4] developed a fuzzy stability controller by using a PD controller to train an adaptive neural fuzzy inference system. Fuzzy control relies on two controllers until the fuzzy control takes complete action making the task of design more intensive considering two controllers.

[6] uses an ADP approach, but the system parameters are relied upon heavily. Neural networks have been proposed in [9], but the use of high-order neural networks are computationally tedious. Reinforcement learning is used in [16], but stabilization is controlled first then parameter estimation of the system.

The problem with the aforementioned techniques is they either involve considerable mathematics not easily implemented in embedded systems or known parameters which may be difficult to calculate. That is not to say that understanding the dynamics of quadcopters and their control in general is easy. Dynamics of quadcopters were discussed in [7, 14, 11]. [14] implemented PD stabilization, and [11] developed position control on a quadcopter. [3] implemented quadcopter control via a PID controller tuned with an LQR loop. [15] developed a real-time, adaptive high-gain EKF, extended Kalman filter, which regulates system angles suitable for inertial navigation.

The Quanser AERO and quadcopters in general have seen significant research over the past few years, but not many approaches prove adaptable to the system and environment.

1.2 Problem Statement

As discussed briefly earlier, the difficulty with controlling a nonlinear system is determining the optimal state-feedback gain. ADP is one such method that hopes to make this difficulty possible. In order to implement ADP on a physical system, standard engineering practices must be followed to have successful implementation. The steps leading to implementation include analyzing the method, modeling the system, simulating the method on the system, and finally implementing the method on the physical system.

In our case, the method, ADP, was analyzed and modified to control our specific system, the Quanser AERO. Because ADP utilizes an approximate system model, the Quanser AERO's system model needed to be derived. In order to see the effectiveness of ADP, we must perform simulations verifying our method and system. These simulations were conducted in both MATLAB and V-REP. With promising simulation results, physical implementation could begin. Implementation included implementing ADP in Simulink directly, on a Raspberry Pi, and using a Raspberry Pi/Android smart phone.

1.3 Report Organization

This report is organized as follows. Chapter 1 gives an introduction to the project and problem at hand. Chapter 2 discusses the model-based reinforcement learning approach

known as approximate dynamic programming, or ADP. Chapter 3 discusses the modeling of the system for ADP to be applied to, the Quanser AERO. Chapter 4 discusses the MATLAB simulations conducted to verify that the use of ADP for the Quanser AERO is possible. Chapter 5 extends the simulations conducted in Chapter 4 to provide a virtual representation of the system. Chapter 6 discusses how ADP can be physically implemented onto hardware. Finally, Chapter 7 discusses the conclusions found throughout this project.

Further resources are attached as appendices. Appendix A provides the MATLAB simulation code. Appendix B provides a tutorial and code pertaining to the real-time simulations. Appendix C provides a tutorial and MATLAB code pertaining to controlling the Quanser AERO via the QFLEX 2 USB panel. Appendix D provides a tutorial and MATLAB code pertaining to controlling the Quanser AERO via the QFLEX 2 Embedded panel and a Raspberry Pi. Appendix E provides a tutorial of implementing an Android smart phone application.

Chapter 2

Model-Based Reinforcement Learning

Reinforcement learning is a type of machine learning where an algorithm tries to determine proper actions in order to minimize or maximize some sort of performance measurement. The question is how can reinforcement learning be applied to the area of control theory? A perfect example is a simple state-feedback diagram shown in Figure 2. For most cases



Figure 2.1: Generic state-feedback block diagram.

studied in an academic environment, the system in Figure 2 would be a linear system. The problem with this assumption is that most physical systems are in fact nonlinear systems. For a linear system, the state-feedback gain, K, is easily computed using conventional control techniques. For the nonlinear systems on the other hand, an accurate calculation for the state-feedback gain can be either difficult or impossible. This is where approximate dynamic programming, ADP, comes into play.

The main objective of ADP is to update the state-feedback gain in Figure 2 by using a model-based reinforcement learning approach. ADP accomplishes this task by utilizing measured state error data and an approximate system model. State error data is collected for T seconds at sample times of τ seconds. Every T seconds the state-feedback gain is updated by using ADP. ADP uses the measured state error data as initial conditions to the approximate system model. ADP predicts future state errors as the initial conditions

would be propagated through system model. By adjusting the value of K, ADP can try to minimize or maximize a performance measurement to determine how the updated K values would affect the system. This seems like a rather simple approach, but ADP can become more complex mathematically.

The rest of this chapter is organize as follows. Section 2.1 goes into more depth concerning the mathematics of ADP. Section 2.2 discusses some situations which may deter ADP and how they can be overcame.

2.1 ADP

Let us denote the state and control variables of the system to be controlled with $\mathbf{x}^{T}(t)$ and $\mathbf{u}^{T}(t)$ at time $t \geq 0$, respectively. In order to utilize ADP, an approximate system model is needed which can be in the form of a typical state-space model seen in Equation 2.1.

$$\dot{\mathbf{x}}(t) = \mathbf{A}\mathbf{x}(t) + \mathbf{B}\mathbf{u}(t) \tag{2.1}$$

We suppose that a more exact nonlinear system model could be used, but in this project, an approximated linear system model was used in ADP. Suppose that the system is supposed to reach a predefined desired (reference) state $\mathbf{x}^{[d]}$ yielding its state error denoted by $\mathbf{e}(t) = \mathbf{x}^{[d]} - \mathbf{x}(t)$. Because all simulations and implementation must be considered in discrete-time, a sampling time must be defined for data collection and input update. Let $t = k\tau$, where $k = 0, 1, 2, \ldots$ is the discrete-time index, and $\tau > 0$ is the sampling time between data collection and model update. The sampling time in which ADP updates the state-feedback gain is T.

Using the first-order Euler integration, the discrete-time error model of the system can then be written as

$$\mathbf{e}[k+1] = \mathbf{f}(\mathbf{e}[k]) + \mathbf{G}\mathbf{u}[k]$$
(2.2)

where $\mathbf{f}(\mathbf{e}[k])$ and matrix \mathbf{G} are given as

$$\mathbf{f}(\mathbf{e}[k]) = e[k] + \tau \mathbf{A}e[k] - \tau \mathbf{A}\mathbf{x}^{[d]}$$
(2.3a)

$$\mathbf{G} = -\tau \mathbf{B}.\tag{2.3b}$$

The problem is to now find the optimal $\mathbf{u}^*[k]$ such that the system's state error $\mathbf{e}[k]$ asymptotically converges to zero, *i.e.*, $\mathbf{e}[k] \to \mathbf{0}$ as $k \to \infty$.

The model-based reinforcement learning approach is based on the conventional dynamic programming technique which has the ability to determine optimal control/action inputs by considering possible future system states without actually experiencing them [8]. In

other words, the task is to find the sequence of actuator inputs $\{\mathbf{u}(k)\}_{k=0}^{\infty}$ of the system that minimizes the cost function defined by

$$J(\mathbf{u}) = \sum_{k=0}^{\infty} \left(\mathbf{e}^{T}(k) \mathbf{Q} \mathbf{e}(k) + \mathbf{u}^{T}(k) \mathbf{R} \mathbf{u}(k) \right)$$
(2.4)

subject to (2.2), where $\mathbb{R}^{4\times4} \ni \mathbf{Q} = \mathbf{Q}^T \ge \mathbf{0}$ and $\mathbb{R}^{2\times2} \ni \mathbf{R} = \mathbf{R}^T > \mathbf{0}$ penalize the system's error deviation from its desired state and its control effort, respectively. It should be noted that these dimensions are with respect to our specific system which will be discussed in the next chapter.

In addition, the cost-to-go function (value function) of the system is defined as

$$V(\mathbf{e}(k)) = \sum_{\kappa=k}^{\infty} \left(\mathbf{e}^{T}(\kappa) \mathbf{Q} \mathbf{e}(\kappa) + \mathbf{u}^{T}(\kappa) \mathbf{R} \mathbf{u}(\kappa) \right).$$
(2.5)

Note that the second term of the value function (2.5) takes into account the asymptotic energy consumption of the system. Following [13, Ch. 11], the value function (2.5) can be rewritten in the form:

$$V(\mathbf{e}(k)) = \mathbf{e}^{T}(k)\mathbf{Q}\mathbf{e}(k) + \mathbf{u}^{T}(k)\mathbf{R}\mathbf{u}(k) + V(\mathbf{e}(k+1)).$$
(2.6)

Therefore, the optimal control input $\mathbf{u}(k)$ that drives the system using (2.1) to its desired configuration can be obtained by solving the following minimization problem:

$$\mathbf{u}^{*}(k) = \operatorname{argmin}_{\mathbf{u}(k)} \Big(\mathbf{e}^{T}(k) \mathbf{Q} \mathbf{e}(k) + \mathbf{u}^{T}(k) \mathbf{R} \mathbf{u}(k) + V^{*}(\mathbf{e}(k+1)) \Big).$$
(2.7)

A standard solution to the minimization problem (2.7) is obtained by solving the discretetime Hamilton–Jacobi–Bellman equation given by

$$V^*(\mathbf{e}(k)) = \min_{\mathbf{u}(k)} \left(\mathbf{e}^T(k) \mathbf{Q} \mathbf{e}(k) + \mathbf{u}^T(k) \mathbf{R} \mathbf{u}(k) + V^*(\mathbf{e}(k+1)) \right).$$
(2.8)

Following [2], calculating the gradient of the right side of (2.8) and setting it equal to zero yields

$$\frac{\partial}{\partial \mathbf{u}(k)} \left(\mathbf{e}(k)^T \mathbf{Q} \mathbf{e}(k) + \mathbf{u}^T(k) \mathbf{R} \mathbf{u}(k) \right) + \left(\frac{\partial \mathbf{e}(k+1)}{\partial \mathbf{u}(k)} \right)^T \nabla V^* (\mathbf{e}(k+1)) = \mathbf{0}$$
(2.9)

where ∇ is the gradient operator which defines $\nabla V^*(\mathbf{e}(k+1)) = \frac{\partial V^*(\mathbf{e}(k+1))}{\partial \mathbf{e}(k+1)}$ as a column vector. Using (2.2), the optimal control action of the system at time instant k is given by

$$\mathbf{u}^*(k) = -\frac{1}{2}\mathbf{R}^{-1}\mathbf{G}^T\nabla V^*(\mathbf{e}(k+1)).$$
(2.10)

Note that the solution $\mathbf{u}^*(k)$ in (2.10) is straight forward given the fact that $\nabla V^*(\mathbf{e}(k+1))$ is defined. Since the value function (2.5) constitutes an infinite summation that takes into account the changes of the system's state-space, this function is to be approximated based on the data collected along the system trajectory (2.2).

Similar to the conventional LQR-based optimal control technique, the cost-to-go function (2.5) can be expressed as a quadratic function, *i.e.*,

$$V(\mathbf{e}(k)) = \mathbf{e}^{T}(k)\mathbf{P}\mathbf{e}(k) \tag{2.11}$$

for some $\mathbf{P} \in \mathbb{R}^{4 \times 4}$. It should be noted that these dimensions are for our specific system which will be discussed in the next chapter. The cost-to-go function (2.11) can be approximated by

$$V(\mathbf{e}(k)) = (\operatorname{vec}(\mathbf{P}))^T(\mathbf{e}(k) \otimes \mathbf{e}(k)) \equiv \mathbf{w}_c^T \phi(\mathbf{e}(k))$$
(2.12)

where \otimes is the Kronecker product operator, and the weight $\mathbf{w}_c = \operatorname{vec}(\mathbf{P})$, where the operator $\operatorname{vec}(\mathbf{P})$ forms the vector by stacking columns of the matrix \mathbf{P} . The weight vector \mathbf{w}_c can be approximated using the critic neural network which will be discussed shortly. The vector-valued function $\phi(\mathbf{e}(k)) = \mathbf{e}(k) \otimes \mathbf{e}(k)$ is the quadratic polynomial vector containing all possible pairwise products of the two components of $\mathbf{e}(k)$. According to the Weierstrass approximation theorem, the function $V(\mathbf{e}(k))$ can be approximated using basis functions which are polynomials of the combination of state elements $(e^{(1)}(k), e^{(2)}(k), e^{(3)}(k), e^{(4)}(k))$. It should be noted again that these dimensions are for our specific system which will be discussed in the chapter. Here we choose the basis functions up to the second order. Therefore, $\phi(\mathbf{e}(k)) = [\phi_1, \phi_2, \phi_3, \phi_4, \phi_5, \phi_6, \phi_7, \phi_8, \phi_9, \phi_{10}, \phi_{11}, \phi_{12}, \phi_{13}, \phi_{14}]^T$ with

$$\phi_1(\mathbf{e}(k)) = e^{(1)}(k) \tag{2.13a}$$

$$\phi_2(\mathbf{e}(k)) = e^{(2)}(k)$$
 (2.13b)

$$\phi_3(\mathbf{e}(k)) = e^{(3)}(k) \tag{2.13c}$$

$$\phi_4(\mathbf{e}(k)) = e^{(4)}(k) \tag{2.13d}$$

$$\phi_5(\mathbf{e}(k)) = (e^{(1)}(k))^2 \tag{2.13e}$$

$$\phi_6(\mathbf{e}(k)) = e^{(1)}(k)e^{(2)}(k) \tag{2.13f}$$

$$\phi_7(\mathbf{e}(k)) = e^{(1)}(k)e^{(3)}(k)$$
 (2.13g)

$$\phi_8(\mathbf{e}(k)) = e^{(1)}(k)e^{(4)}(k)$$
(2.13h)
$$\phi_8(\mathbf{e}(k)) = (e^{(2)}(k))^2$$
(2.13i)

$$\varphi_{9}(\mathbf{e}(k)) = (e^{\chi}(k))$$
(2.131)
$$\varphi_{0}(\mathbf{e}(k)) = e^{(2)}(k)e^{(3)}(k)$$
(2.13i)

$$\psi_{10}(\mathbf{e}(k)) = e^{(k)}(k)e^{(k)}(k)$$
(2.13)

$$\phi_{11}(\mathbf{e}(k)) = e^{(2)}(k)e^{(2)}(k)$$
(2.13k)

$$\phi_{12}(\mathbf{e}(k)) = (e^{(3)}(k))^2 \tag{2.131}$$

$$\phi_{13}(\mathbf{e}(k)) = e^{(3)}(k)e^{(4)}(k) \tag{2.13m}$$

$$\phi_{14}(\mathbf{e}(k)) = (e^{(4)}(k))^2 \tag{2.13n}$$

being the basis functions of the critic neural network, where the value function (2.12) can be approximated using a linear-in-weight critic neural network structure modeled by

$$\hat{V}(\mathbf{e}(k)) \cong \mathbf{w}_c^T \phi(\mathbf{e}(k)) \tag{2.14}$$

where $\mathbf{w}_c \in \mathbb{R}^{10 \times 1}$ is the weight vector for the critic neural network. The output node of the critic neural network produces the approximate output of the cost-to-go function using (2.14). The critic neural network can be seen in Figure 2.2.



Figure 2.2: Critic neural network.

Similar to [10], the critic weight vector, \mathbf{w}_c , is solved using the least square technique, except that the system's data (state error) is collected online along its trajectory. For that, we introduce another sampling time τ such that the sampling time between statefeedback gain updates is $T = \bar{n}\tau$, with $\bar{n} > 14$ being a scalar parameter that quantifies the number of training samples. Note that training samples are simply a sequence of data points $\{\mathbf{e}(k + \kappa), \mathbf{e}(k + \kappa + 1), \ldots\}_{\kappa=0}^{\bar{n}-1}$ collected by the system along its trajectory, where $V(\mathbf{e}(k + \kappa))$ is the desired (reference) value function at time instant $k + \kappa$ with $\kappa = 0, 1, \ldots, \bar{n} - 1$. To solve for the weight vector $\mathbf{w}_c = [w_1, w_2, \ldots, w_{14}]^T$ using the batch least square technique, the number of training samples of the critic neural network is chosen such that $\bar{n} > 14$. The target value function is determined using

$$V(\mathbf{e}(k)) = \mathbf{e}^{T}(k)\mathbf{Q}\mathbf{e}(k) + \mathbf{u}^{T}(k)\mathbf{R}\mathbf{u}(k) + \mathbf{w}_{c}^{T}\phi(\mathbf{e}(k+1)), \qquad (2.15)$$

and the estimated value function is the output of the critic neural network given by (2.14). The least square error is then defined as

$$\delta_c = \frac{1}{2} \sum_{\kappa=0}^{\bar{n}-1} [V(\mathbf{e}(k)) - \hat{V}(\mathbf{e}(k))]^2.$$
(2.16)

The weight vector, \mathbf{w}_c , that minimizes the sum-of-square error δ_c is given by

$$\mathbf{w}_c = \left(\mathbf{\Lambda}^T \mathbf{\Lambda}\right)^{-1} \mathbf{\Lambda}^T \mathbf{V}$$
(2.17)

where the matrix $\mathbb{R}^{14 \times \bar{n}} \ni \mathbf{\Lambda} = [\mathbf{a}_0, \mathbf{a}_1, \dots, \mathbf{a}_{\bar{n}-1}]^T$ with $\mathbf{a}_{\kappa} = \phi^T(\mathbf{e}(k+\kappa))$ and $\mathbf{V} = [v_0, v_1, \dots, v_{\bar{n}-1}]^T$ with $v_{\kappa} = V(\mathbf{e}(k+\kappa))$ for $\kappa = 0, 1, \dots, \bar{n}-1$.

The key steps of the ADP technique implemented in this work are given in Algorithm 1. A timing flowchart of ADP can be seen in Figure 2.3.



Figure 2.3: ADP timing flowchart.

Algorithm 1: ADP Outline			
Input: $\mathbf{x}^{[d]}$			
Output: x			
1 begin			
2 repeat			
$\mathbf{s} \qquad \bullet \ t = k\tau$			
4 • Apply $\mathbf{u}[k]$ to system model $\rightarrow \mathbf{x}[k]$			
• Apply Equation 2.1 to $\mathbf{x}[k] \to \mathbf{x}[k+1]$			
6 • Constrain ψ on $[-180^\circ, 180^\circ]$ and θ on $[-90^\circ, 90^\circ]$			
• Calculate error $\rightarrow \mathbf{e}[k] = \mathbf{x}^{\text{rer}}[k] - \mathbf{x}[k]$			
s • Calculate updated \mathbf{w}_c if $t = T$			
9 repeat			
10 • $\mathbf{w}_{last} = \mathbf{w}_c, i = 0$			
11 repeat $ $			
12 • $\mathbf{u}[i] = [0, 0]^{T}, j = 0$			
13 repeat			
14 \bullet $\mathbf{u}_{last}[i] = \mathbf{u}[i]$ \bullet Find $c[i + 1]$ using collected space data and Equation 2.2			
• Find $\mathbf{e}[i+1]$ using collected error data and Equation 2.2			
• Compute new $\mathbf{u}[i]$ using Equation 2.10			
17 $ \begin{bmatrix} 0 \\ j = j + 1 \end{bmatrix} $			
18 $ \mathbf{u}[i] - \mathbf{u}_{last}[i] < \epsilon \text{ or } j = j_{max}$			
$\begin{array}{c} \textbf{Output: } \mathbf{u}[i] \\ \textbf{Distribut: } $			
• Find $V(i)$ using Equation 2.15			
• $\Lambda(i) = \phi(\mathbf{e}(i))$			
$\begin{bmatrix} 21 \\ 0 \end{bmatrix} = \begin{bmatrix} 0 \\ i \end{bmatrix} \begin{bmatrix} 0 \\ i \end{bmatrix} \begin{bmatrix} 0 \\ i \end{bmatrix} \begin{bmatrix} 1 \\ i \end{bmatrix}$			
22 Until $i = n - 1$ Output: A V			
Output: A, V			
• Update \mathbf{w}_c using Equation 2.17			
$\ \mathbf{until} \ \mathbf{w}_c - \mathbf{w}_{last} \ < \epsilon_c$			
Output: w_c			
• Calculate state-reedback gain using Equation 2.10 and Equation 2.11			
• Constrain optimal inputs on $[-10, 10]$			
$ -\kappa - \kappa + 1 $			
28			

2.2 ADP Conditions

ADP utilizes a critic neural network to approximate the value function. The weights of this critic neural network are determined recursively. The problem with this approach, and ADP in general, is that these weights can diverge. It was initially suspected that the weights would converge on almost all instances, but it turns out this is not the case. Initial MATLAB simulations, that will be discussed Chapter 4, showed that the weights always seemed to diverge for a linear model. It appears that the neural network cannot handle the error data of a linear system. The cause may be due to the error data not being random enough because of a truly linear system. This divergence of the weights make ADP useless if it cannot support simple linear systems.

After much research and trial-and-error with the MATLAB simulations, it was determined that several conditional cases should be considered when performing the algorithm. Research showed that if random error data was inputted to ADP, the weights would converge on almost all instances. This findings still baffle us as to why the random error data causes convergent weights while the linear system error data causes divergent weights. Further research will be needed to determine the cause of this phenomenon.

The conditional cases implemented in ADP utilize the fact that the weights converge for random error data. Before ADP is used, it is ran with random error data to obtain convergent weights; these weights are then saved. When ADP is called again, several types of errors can cause divergent weights. Because the algorithm is recursive, we can reach the end of one of the recursive loops. If this is the case, the initial convergent weights should be used to save time instead of continuing for further loop iterations. The weights may also not be found because the least-squares regression calculation cannot be performed. If this is the case, the initial convergent weights should be used again as well. The last possible failure would be if the weights actually begin to diverge to infinity. If this is the case, the weights will be assigned a very large number which is user-defined in the algorithm.

These conditional cases in ADP algorithm have been shown effective, and they guarantee that ADP will output some type of weights instead of failing.

Chapter 3 Quanser AERO System Modeling

In order to test the effectiveness of the ADP algorithm, the ideal testing platform would be a nonlinear system that is very difficult to control. This almost chaotic system would then highlight the effectiveness of ADP in its adaptive learning capabilities. Our problem is we do not have such a system readily available for testing, and it is also difficult to simulate such a system. The closest physical system we have is the Quanser AERO (https: //www.quanser.com/products/quanser-aero/) seen in Figure 3.1(a).



Figure 3.1: Quanser AERO. (a) Quanser AERO configured as a 2-DOF helicopter. (b) A simplistic view of the Quanser AERO's free body diagram and its orientation.

The Quanser AERO is a configurable testing platform that utilizes two fans to mimic a 2-DOF helicopter. It should be noted that this system does not actually model a physical helicopter, but it does model something similar. The Quanser AERO can either be configured as a 2-DOF helicopter or a half-quadcopter both of which will be discussed. This reconfiguration can be done by unscrewing various screws and rotating the fan assemblies. Control of the Quanser AERO can be achieved by applying the proper input voltages V_p and V_y to the main and tail DC motors, respectively. Determining the values of these motor input voltages is the task of ADP.

In order to be able use ADP, we need to know the system model of our particular Quanser AERO configuration. This system model can be approximated, but a relatively accurate model compared to that of the actual system model is needed for ADP. In order to utilize state-space modeling, we need to determine what states we will be considering. The Quanser AERO utilizes many sensors for measurement, but we will only focus on the sensors measuring pitch (θ) and yaw (ψ). Let us denote the state and control (DC motor input voltages) variables of the state-space models as $\mathbf{x}^{T}(t) \equiv [\theta(t), \psi(t), \dot{\theta}(t), \dot{\psi}(t)]$ and $\mathbf{u}^{T}(t) \equiv [V_{p}(t), V_{y}(t)]$ at time $t \geq 0$, respectively. It should be noted that $\psi > 0$ when the Quanser AERO moves in a counter-clockwise direction, and *theta* > 0 when the Quanser AERO raises above the horizontal plane.

This chapter examines the state-space models of the Quanser AERO configured as a halfquadcopter and 2-DOF helicopter. This chapter is organized as follows. Section 3.1 shows the derivation of the Quanser AERO configured as a half-quadcopter. Section 3.2 shows the derivation of the Quanser AERO configured as a 2-DOF helicopter.

3.1 Half-Quadcopter

The propeller-arm system comprises half of the Quanser AERO, shown in Figure 3.2. The movement of the arm is characterized by a thrust generated by the propeller and a resultant force, perpendicular to the thrust, generated as a result of Newton's Third Law by virtue of the motor accelerating the propeller and the propeller driveshaft.



Figure 3.2: (a) Rotor arm dynamics (top view) and (b) blade rotations $(V_0 = V_1)$.

We'll assume that each rotor produces an upward force, or *thrust*. T_i , that is proportional to the square of the angular velocity of the propeller times an empirically derived constant, c_t . c_t is a constant empirically derived from various characteristics of the surrounding air

[14].

$$F_{prop_i} = c_t \omega_i^2, \qquad \text{for } i = 0, 1 \tag{3.1}$$

The thrust force will be applied at radius, R, creating a torque on the body in the θ plane.

$$\tau_{\theta_i} = RF_{prop_i}, \qquad \text{for } i = 0, 1 \tag{3.2}$$



Figure 3.3: (a) Force diagram (top view) and (b) force diagram (side view).

Gyroscopic effects from the propeller on the body and torque on body from the drag of the propeller generate torque opposite of the direction of ω_{prop} . The torque will be a function of the squared angular speed of the propeller multiplied by a constant, b. b is empirically determined and understood to be a function of the propeller cross sectional area and surrounding air density.

The gyroscopic effects on the body are defined as torques τ_i and will articulate the Quanser AERO perpendicular to the thrust causing movement in the ψ plane:

$$\tau_{\psi_i} = b\omega_i^2 + J_{prop}\dot{\omega}_i, \qquad \text{for } i = 0, 1 \tag{3.3}$$

Damping torques exist in the Quanser AERO in the form of bearing stiffness. D_{θ} and D_{ψ} are constants empirically determined for the system. These torques, τ_{θ_d} and τ_{ψ_d} , are expressed as follows

$$\tau_{\theta_d} = -D_\theta \dot{\theta} \tag{3.4a}$$

$$\tau_{\psi_d} = -D_\psi \dot{\psi} \tag{3.4b}$$

Finally, in the pitch plane, a gravitational torque exists on the Quanser AERO. The gravitational torque is found at the center of mass which is R_c from the pivot. The gravitational

force is then the mass of the body and the two propellers. g is the acceleration due to gravity.



Figure 3.4: Gravitational torque exerted on Quanser AERO in θ plane.

$$\tau_{\theta_a} = R_c M_b g \sin(\theta) \tag{3.5}$$

Next, we use a definition of torque, τ :

$$\sum_{n} \tau = J_{\text{total}} \alpha_{\text{total}} = J_{\theta} \ddot{\theta} + J_{\psi} \ddot{\psi}$$
(3.6)

We use components of torque to generalize:

$$\tau_{\theta} = \tau_{\theta_0} + \tau_{\theta_1} + \tau_{\theta_d} + \tau_{\theta_q} \tag{3.7a}$$

$$\tau_{\psi} = (\tau_{\psi_0} + \tau_{\psi_1} + \tau_{\psi_d}) cos(\theta) \tag{3.7b}$$

Next, we consider the coordinates of our system to finalize the equations of motion:

$$\tau_{\theta} = J_{\theta}\ddot{\theta} = Rc_t(\omega_0^2 - \omega_1^2) - R_c M_b gsin(\theta) - D_{\theta}\dot{\theta}$$
(3.8a)

$$\tau_{\psi} = J_{\psi}\ddot{\psi} = b\omega_0^2 - b\omega_1^2 + J_{prop}\dot{\omega}_0 - J_{prop}\dot{\omega}_1 - D_{\psi}\dot{\psi}$$
(3.8b)

The linearized form of the controls are made with the following assumptions:

$$R_c M_b gsin(\theta) \approx K_{sp} \theta$$
 (3.9a)

$$\cos(\theta) \approx 1$$
 (3.9b)

$$J_{prop}\dot{\omega}_{0,1} \approx 0 \tag{3.9c}$$

$$Rc_t \omega_{0,1}^2 \propto K_{pp} V_{0,1} \tag{3.9d}$$

$$b\omega_1^2 \propto K_{yy} V_{0,1} \tag{3.9e}$$

where K_{yy} and K_{pp} are thrust gains, and K_{sp} is the stiffness constant.

$$\tau_{\theta} = J_{\theta} \ddot{\theta} = K_{pp} V_0 - K_{pp} V_1 - K_{sp} \theta - D_{\theta} \dot{\theta}$$
(3.10a)

$$\ddot{\theta} = \frac{K_{pp}V_0 - K_{pp}V_1 - R_c M_b g\theta - D_\theta \theta}{J_\theta}$$
(3.10b)

$$\tau_{\psi} = J_{\psi} \ddot{\psi} = K_{yy} V_0 - K_{yy} V_1 + -D_{\psi} \dot{\psi}$$
(3.11a)

$$\ddot{\psi} = \frac{K_{yy}V_0 - K_{yy}V_1 + -D_{\psi}\psi}{J_{\psi}}$$
(3.11b)

Next, the linearized equations of motion are realized in a state-space representation with Table 3.1 representing the constant values:

$$\begin{bmatrix} \dot{\theta} \\ \dot{\psi} \\ \ddot{\theta} \\ \ddot{\psi} \end{bmatrix} = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ -K_{\rm sp}/J_{\theta} & 0 & -D_{\theta}/J_{\theta} & 0 \\ 0 & 0 & 0 & -D_{\psi}/J_{\psi} \end{bmatrix} \begin{bmatrix} \theta \\ \dot{\psi} \\ \dot{\theta} \\ \dot{\psi} \end{bmatrix} + \begin{bmatrix} 0 & 0 \\ 0 & 0 \\ K_{\rm pp}/J_{\theta} & -K_{\rm pp}/J_{\theta} \\ K_{\rm yy}/J_{\psi} & -K_{\rm yy}/J_{\psi} \end{bmatrix} \begin{bmatrix} V_0 \\ V_1 \end{bmatrix}$$
(3.12)

Table 3.1: Parameters calculated and measured by Quanser.

$\begin{array}{cccc} J_{\theta} & 0.0215 & [\mathrm{kg}\mathrm{m}^2] \\ J_{\psi} & 0.0237 & [\mathrm{kg}\mathrm{m}^2] \\ D_{\theta} & 0.00711 & [\mathrm{N}\mathrm{m}\mathrm{s}\mathrm{rad}^{-1}] \\ D_{\psi} & 0.0220 & [\mathrm{N}\mathrm{m}\mathrm{s}\mathrm{rad}^{-1}] \\ K_{\mathrm{sp}} & 0.0375 & [\mathrm{N}\mathrm{m}\mathrm{rad}^{-1}] \\ K_{\mathrm{pp}} & 0.0011 & [\mathrm{N}\mathrm{m}\mathrm{V}^{-1}] \\ K_{\mathrm{yy}} & 0.0022 & [\mathrm{N}\mathrm{m}\mathrm{V}^{-1}] \end{array}$	Parameter	Value	Unit
$\begin{array}{lll} J_{\psi} & 0.0237 & [\mathrm{kg}\mathrm{m}^2] \\ D_{\theta} & 0.00711 & [\mathrm{N}\mathrm{m}\mathrm{s}\mathrm{rad}^{-1}] \\ D_{\psi} & 0.0220 & [\mathrm{N}\mathrm{m}\mathrm{s}\mathrm{rad}^{-1}] \\ K_{\mathrm{sp}} & 0.0375 & [\mathrm{N}\mathrm{m}\mathrm{rad}^{-1}] \\ K_{\mathrm{pp}} & 0.0011 & [\mathrm{N}\mathrm{m}\mathrm{V}^{-1}] \\ K_{\mathrm{yy}} & 0.0022 & [\mathrm{N}\mathrm{m}\mathrm{V}^{-1}] \end{array}$	$J_{ heta}$	0.0215	$[\mathrm{kg}\mathrm{m}^2]$
$ \begin{array}{ll} D_{\theta} & 0.00711 & [{\rm Nmsrad^{-1}}] \\ D_{\psi} & 0.0220 & [{\rm Nmsrad^{-1}}] \\ K_{\rm sp} & 0.0375 & [{\rm Nmrad^{-1}}] \\ K_{\rm pp} & 0.0011 & [{\rm NmV^{-1}}] \\ K_{\rm yy} & 0.0022 & [{\rm NmV^{-1}}] \end{array} $	J_ψ	0.0237	$[{ m kgm^2}]$
$\begin{array}{lll} D_{\psi} & 0.0220 & [\mathrm{Nmsrad^{-1}}] \\ K_{\mathrm{sp}} & 0.0375 & [\mathrm{Nmrad^{-1}}] \\ K_{\mathrm{pp}} & 0.0011 & [\mathrm{NmV^{-1}}] \\ K_{\mathrm{yy}} & 0.0022 & [\mathrm{NmV^{-1}}] \end{array}$	$D_{ heta}$	0.00711	$[N \mathrm{m}\mathrm{s}\mathrm{rad}^{-1}]$
$K_{\rm sp}$ 0.0375 $[{\rm N}{\rm m}{\rm rad}^{-1}]$ $K_{\rm pp}$ 0.0011 $[{\rm N}{\rm m}{\rm V}^{-1}]$ $K_{\rm yy}$ 0.0022 $[{\rm N}{\rm m}{\rm V}^{-1}]$	D_ψ	0.0220	$[N \mathrm{m}\mathrm{s}\mathrm{rad}^{-1}]$
$\begin{array}{ll} K_{\rm pp} & 0.0011 & [{\rm NmV^{-1}}] \\ K_{\rm yy} & 0.0022 & [{\rm NmV^{-1}}] \end{array}$	$K_{\rm sp}$	0.0375	$[N \operatorname{m} \operatorname{rad}^{-1}]$
K_{yy} 0.0022 [N m V ⁻¹]	$K_{\rm pp}$	0.0011	$[\mathrm{NmV^{-1}}]$
	$K_{\rm yy}$	0.0022	$[\rm NmV^{-1}]$

Using Equations 3.12 and Table 3.1, the model can explicitly be stated as:

$$\begin{bmatrix} \dot{\theta} \\ \dot{\psi} \\ \ddot{\theta} \\ \ddot{\psi} \end{bmatrix} = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ -1.7442 & 0 & -0.3307 & 0 \\ 0 & 0 & 0 & -0.9283 \end{bmatrix} \begin{bmatrix} \theta \\ \psi \\ \dot{\theta} \\ \dot{\psi} \end{bmatrix} + \begin{bmatrix} 0 & 0 \\ 0 & 0 \\ 0.05116 & -0.05116 \\ 0.09282 & -0.09282 \end{bmatrix} \begin{bmatrix} V_0 \\ V_1 \end{bmatrix}$$
(3.13)

3.2 2-DOF Helicopter

In order to utilize ADP for the 2-DOF helicopter, the state-space model of the system must be derived. Quanser documentation has provided this derivation, but we started from the beginning in order to verify their results and better understand the system. It should be noted that the system model derived here contains many assumptions and linearizations. These same assumptions and linearizations were assumed by Quanser in their documentation. It should also be noted that this derivation is very different from the one derived in Section 3.1 where Quanser documentation was not provided.

To begin the state-space model derivation, let us draw a free body diagram of the Quanser AERO which can be seen in Figure 3.2. We can then look at the free body diagram in terms of the two propellers seen in Figures 3.6(a) and 3.6(b). This also divides the forces in terms of horizontal and vertical planes.



Figure 3.5: General free body diagram of the Quanser AERO.

The symbols represented in Figure 3.2 can be described as follows.

- R_p is the distance from the fork of the Quanser AERO to the center of the main propeller.
- R_y is the distance from the fork of the Quanser AERO to the center of the tail propeller.
- R_c is the distance from the fork of the Quanser AERO to the center of mass of the body. The body is the structure connecting and including the main and tail propellers.
- F_p is the force produced by the main rotor causing lift. We will assume that when the helicopter is rising, F_p is positive.
- F_y is the force produced by the tail rotor causing thrust. We will assume that when the helicopter is rotating counter-clockwise, F_y is positive.



Figure 3.6: (a) Forces acting on the propeller controlling the pitch of the Quanser AERO. (b) Forces acting on the propeller controlling the yaw of the Quanser AERO.

We can look at the forces specifically associated with the main rotor as in Figure 3.6(a). The symbols represented in Figure 3.6(a) can be described as follows.

- F_p is the force produced by the main rotor causing lift. We will assume that when the helicopter is rising, F_p is positive.
- $F_{p,tail}$ is the coupled force generated by the tail rotor. To visualize this force, think of the torques on the tail rotor. As the tail propeller spins, the propeller is causing torque on the actual motor shaft. This torque is translated to the pitch axis. We assume this coupling force is aiding the main rotor force as in [1].
- $F_{friction}$ is the frictional force opposing pitch.
- $F_{gravity}$ is the gravitational force exerted on the main rotor. This force only exists in the vertical plane.

We can look at the forces specifically associated with the main rotor as in Figure 3.6(b). The symbols represented in Figure 3.6(b) can be described as follows.

- F_y is the force produced by the tail rotor causing thrust. We will assume that when the helicopter is rotating counter-clockwise, F_y is positive.
- $F_{y,main}$ is the coupled force generated by the main rotor. Again, we used the directions used in [1].
- $F_{friction}$ is the frictional force opposing yaw.

We know that the forces in each of the planes are tangential to the path of motion except the force of gravity. This means we can find the torques associated with each of the forces. Let the torques be defined as follows.

• $\tau_{p,main}$ is the torque associated with the force generated by the main rotor on the main rotor.

- $\tau_{p,tail}$ is the torque associated with the force generated by the tail rotor on the main rotor.
- $\tau_{p,friction}$ is the frictional torque opposing a change in pitch.
- $\tau_{p,gravity}$ is the torque associated with the force of gravity on the main rotor.
- $\tau_{y,tail}$ is the torque associated with the force generated by the tail rotor on the tail rotor.
- $\tau_{y,main}$ is the torque associated with the force generated by the main rotor on the tail rotor.
- $\tau_{y,friction}$ is the frictional torque opposing a change in yaw.
- $\tau_{p,net}$ is the net torque on the main rotor.
- $\tau_{y,net}$ is the net torque on the tail rotor.

We also know that the net torque is the rotational inertia times the tangential angular acceleration. Let the rotational inertia for the the main rotor be J_p , and let the rotational inertia for the tail rotor be J_y . The angular acceleration for the pitch and yaw are $\ddot{\theta}$ and $\ddot{\psi}$, respectively.

$$\tau_{p,net} = J_p \ddot{\theta} \tag{3.14a}$$

$$\tau_{y,net} = J_y \ddot{\psi} \tag{3.14b}$$

Now we can calculate the individual torques because we know that torque is the tangential force times the distance from the pivot. Let us also make one assumption here. Let us assume that the torque generated on the main (tail) rotor from the main (tail) rotor is directly proportional to V_p and V_y , respectively. The proportional gains here will be classified as thrust gains.

$$\tau_{p,main} = R_p F_p = K_{p,main} V_p \tag{3.15a}$$

$$\tau_{p,main} = G_p \tag{3.15b}$$

$$\tau_{p,main} = R_p \beta_p \dot{\theta} \tag{3.15c}$$

 G_p is the torque coupling which we have yet determined. The tangential frictional force is the equal to a damping coefficient, β_p , times the velocity, $\dot{\theta}$.

$$\tau_{p,main} = R_y F_y = K_{y,tail} V_y \tag{3.16a}$$

$$\tau_{p,main} = G_y \tag{3.16b}$$

$$\tau_{p,main} = R_y \beta_y \dot{\psi} \tag{3.16c}$$

$$\tau_{p,qravity} = R_c m_{body} gsin(\theta) \tag{3.16d}$$

 G_y is the torque coupling which we have yet determined. The tangential frictional force is the equal to a damping coefficient, β_y , times the velocity, $\dot{\psi}$. The gravitational torque is found at the center of mass which is R_c from the pivot. The gravitational force is then the mass of the body and the two propellers. g is the acceleration due to gravity.

We can now combine Equations 3.14, 3.15, and 3.16 using the same sign conventions as used in Figures 3.6(a) and 3.6(b).

$$J_p \ddot{\theta} = K_{p,main} V_p + G_p - R_p \beta_p \dot{\theta} - R_c m_{body} gsin(\theta)$$
(3.17a)

$$J_y \ddot{\psi} = K_{y,tail} V_y - G_y - R_y \beta_y \dot{\psi}$$
(3.17b)

We can rewrite Equations 3.17a and 3.17b with the second derivative state variables set alone.

$$\ddot{\theta} = \frac{K_{p,main}V_p}{J_p} + \frac{G_p}{J_p} - \frac{R_p\beta_p\dot{\theta}}{J_p} - \frac{R_c m_{body}gsin(\theta)}{J_p}$$
(3.18a)

$$\ddot{\psi} = \frac{K_{y,tail}V_y}{J_y} - \frac{G_y}{J_y} - \frac{R_y\beta_y\dot{\psi}}{J_y}$$
(3.18b)

We can now write Equation 3.18 in state-space form.

$$\begin{bmatrix} \dot{\theta} \\ \dot{\psi} \\ \ddot{\theta} \\ \ddot{\psi} \end{bmatrix} = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & -\frac{R_p \beta_p}{J_p} & 0 \\ 0 & 0 & 0 & -\frac{R_y \beta_y}{J_y} \end{bmatrix} \begin{bmatrix} \theta \\ \dot{\psi} \\ \dot{\theta} \\ \dot{\psi} \end{bmatrix} + \begin{bmatrix} 0 & 0 \\ 0 & 0 \\ \frac{K_{p,main}}{J_p} & 0 \\ 0 & \frac{K_{y,tail}}{J_y} \end{bmatrix} \begin{bmatrix} V_p \\ V_y \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ \frac{G_p}{J_p} - \frac{R_c m_{body} gsin(\theta)}{J_p} \\ -\frac{G_y}{J_y} \end{bmatrix}$$
(3.19)

At this point, we shall make a few more assumptions to obtain a linearized model of the 2-DOF helicopter. ADP, in theory, should be able to withstand the inconsistencies between

the linearized model and the physical system. Let us assume that we do not the directions of the coupling forces acting on each of the rotors. This means we do not know the direction of G_p and G_y . Let us also assume that these torques are proportional to the voltages that generate them, V_p and V_y . This means we can set up two more equations in terms of thrust gains as we did in Equations 3.15a and 3.16a.

$$G_p = K_{p,tail} V_y \tag{3.20a}$$

$$G_p = K_{y,main} V_p \tag{3.20b}$$

Let us also use the small angle approximation method to get rid of the sinusoidal term. This means $sin(\theta) = \theta$ for small angles. Combining assumptions, we can rewrite the state-space model in a linear form.

$$\begin{bmatrix} \dot{\theta} \\ \dot{\psi} \\ \ddot{\theta} \\ \ddot{\psi} \end{bmatrix} = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ -\frac{R_c m_{body} g}{J_p} & 0 & -\frac{R_p \beta_p}{J_p} & 0 \\ 0 & 0 & 0 & -\frac{R_y \beta_y}{J_y} \end{bmatrix} \begin{bmatrix} \theta \\ \psi \\ \dot{\theta} \\ \dot{\psi} \end{bmatrix} + \begin{bmatrix} 0 & 0 \\ 0 & 0 \\ \frac{K_{p,main}}{J_p} & \frac{K_{p,tail}}{J_p} \\ \frac{K_{y,main}}{J_y} & \frac{K_{y,tail}}{J_y} \end{bmatrix} \begin{bmatrix} V_p \\ V_y \end{bmatrix}$$
(3.21)

We can calculate the rotational inertias because we know the Quanser AERO configuration.

$$J_p = I_{body} + 2I_{prop} = \frac{m_{body} L_{body}^2}{12} + 2m_{prop} r_{prop}^2$$
(3.22a)

$$J_y = I_{body} + 2I_{prop} + I_{yoke} = \frac{m_{body}L_{body}^2}{12} + 2m_{prop}r_{prop}^2 + \frac{m_{yoke}r_{fork}^2}{2}$$
(3.22b)

Finally, we can add some notations to make the state-space model a little more concise. Let us define the stiffness about the pitch axis as $K_{sp} = R_c m_{body} g$. Let us define the damping constant about the pitch axis as $D_p = R_p \beta_p$. Let us define the damping constant about the yaw axis as $D_y = R_y \beta_y$.

$$\begin{bmatrix} \dot{\theta} \\ \dot{\psi} \\ \ddot{\theta} \\ \ddot{\psi} \end{bmatrix} = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ -\frac{K_{sp}}{J_p} & 0 & -\frac{D_p}{J_p} & 0 \\ 0 & 0 & 0 & -\frac{D_y}{J_y} \end{bmatrix} \begin{bmatrix} \theta \\ \dot{\psi} \\ \dot{\theta} \\ \dot{\psi} \end{bmatrix} + \begin{bmatrix} 0 & 0 \\ 0 & 0 \\ \frac{K_{p,main}}{J_p} & \frac{K_{p,tail}}{J_p} \\ \frac{K_{y,main}}{J_y} \end{bmatrix} \begin{bmatrix} V_p \\ V_y \end{bmatrix}$$
(3.23)

For conciseness, we will also rename the thrust gains. The first term of the subscript will be the rotor that is affected [pitch (main) or yaw (tail)], and the second term of the subscript will be the rotor causing the effect [pitch (main) or yaw (tail)].

$$\begin{bmatrix} \dot{\theta} \\ \dot{\psi} \\ \ddot{\theta} \\ \ddot{\psi} \end{bmatrix} = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ -\frac{K_{sp}}{J_p} & 0 & -\frac{D_p}{J_p} & 0 \\ 0 & 0 & 0 & -\frac{D_y}{J_y} \end{bmatrix} \begin{bmatrix} \theta \\ \psi \\ \dot{\theta} \\ \dot{\psi} \end{bmatrix} + \begin{bmatrix} 0 & 0 \\ 0 & 0 \\ \frac{K_{pp}}{J_p} & \frac{K_{py}}{J_p} \\ \frac{K_{yp}}{J_y} & \frac{K_{yy}}{J_y} \end{bmatrix} \begin{bmatrix} V_p \\ V_y \end{bmatrix}$$
(3.24)

The derived state-space model is the same as the one derived by Quanser documentation provided in the licensed software package resources. The only item left is to determine the constants used in the state-space model. Some of the constants can be calculated and others need to be measured. The rotational inertias and the stiffness can be be calculated by provided product features, but the damping constants and the thrust gains need to be measured. Quanser documentation provides experimental procedures to measure these values including the signs for the thrust gains. The constants calculated and measured by Quanser for the Quanser AERO can be seen in Table 3.2. With the parameters known, we

Parameter	Value	Unit
J_p	0.0215	$[\mathrm{kg}\mathrm{m}^2]$
J_y	0.0237	$[\mathrm{kg}\mathrm{m}^2]$
$D_p = R_p \beta_p$	0.00711	$[N \mathrm{m}\mathrm{s}\mathrm{rad}^{-1}]$
$D_y = R_y \beta_y$	0.0220	$[N \mathrm{m}\mathrm{s}\mathrm{rad}^{-1}]$
$K_{\rm sp} = R_{\rm c} m_{\rm body} g$	0.0375	$[N \mathrm{m}\mathrm{rad}^{-1}]$
$K_{\rm pp}$	0.0011	$[{ m NmV^{-1}}]$
$K_{ m py}$	0.0021	$[\mathrm{NmV^{-1}}]$
$K_{\rm yy}$	0.0022	$[{ m NmV^{-1}}]$
$K_{ m yp}$	-0.0027	$[\mathrm{NmV^{-1}}]$

Table 3.2: Parameters calculated and measured by Quanser.

can calculate the numeric state-space model of the Quanser AERO configured as a 2-DOF helicopter as defined the Quanser measurements.

$$\begin{bmatrix} \dot{\theta} \\ \dot{\psi} \\ \ddot{\theta} \\ \ddot{\psi} \end{bmatrix} = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ -1.7442 & 0 & -0.3307 & 0 \\ 0 & 0 & 0 & -0.9283 \end{bmatrix} \begin{bmatrix} \theta \\ \psi \\ \dot{\theta} \\ \dot{\psi} \end{bmatrix} + \begin{bmatrix} 0 & 0 \\ 0 & 0 \\ 0.0512 & 0.0977 \\ -0.1139 & 0.0928 \end{bmatrix} \begin{bmatrix} V_p \\ V_y \end{bmatrix}$$
(3.25)
Chapter 4

MATLAB Simulations

In any electrical engineering project, one must perform theoretical research which includes a mathematical analysis before they can continue to implementation. That research was performed in the previous chapters. The next step is combining the theoretical aspect with the defined system, the Quanser AERO. We cannot go straight to implementation because we need to verify that the ADP algorithm does, in fact, behave as intended for the Quanser AERO. In our case, the most viable option was to perform MATLAB simulations using the derived models discussed in Chapter 3 and the ADP algorithm discussed in Chapter 2.

The MATLAB simulation code can be seen in Appendix A.1. The MATLAB simulation can be ran for both the half-quadcopter and the 2-DOF helicopter configurations. Both ADP and LQR are ran in a single simulation to compare the results of the two control methods. The desired trajectory can be set, and then both control approaches are ran for the state-space model.

A few assumptions where made when writing the MATLAB simulation code, and these assumptions can be seen in the code provided. Parameters discussed in Chapter 3 were used to describe the state-space models. No measurements were used from our physical Quanser AERO. Quanser documentation also provided reasonable matrices for **R** and **Q** used in the LQR control approach. These matrices were adjusting slightly to make them positive semi-definite and then used for both ADP and LQR. Sampling times of $\tau = 0.01$ and T = 0.2 seconds were used for timing. These constraints were provided by Dr. Miah at the beginning of the project.

The disadvantage of using MATLAB is the state-space models used for the system are linear. We derived linear state-space models in Chapter 3. We cannot simulate a truly nonlinear system for the Quanser AERO because the coupling is described as linear. In essence, the MATLAB simulations only provide a comparison of how ADP and LQR control a linear system. One would suspect both of them to perform well given the fact that LQR is designed for linear systems. The real question is whether ADP will out-perform LQR for a linear system. To try to mimic a nonlinear system, the MATLAB simulation code has a

nonlinear adjustment factor. This factor adds a random few degrees to the output of the system every sampling time. The idea of this factor is to try to mimic some pseudo-random noise or nonlinearity to the system. This factor is adjustable, but it is currently set to zero in Appendix A.1.

For both the half-quadcopter and 2-DOF helicopter, three MATLAB simulations are provided. The initial conditions are the same for all of the MATLAB simulations. The first MATLAB simulation is setting the desired pitch and yaw to a constant value. The second MATLAB simulation imposes time-varying constant trajectories for both the desired pitch and yaw. The last MATLAB simulation provided imposes sinusoidal desired trajectories for both the pitch and yaw. The desired angular velocities are always zero.

This chapter is organized as follows. Section 4.1 provides and analyzes the MATLAB simulation results for the half-quadcopter. MATLAB simulation results for the 2-DOF helicopter are provided and analyzed in Section 4.2.

4.1 Half-Quadcopter

4.1.1 Results

The results of the MATLAB simulations for the half-quadcopter can be seen in Figures 4.1, 4.2, and 4.3.

4.1.2 Analysis

The MATLAB simulation results for constant desired pitch and yaw can be seen in Figure 4.1. The angular positions for both ADP and LQR can be seen in Figure 4.1(a). The MATLAB simulation shows that both control techniques have convergent pitch and yaw values. It does appear that the ADP technique does oscillate initially slightly more than LQR, but ADP does reach the desired values. The same results can be seen with the angular velocities shown in Figure 4.1(b). The voltage inputs seen in Figure 4.1(e) show similar traits for ADP and LQR, but no technique has a distinct advantage.

For the time-varying constant desired trajectories, the MATLAB simulations can be seen in Figure 4.2. The angular position seen in Figure 4.2(a) shows faster convergence for LQR and large initial oscillations for ADP. The angular velocities seem to have difficulty converging to zero because of the changing trajectories, but both ADP and LQR have similar traits as seen in Figure 4.2(b). Notice that the voltage inputs of opposite motors mirror each other. Again for the voltage inputs, there is no clear advantage for either ADP or LQR.



Figure 4.1: Constant desired pitch and yaw for half-quadcopter. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.

The most surprising MATLAB simulation is the sinusoidal desired trajectory as seen in Figure 4.3. For the angular position, we see the initial oscillations from ADP, while LQR



Figure 4.2: Time-varying constant desired pitch and yaw for half-quadcopter. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.

seems to converge quicker which can be seen in Figure 4.3(a). The difference with this MATLAB simulation is that ADP actually appears closer to the desired trajectory than



Figure 4.3: Sinusoidal desired pitch and yaw for half-quadcopter. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.

LQR. This can be verified in the angular position error plot seen in Figure 4.3(c). As expected, the angular velocity is no where close to zero because we have a changing desired

trajectory. Also, there is again no clear advantage for either ADP or LQR in terms of input voltage as seen in Figure 4.3(e).

Overall, both ADP and LQR perform well for the linear system of the half-quadcopter as seen in the MATLAB simulation results. We usually see more initial oscillation from ADP than LQR, but this might be the algorithm learning the system. Both control techniques, however, converge to the desired trajectories. On the down side, there is no clear advantage to ADP over LQR for a linear system as hoped for in the beginning of this project.

4.2 2-DOF Helicopter

4.2.1 Results

The results of the MATLAB simulations for the 2-DOF helicopter can be seen in Figures 4.4, 4.5, and 4.6.

4.2.2 Analysis

The MATLAB simulation results for constant desired pitch and yaw can be seen in Figure 4.4. The angular positions for both ADP and LQR can be seen in Figure 4.4(a). The MATLAB simulation shows that both control techniques have convergent pitch and yaw values. It does appear that the ADP technique does oscillate initially much more than LQR, but ADP does reach the desired values. The same results can be seen with the angular velocities shown in Figure 4.4(b). The voltage inputs seen in Figure 4.4(e) show similar traits for ADP and LQR, but no technique has a distinct advantage.

For the time-varying constant desired trajectories, the MATLAB simulations can be seen in Figure 4.5. The results of the time-varying trajectories are very similar to those of the constant desired trajectories. The angular position seen in Figure 4.5(a) shows faster convergence for LQR and slight initial oscillations for ADP. This is exactly what we saw in the constant desired trajectories. The angular velocities seem to have difficulty converging to zero because of the changing trajectories, but both ADP and LQR have similar traits as seen in Figure 4.5(b). Again for the voltage inputs, there is no clear advantage for either ADP or LQR.

The most surprising MATLAB simulation is the sinusoidal desired trajectory as seen in Figure 4.6. For the angular position, we see the initial oscillations from ADP, while LQR seems to converge quicker which can be seen in Figure 4.6(a). The difference with this MATLAB simulation is that ADP actually appears closer to the desired trajectory than LQR. This can be verified in the angular position error plot seen in Figure 4.6(c). As expected, the angular velocity is no where close to zero because we have a changing desired trajectory. Also, there is again no clear advantage for either ADP or LQR in terms of input voltage as seen in Figure 4.6(e).



Figure 4.4: Constant desired pitch and yaw for 2-DOF helicopter. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.

Overall, both ADP and LQR perform well for the linear system of the 2-DOF helicopter as seen in the MATLAB simulation results. We usually see more initial oscillation from ADP



Figure 4.5: Time-varying constant desired pitch and yaw for 2-DOF helicopter. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.

than LQR, but this might be the algorithm learning the system. Both control techniques, however, converge to the desired trajectories. On the down side, there is no clear advantage



Figure 4.6: Sinusoidal desired pitch and yaw for 2-DOF helicopter. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.

to ADP over LQR for a linear system as hoped for in the beginning of this project.

Chapter 5

Real-Time Simulation

V-REP, Virtual Robot Experimentation Platform, is a program made by Coppelia Robotics that offers an integrated development environment for prototyping robots and testing control algorithms on different robots among many other capabilities.

On the most basic level, V-REP offers an environment, called a "scene," in which the user places components of a robot, such as joints, sensors, grabbers, and basic shapes. When implemented, these items are called "scene objects." Each scene object can be assigned code called by V-REP at each time step in a simulation. The code, written in the LUA language, allows the user to use V-REP's built-in API to have scene objects interact, governed by the different physics engines available within V-REP.

V-REP can be run solely by internal LUA code, and many of the provided robots come with some short script attached to each robot demonstrating functionality. One of the deliverables for this project is to have V-REP be controlled remotely by MATLAB.

This chapter is organized as follows. Section 5.1 discusses the software used throughout this project pertaining to V-REP. Section 5.2 discusses the data transfer between MATLAB and V-REP. Section 5.3 discusses the design of the Quanser AERO in V-REP. Section 5.3.1 discusses the design of a working Quanser AERO model, and Section 5.3.2 discusses a model still under development. Section 5.4 discusses the results including RMSE measurements and the plots from simulations performed throughout the project.

5.1 Required Software

At the time of this publication, the software simulations were performed using V-REP PRO EDU V3.5.0.4 and MATLAB 2016B. It is not required, but it is recommended to use the Vortex Physics engine to control the V-REP simulation.

NOTE: Different versions of V-REP may affect model functionality.



Figure 5.1: Communication scheme between MATLAB (acting as controller) and V-REP (acting as environment).

5.2 Communication

V-REP and MATLAB communicate through a communication thread defined at port 19999. This process is shown in Figure 5.1. Next, the V-REP API offers a function to package a MATLAB vector and send that through the port for V-REP to read. Once V-REP reads the signal, it clears the signal, reading it for the next MATLAB assertion. A simple MATLAB and LUA setup are provided in Appendices B.2.1 and B.2.2.

5.3 Designing Quanser AERO

The V-REP design started with considering the deliverables. The initial design was to use the V-REP quadcopter model provided. To adapt this model to the Quanser AERO system, the center was to be set at a fixed point, while two of the four propellers were to be used to actuate the system. This plan was modified to a two pronged approach at designing and implementing a Quanser AERO base, each utilizing different API function calls within V-REP. The first design will make use of the linearized model and exert control in V-REP via API calls concerning the pitch and yaw joints, configured as motors. The second model will make use of the motor circuit to calculate thrust and resultant torques.

Clearly, factors such as the weight, size, and location of each shape that comprise the V-REP model are important to accurately define the **A** and **B** matrices of the statespace model. The model uses only cylinders and rectangular prisms such that algebraic moments of inertia could be utilized for convenience. To compute the moment of inertia of a cylinder, Equation 5.1 is used. To compute the moment of inertia of a rectangular prism, Equation 5.2 is used. To compute the moment of a cylinder or rectangular prism that is off of its axis of rotation by a distance d, the Parallel Axis Theorem, Equation 5.3, is used.

$$I_{cyl} = m * r^2 \tag{5.1}$$

$$I_{prism} = \frac{m}{12}(a^2 + b^2)$$
(5.2)

$$I_{PAX} = I + md^2 \tag{5.3}$$

Table 5.1 displays the similarity of the moment of inertia about θ and ψ for V-REP and the Quanser documentation values. Code to calculate the moment of inertia for the V-REP Quanser AERO in yaw and pitch is included in Appendix B.3.

Parameter	Source	Value	Unit
$J_{ heta}$	Quanser	0.215	$[\mathrm{kg}\mathrm{m}^2]$
$J_{ heta}$	V-REP	0.0214	$[\mathrm{kg}\mathrm{m}^2]$
J_ψ	Quanser	0.237	$[\mathrm{kg}\mathrm{m}^2]$
J_ψ	V-REP	0.0244	$[\mathrm{kg}\mathrm{m}^2]$

Table 5.1: Comparing V-REP and Quanser AERO Moments of Inertia

5.3.1 Joint Control

In this approach, the V-REP model makes use of the model constants provided by Quanser. MATLAB is configured to send the optimized motor voltages decided by the control algorithm. V-REP then receives this signal, and sets the corresponding motor's input voltage in the simulation. LUA script then estimates the current angular velocities numerically by trapezoidal integration. The angular velocities are applied to the scene joints. The design of the LUA script attached to the Quanser AERO is shown in Figure 5.2.

In this design, the joints are configured as motors capable of receiving angular velocities from the V-REP API's setJointTargetVelocity function. In this configuration, V-REP will command the physics engine running the scene to apply torque to the selected joint until the target velocity is reached. V-REP will continue adjusting this torque to maintain the target velocity.

For the 2-DOF helicopter, we can write the angular accelerations as:

$$\ddot{\theta} = \frac{-K_{\rm sp}}{J_{\theta}}\theta + \frac{-D_{\theta}}{J_{\theta}}\dot{\theta} + \frac{K_{\rm pp}}{J_{\theta}}V_0 + \frac{K_{\rm py}}{J_{\theta}}V_1$$
(5.4a)

$$\ddot{\psi} = \frac{-D_{\psi}}{J_{\psi}}\dot{\psi} + \frac{K_{yp}}{J_{\psi}}V_0 + \frac{K_{yy}}{J_{\psi}}V_1$$
(5.4b)

For the half-quadcopter, we can write the angular accelerations as:

$$\ddot{\theta} = \frac{-K_{\rm sp}}{J_{\theta}}\theta + \frac{-D_{\theta}}{J_{\theta}}\dot{\theta} + \frac{-K_{\rm pp}}{J_{\theta}}V_0 + \frac{K_{\rm pp}}{J_{\theta}}V_1$$
(5.5a)

$$\ddot{\psi} = \frac{-D_{\psi}}{J_{\psi}}\dot{\psi} + \frac{-K_{yy}}{J_{\psi}}V_0 + \frac{K_{yy}}{J_{\psi}}V_1$$
(5.5b)

The LUA script for the half-quadcopter is included as Appendix B.4.1, and the 2-DOF helicopter script is included as Appendix B.4.2. Each of these scripts are attached to the base of the Quanser AERO in the V-REP scene. Table 5.2 provides the reader with information about the MATLAB script that runs each V-REP scene.

Table 5.2: File Sets				
Model	MATLAB File	V-REP Scene		
Half-quadcopter	B.4.3	AeroHalfquadcopter		
2-DOF Helicopter	B.4.4	AeroHelicopter		
Half-quadcopter	B.4.5	AeroHalfquadcopter		
2-DOF Helicopter	B.4.6	AeroHelicopter		



Figure 5.2: V-REP joint model flowchart.



Figure 5.3: Quanser AERO developed for joint control. (a) 2-DOF helicopter mode. (b) Half-quadcopter mode.

5.3.2 Motor Control

For the motor control approach, the LUA script applies a force and a torque on the Quanser AERO at the location of the motor. This will simulate the propeller's lift and the motor's counter-torque on the Quanser AERO through V-REP's selected physics engine.

V-REP joints in this model must be configured as free spinning connections and contain

no internal friction. Therefore, $D_{\theta} = D_{\psi} = 0$. Further, the model is balanced about the pitch joint, therefore, $K_{sp} = 0$. Using these values, the state-space matrices for the 2-DOF helicopter become:

$$\begin{bmatrix} \dot{\theta} \\ \dot{\psi} \\ \ddot{\theta} \\ \ddot{\psi} \end{bmatrix} = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} \theta \\ \psi \\ \dot{\theta} \\ \dot{\psi} \end{bmatrix} + \begin{bmatrix} 0 & 0 \\ 0 & 0 \\ 0.0515 & 0.0983 \\ -0.1105 & 0.0901 \end{bmatrix} \begin{bmatrix} V_0 \\ V_1 \end{bmatrix}$$
(5.6)

And for the half-quadcopter, the state-space matrices become:

$$\begin{bmatrix} \dot{\theta} \\ \dot{\psi} \\ \ddot{\theta} \\ \ddot{\psi} \end{bmatrix} = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} \theta \\ \psi \\ \dot{\theta} \\ \dot{\psi} \end{bmatrix} + \begin{bmatrix} 0 & 0 \\ 0 & 0 \\ -0.0450 & 0.0450 \\ -0.0901 & 0.0901 \end{bmatrix} \begin{bmatrix} V_0 \\ V_1 \end{bmatrix}$$
(5.7)

Figure 5.5 shows the modeling circuit of the electric motors driving each propeller in the Quanser AERO. Both the ADP and LQR will decide an optimal voltage to regulate the error states to zero, so the goal will be to calculate the lift and torque from equations provided in Quanser literature.

First, we calculate the lifting force created by the propeller. We express τ_{θ} , assuming 90[°] application:

$$\tau_{\theta} = R \cdot F_0 \sin 90 - R \cdot F_1 \sin 90 = K_{\rm pp} V_0 - K_{\rm pp} V_1 \tag{5.8}$$

Therefore:

$$F_{prop_{0,1}} = \frac{K_{\rm pp}}{R} V_{0,1} \tag{5.9}$$

Next, the torque generated against the motor by the propeller dynamics is considered. V-REP provides a local coordinate system for the motor, and the torque is applied about the Z-axis. The model defines Motor#0 as having upwards thrust with a counterclockwise spin, and Motor#1 as having upwards thrust with a clockwise spin. Likewise, when the thrust changes direction as the polarity of voltage changes, the torque vector has to switch directions.

Quanser literature expresses drag torque as:

$$\tau_{drag} = k_d \omega_m \tag{5.10}$$

And gyroscopic torque as:

$$\tau_{gryo} = k_m i_m \tag{5.11}$$

Creating a total torque about the motor's body frame Z-axis as:

$$\tau_z = k_d \omega_m + k_m i_m \tag{5.12}$$

Using Equation 5.12, the input voltage from the control algorithm, and the motor current, i_m , the torque created by the motor can be modelled. However, we don't have an expression for the motor current, i_m , or the propeller speed, ω_m .

To model the current in V-REP, a sweep was performed on the actual Quanser AERO with the Simulink model developed in Section 6.1. For this test, we enabled the yaw and pitch locks on the Quanser AERO, applied voltages in steps of 2V from -22 to 22 Volts, and logged the currents. A cubic function estimating ($R^2 = 0.997$) the motor current, given the input voltage, was generated shown in Equation 5.13. The recorded data points and the current plot are shown in Figure 5.4.

$$i(v_{0,1}) = 2.77 \times 10^{-5} v^3 + -9.66 \times 10^{-6} v^2 + 2.13 \times 10^{-2} v + -3.19 \times 10^{-3}$$
(5.13)



Figure 5.4: Measured currents and estimate function.

Using the current model allows for calculation of the torque from the drag force of the propeller and from the motor accelerating the propeller. Referencing Quanser literature, Equation 5.14 represents the back EMF, a relationship between the voltage across the motor and the motor speed.



Figure 5.5: Circuit modeling Quanser AERO motors.

$$e_b = k_m \omega_m \tag{5.14}$$

Using Figure 5.5 and KVL:

$$v_m - R_m i_m - L_m \frac{di_m}{dt} - k_m \omega_m = 0 \tag{5.15}$$

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			0
Parameter	Description	Value	Unit
R_m	Terminal Resistance	8.4	$[\Omega]$
k_t	Torque Constant	0.042	$[\mathrm{NmA^{-1}}]$
k_m	Back EMF Constant	0.042	$[V rad^{-1} s^{-1}]$
J_m	Rotor Inertia	4×10^{-6}	$[\mathrm{kg}\mathrm{m}^2]$
L_m	Rotor Inductance	0.00116	[H]
k_d	Drag Coefficient	1×10^{-5}	$[N m rad^{-1} s^{-1}]$

Leading to:

$$\omega_m = \frac{v_m - R_m i_m - L_m \frac{di_m}{dt}}{k_m} \tag{5.16}$$

Equation 5.16 represents the model to find the propeller speed, given the input voltage, and the current. Therefore, using Equations 5.13, 5.15, 5.10, and 5.11 allows for estimated forces and torques to be calculated. Each propeller's LUA script is described by the flowchart in Figure 5.3.2.

The LUA script is implemented as follows. The Quanser AERO base receives Appendix B.5.1, then each motor gets its own script, Motor#0: Appendix B.5.2, Motor#1: Appendix B.5.3. For now, the only implementation of MATLAB code is provided as an untuned PID controller, provided as Appendix B.5.4. Thus far, LQR and ADP algorithms have not been applicable to the system.



Figure 5.6: V-REP motor model flowchart.



Figure 5.7: Quanser AERO developed for motor control. (a) Half-quadcopter mode (side). (b) Half-quadcopter mode (top).

5.4 Simulation Results

Figures 5.8, 5.9, 5.10, 5.11, 5.12, and 5.13 are presented to demonstrate the V-REP joint control model performing in both 2-DOF helicopter and half-quadcopter mode with the LQR algorithm acting as the controller. The results of the RMSE of the measurements are summarized in Table 5.4.

Figures 5.14, 5.15, 5.16, and 5.17 are presented to demonstrate the V-REP joint control model performing in both 2-DOF helicopter and half-quadcopter mode with ADP acting

as the controller.

We can see that LQR performs well for this particular system. ADP under-performs LQR most notably in smoothness; however, ADP does track the referenced angular trajectory. The results are similar to the MATLAB simulations, but there are more disturbances. The results of the RMSE measurements are summarized in Table 5.5.

Figures 5.18, 5.19 and 5.20 are provided as proof of concept for the motor modeled V-REP scene. In these simulations, $K_p = 8$, $K_i = 10$, and $K_d = 3$.

5.4.1 LQR

Figure	Configuration	θ LQR [deg]	ψ LQR [deg]	$\dot{\theta}$ LQR [deg /s]	$\dot{\psi}$ LQR [deg/s]
5.8	2-DOF Helicopter	3.60063	1.42257	0.00034832	0.00082065
5.9	2-DOF Helicopter	0.984427	1.75581	0.000235371	0.000358396
5.10	2-DOF Helicopter	0.433482	0.309782	0.000207048	0.000369757
5.11	Half-quadcopter	5.03313	3.18276	0.000334748	0.00081096
5.12	Half-quadcopter	1.25609	1.15357	0.000226184	0.000364947
5.13	Half-quadcopter	0.643793	0.170658	0.000205815	0.000367207

Table 5.4: Experimental V-REP LQR angular RMSE measurements.

5.4.2 ADP

Table 5.5: Experimental V-REP ADP angular RMSE measurements.

Figure	Configuration	θ ADP [deg]	ψ ADP [deg]	$\dot{\theta}$ ADP [deg/s]	$\dot{\psi}$ ADP [deg/s]
5.14	2-DOF Helicopter	3.94776	1.655	0.000159409	0.00091622
5.15	2-DOF Helicopter	0.449811	1.58411	0.000406199	0.00139641
5.16	Half-quadcopter	6.58232	3.27046	0.00096847	0.00108425
5.17	Half-quadcopter	0.742075	1.95259	0.000410448	0.00106001

5.4.3 PID

5.4.4 Analysis

From the simulation results provided in Sections 5.4.1, 5.4.2, and 5.4.3, we can see that the V-REP models seem to be working correctly. There is, however, one difficulty associated with ADP compared to both LQR and PID control. Because of the computational



Figure 5.8: Time-varying constant desired pitch and yaw for 2-DOF helicopter. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.



Figure 5.9: Time-varying constant desired pitch and yaw for 2-DOF helicopter. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.



Figure 5.10: Time-varying constant desired pitch and yaw for 2-DOF helicopter. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.



Figure 5.11: Time-varying constant desired pitch and yaw for half-quadcopter. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.



Figure 5.12: Time-varying constant desired pitch and yaw for half-quadcopter. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.



Figure 5.13: Time-varying constant desired pitch and yaw for half-quadcopter. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.



Figure 5.14: Time-varying constant desired pitch and yaw for 2-DOF helicopter. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.



Figure 5.15: Time-varying constant desired pitch and yaw for 2-DOF helicopter. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.



Figure 5.16: Time-varying constant desired pitch and yaw for half-quadcopter. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.



Figure 5.17: Time-varying constant desired pitch and yaw for half-quadcopter. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.



Figure 5.18: Time-varying constant desired pitch and yaw for half-quadcopter with PID controller. (a) Angular position. (b) Voltage inputs.



Figure 5.19: Time-varying constant desired pitch and yaw for half-quadcopter with PID controller. (a) Angular position. (b) Voltage inputs.

complexity of ADP, latency issues arise between MATLAB and V-REP. MATLAB is not able to perform fast enough for ADP to simulate in real-time in V-REP. Options to fix this could either be optimizing ADP or using a faster language besides MATLAB. For now, it suffices to run V-REP not in real-time mode, made possible by the MATLAB calling function. The results predict that ADP will work in the V-REP simulations, but that optimization could greatly be advanced. Further research needs to be conducted to address this issue.



Figure 5.20: Time-varying constant desired pitch and yaw for half-quadcopter with PID controller. (a) Angular position. (b) Voltage inputs.

Chapter 6

Hardware Implementation

Once MATLAB simulations showed that ADP can be used as a controller for the Quanser AERO, ADP could be implemented to control the physical Quanser AERO. To accomplish this hardware implementation, we used the graphical programming language Simulink which is recommended by Quanser. Simulink is an industry-standard graphical programming language used for the purpose of C-code generation from the graphical interface. This is exactly why it has been implemented by Quanser. Our initial task was to implement ADP into a Simulink model. This is actually more difficult than it sounds because of the complexity of ADP.

The Quanser AERO can be configured with two different interface panels, the QFLEX 2 USB and the QFLEX 2 Embedded. The QFLEX 2 USB is the panel that allows the Quanser AERO to be controlled via USB from a personal computer. This panel requires specialized Quanser software which will be discussed. Quanser also provides plenty of documentation and Simulink examples for the QFLEX 2 USB panel.

The other panel provided by Quanser is the QFLEX 2 Embedded. This panel allows communication to the Quanser AERO via SPI communication. The QFLEX 2 Embedded is designed to be used with embedded devices such as a Raspberry Pi. We implemented this panel using a Raspberry Pi 3 (https://www.raspberrypi.org/products/raspberry-pi-3-model-b/) which can be seen in Figure 6.1. The problem with the QFLEX 2 Embedded panel is that Quanser does not have much documentation or any examples for this panel. All of the results that will be discussed about the Raspberry Pi 3 were made from our own research and technical experiences.

One advantage of starting from the beginning with the Raspberry Pi 3 is that we had the freedom to expand its applications once we were able to get the basics under control. This can be seen with the application of the Android smart phone application which will be discussed as well.

Our initial goal for the project was to be able to implement ADP on the Quanser AERO through various media. The options can be seen in Figure 6.2. Not all media have been achieved, but the ones completed will be described in this chapter.



Figure 6.1: Raspberry Pi 3.



Figure 6.2: From top to bottom. 1) User controls the Quanser AERO from their laptop via Wi-Fi through the Raspberry Pi 3. 2) User controls the Quanser AERO from their laptop via Ethernet through the Raspberry Pi 3. 3) User controls the Quanser AERO from their cell phone via Wi-Fi through the Raspberry Pi 3. 4) User controls the Quanser AERO via a USB connection using the QFLEX 2 USB panel.

This chapter is organized as follows. Section 6.1 discusses the Simulink model of ADP. Section 6.2 discusses the QFLEX 2 USB panel and its use in running a Simulink model on the Quanser AERO. The use of the QFLEX 2 Embedded panel is discussed in Section 6.3 along with the use of the Raspberry Pi 3 and its complexities. Section 6.4 showcases the development of a Android smart phone application created by Simulink to control the Quanser AERO via the Raspberry Pi 3.

NOTE: MATLAB 2016b and its associated Simulink version were used for the entirety of this project. Different versions may affect some of the Simulink models.

6.1 ADP Algorithm in Simulink

In order to utilize ADP on the Quanser AERO, we have to convert the ADP MATLAB code to a Simulink model. The Simulink model is essentially a block diagram of ADP. We need to use Simulink to model ADP because the QUARC software, which will be discussed later, generates C-code from the Simulink model. This C-code is what drives the QFLEX 2 USB panel.

The Simulink model of ADP can be seen in Figure 6.3. The Simulink model seen in



Figure 6.3: ADP algorithm in the form of a Simulink model.

Figure 6.3 is actually combined with the communication necessary for the QFLEX 2 USB, but we will only focus on the ADP aspects in this section.

To analyze the Simulink model shown in Figure 6.3, we can briefly describe the model clockwise starting in the top-left. Before we describe the model, it should be noted that the state-feedback gain needs to be initialized before ADP can be ran. This is accomplished in Simulink by calling an "init" function which is declared in the Model Properties. This initialization function essentially calculates the state-feedback gain using random error data. It then saves the state-feedback gain and other important values such as sampling times in the workspace. The specific initialization function can be seen in Appendix C.2. The MATLAB code is very similar to that of the MATLAB simulations because MATLAB executes this script before executing the Simulink model.

Now, observing the Simulink model in Figure 6.3, we can begin analyzing it starting with the inputs. The inputs to the Simulink model are the desired pitch and yaw and their

corresponding desired angular velocities. The inputs are setup such that they can be easily switched. Moving to right, the desired configurations are combined into a vector with the mux block. The vector then enters the state-feedback system. We can see the state-feedback gain and the system which is the the largest block in the Simulink model. The inputs are the two motor voltages which are saturated between -24 and 24 Volts by the saturation block. We saturated the voltage inputs to 18 Volts in the MATLAB simulations, but Quanser Simulink examples use a saturation level of 24 Volts, so we will implement that method.

Once the system uses the input voltages to update the position, we use feedback to find the pitch, yaw, and angular velocity error used to find new inputs. Because we are not measuring the angular velocity, we use the derivative block to find the angular velocity. The model thus far essentially represents a state-feedback system, but we have not discussed ADP yet.

ADP is the subsystem in the lower center of the Simulink model. This subsystem is only triggered when the sampling is a multiple of T. The subsystem contains the user-defined function in Figure 6.4. This user-defined function essentially contains another version of





ADP in MATLAB code. When designing the ADP Simulink model, it was determined that

it would be easier to utilize ADP MATLAB code rather than converting it to a model itself. The disadvantage is the MATLAB code must be compatible for C-code generation. The reason for this will be discussed in later sections, but for now, the ADP update MATLAB code can be seen in Appendix C.3. One can notice that certain functions used in the original MATLAB simulations had to be changed to overcome this C-code generation challenge. One example is the use of anonymous functions. It should also be noted that the error data needed to be saved for use in the ADP subsystem. To save the error data, a tapped delay block was used for each of the four states. The tapped delay allows you to specify the number of delays you would like stored in the block saved as an array.

The last subsystem yet to be discussed in the Simulink model in Figure 6.3 is the subsystem in the center. This subsystem will be used later for the QFLEX 2 USB, but it can be discussed now. The subsystem changes the color of the base LED. In the subsystem is another user-defined function based on the magnitude of the error seen in Figure 6.5. The



Figure 6.5: User-defined function to change the base LED color.

user-defined MATLAB function code can be seen in Appendix C.4. The output of the MATLAB code is the intensities of red, blue, and green values to be sent to the Quanser AERO.

The items mentioned here were the main components of ADP in the form of a Simulink model. It is actually a rather simple model since we already had the MATLAB code ready from the simulations. The model could have been made a lot more difficult if the user-defined MATLAB functions were not used. As for the other components of the model, there are multiple scopes and gotos. The gotos make signal routing much cleaner. They are also used to save data to a .mat file. This data can be used for plotting and comparing experiments.

6.2 QFLEX 2 USB

As mentioned before, the QFLEX 2 USB panel provides an interface between the Quanser AERO and a computer running QUARC which will be discussed next. The QUARC software is the software that uses the C-code generated by Simulink and makes it compatible to run on the QFLEX 2 USB panel. We have a control method set up, so the only item left is to send the actuator commands to the Quanser AERO with the use of the QFLEX 2

USB panel. Quanser provides the necessary Simulink blocks for this communication which will be discussed in the following sections.

NOTE: A tutorial is provided in Appendix C.1 on how to run a complete Simulink model for the Quanser AERO. The Simulink model must contain a control and communication method to send the actuator commands to the Quanser AERO.

6.2.1 Required Software

In order to utilize the QFLEX 2 USB panel, some extra software is needed. To obtain and use the Simulink blocks provided by Quanser, a licensed version of QUARC must be obtained. QUARC is the proprietary software of Quanser that installs the Simulink QUARC package that provides communication blocks for the QFLEX 2 USB panel. QUARC also provides additional Simulink blocks such as filters and interfaces in Simulink.

Additional MATLAB and Simulink packages must also be obtained in order for QUARC to operate correctly. The QFLEX 2 USB panel cannot understand a Simulink model; some form of C-code is needed for the Quanser AERO to operate correctly. Because we are using Simulink, we need to convert the graphical model to C-code. To do this, we must have the MATLAB Coder (https://www.mathworks.com/products/matlab-coder.html) and Simulink Coder (https://www.mathworks.com/products/simulink-coder.html) packages installed. These packages can be obtained by going to one's MATLAB. In the Home tab, click on Add-Ons. You can either use Manage Add-Ons or Get Add-Ons to find the MATLAB Coder and Simulink Coder. It should be noted that these add-ons are not free, so a MathWorks account is needed.

6.2.2 Quanser AERO Simulink Model

With the QUARC software, communication between Simulink and the Quanser AERO is made much easier. As discussed earlier, the QUARC software installs Simulink libraries and blocks supported by the Quanser AERO. For our case, only three additional blocks were needed to communicate ADP to the Quanser AERO. The three blocks can be seen in Figure 6.6. Figure 6.6 is the Simulink subsystem that depicts the physical system in Figure 6.3.

All three new blocks correlate to each other. The HIL Initialize block defines the platform we are implementing once the proper settings are entered. With the proper settings entered in the HIL Initialize block, the QUARC software knows what platform to send the Simulink signals to and from.

The HIL Write block does exactly what its name implies. The HIL Write block writes our provided signals to the Quanser AERO. For our purposes, we are only concerned with motor voltages, enabling motors, and the base color. These inputs to the block can be


Figure 6.6: Simulink subsystem used to communicate with the QFLEX 2 USB panel.

changed by updating the settings inside the block. It is recommended to also refer to the help documentation associated with this block when selecting inputs as the names are abbreviated.

The HIL Read block also does exactly what its name implies. The HIL Read block reads our specified signals from the Quanser AERO. For our purposes, we are only concerned with the pitch, yaw, and motor currents. These outputs to the block can be changed by updating the settings inside the block. It is again recommended to also refer to the help documentation associated with this block when selecting inputs as the names are abbreviated.

With these three Simulink blocks implemented in your model and with the QUARC software installed, communication between Simulink and the Quanser AERO is much simpler. A tutorial on how to run your Simulink model on the Quanser AERO is provided in Appendix C.1.

6.2.3 Results

In order to see the effectiveness of ADP, we needed to see physical measurements from the implementation results. Because we ran ADP through Simulink, we could record various measurements from the Quanser AERO in a .mat file. in Section 4. The plots we generated can be seen in Figures 6.7, 6.8, 6.9, 6.10, 6.11, 6.12, 6.13, 6.14, 6.15, 6.16, 6.17, 6.18, 6.19, and 6.20. For each of the experiments, we ran the experiment with a saturation level of either 18 or 24 Volts. The motors are rated at 18 Volts, but Quanser Simulink models run

them at 24 Volts, so we tried both voltages. In order to have a baseline comparison, we compared ADP to that of LQR. Quanser provides a Simulink model for LQR which we modified slightly to save the measurements. We then ran each test using both ADP and LQR.

It should be noted that ADP needs the **B** matrix of the state-space model to update the state-feedback gains accurately. To make sure we had an accurate **B** matrix, we measured the thrust gains. We measured the thrust gains by running the experiments provided in Quanser documentation. The documentation went through measuring and calculating the thrust gains. We then used the constants provided by Quanser to find the **B** matrix. We did not use any experiments to find the **A** matrix for our system; we used the constants provided. ADP should be able to correct if the **A** matrix is slightly off, but we could see drastic changes in the performance of ADP using the calculated **B** matrix and the one provided in Quanser documentation.

From Figures 6.7, 6.8, 6.9, 6.10, 6.11, 6.12, 6.13, 6.14, 6.15, 6.16, 6.17, 6.18, 6.19, and 6.20 we can see that both ADP and LQR perform well for this particular system. The results are similar to the MATLAB simulations, but that is not enough evidence. To see if there are any bigger differences, we calculated the root-mean-squared-error of the measurements and also the root-mean-squared values of the power of the motors. These values can be seen in Tables 6.1, 6.2, and 6.3. It should also be noted that these experiments were conducted using the inefficient propellers provided with the Quanser AERO.

Figure	θ ADP [deg]	θ LQR [deg]	ψ ADP [deg]	ψ LQR [deg]
6.7	8.794	23.135	8.565	23.308
6.8	8.463	21.205	7.777	21.219
6.9	23.088	21.612	21.144	21.289
6.10	21.478	19.727	18.944	17.942
6.11	9.414	14.833	9.595	19.207
6.12	4.661	8.424	5.925	13.371
6.13	3.787	9.728	3.375	14.155
6.14	3.143	8.885	2.964	13.622
6.15	6.162	14.185	7.116	14.171
6.16	5.973	12.910	6.533	12.812
6.17	14.496	14.939	14.220	15.049
6.18	13.816	14.074	13.368	13.594
6.19	32.126	72.588	30.188	73.744
6.20	30.408	65.609	27.501	65.348

Table 6.1: Experimental Simulink angular RMSE measurements.

From Table 6.1, we can see that ADP performs better at minimizing the error with respect to the desired pitch and yaw compared to LQR. From Table 6.2, we can see that ADP



Figure 6.7: Constant desired pitch and yaw with inputs saturated to 18 Volts. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.

does a somewhat decent job in minimizing the error with respect to the desired angular velocities. ADP only performs slightly better, but this measurement can be difficult to



Figure 6.8: Constant desired pitch and yaw with inputs saturated to 24 Volts. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.

interpret. We are changing the desired trajectory, so the desired velocities cannot always be zero as we intended. As for the power of the motors, Table 6.3 shows that there is no



Figure 6.9: Step desired pitch and yaw with inputs saturated to 18 Volts. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.

clear advantage to either ADP or LQR. The results seem to be evenly split in favor of either. This is not great news because we were hoping that ADP would yield control that



Figure 6.10: Step desired pitch and yaw with inputs saturated to 24 Volts. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.

would minimize the energy consumed. We just cannot say for sure that this occurs with the experiments we have ran here.



Figure 6.11: Sinusoidal desired pitch and yaw with inputs saturated to 18 Volts. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.



Figure 6.12: Sinusoidal desired pitch and yaw with inputs saturated to 24 Volts. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.



Figure 6.13: Constant desired pitch and sinusoidal desired yaw with inputs saturated to 18 Volts. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.



Figure 6.14: Constant desired pitch and sinusoidal desired yaw with inputs saturated to 24 Volts. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.



Figure 6.15: Sinusoidal desired pitch and constant desired yaw with inputs saturated to 18 Volts. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.



Figure 6.16: Sinusoidal desired pitch and constant desired yaw with inputs saturated to 24 Volts. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.



Figure 6.17: Sawtooth desired pitch and constant desired yaw with inputs saturated to 18 Volts. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.



Figure 6.18: Sawtooth desired pitch and constant desired yaw with inputs saturated to 24 Volts. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.



Figure 6.19: Square desired pitch and yaw with inputs saturated to 18 Volts. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.



Figure 6.20: Square desired pitch and yaw with inputs saturated to 24 Volts. (a) Angular position. (b) Angular velocity. (c) Angular position error. (d) Angular velocity error. (e) Voltage inputs.

Figure	$\dot{\theta}$ ADP [deg/s]	$\dot{\theta}$ LQR [deg/s]	$\dot{\psi}$ ADP [deg/s]	$\dot{\psi}$ LQR [deg/s]
6.7	15.338	22.159	11.205	20.706
6.8	19.229	26.309	11.923	23.233
6.9	34.616	20.821	24.156	15.431
6.10	38.509	28.835	23.992	17.961
6.11	11.556	33.329	10.588	32.736
6.12	10.892	32.345	10.616	31.863
6.13	7.162	32.494	5.905	31.616
6.14	6.929	32.827	5.726	31.761
6.15	13.835	13.742	11.990	13.107
6.16	15.412	16.558	12.239	14.713
6.17	22.027	16.116	15.976	12.396
6.18	24.220	19.828	16.577	14.104
6.19	32.972	48.646	26.456	44.624
6.20	39.173	58.155	26.308	52.175

Table 6.2: Experimental Simulink angular velocity RMSE measurements.

Table 6.3: Experimental Simulink power RMS measurements.

Figure	Pitch ADP $[W]$	Pitch LQR $[W]$	Yaw ADP $[W]$	Yaw LQR $[W]$
6.7	8.076	7.102	7.418	6.332
6.8	13.532	11.845	12.628	9.272
6.9	12.140	13.725	11.994	11.271
6.10	19.356	16.368	21.413	12.878
6.11	8.250	10.956	6.247	8.544
6.12	10.275	14.123	7.019	11.655
6.13	5.874	7.358	5.819	6.309
6.14	7.272	9.266	7.241	6.737
6.15	9.305	12.979	5.532	9.620
6.16	11.851	15.868	8.287	10.886
6.17	8.811	9.215	7.004	7.809
6.18	13.406	12.252	12.312	9.589
6.19	11.874	12.993	11.634	10.315
6.20	19.969	19.9	20.693	16.344

6.3 QFLEX 2 Embedded/Raspberry Pi

Quanser provides the QFLEX 2 USB panel with the Quanser AERO making the Quanser AERO simple and easy to use. The problem with the QFLEX 2 USB panel is that a Simulink model must be ran with the QUARC software in order to actually control the Quanser AERO. This can become an issue when multiple licenses for QUARC are not an option or an embedded device is desired as the controller. To overcome these setbacks, Quanser has developed the QFLEX 2 Embedded panel. This panel's datasheet can be found in Appendix F.3.

The QFLEX 2 Embedded panel is different from the QFLEX 2 USB panel in the fact that it uses SPI communication from an embedded device to the Quanser AERO. The SPI protocol will be discussed in more detail later in this section. The universality of the SPI protocol allows most embedded systems to be used for control of the Quanser AERO. In our case, the embedded system is the Raspberry Pi 3. The challenge with the Raspberry Pi is that we have to convert the Simulink model used with the QFLEX 2 USB panel to one compatible with a Raspberry Pi and the QFLEX 2 Embedded panel. These challenges were overcame, but much research and innovation was needed on our part which we will discuss in the following sections.

6.3.1 Raspberry Pi 3

The Raspberry Pi 3 was given to us as a design constraint for the hardware implementation. Dr. Miah chose the Raspberry Pi 3 because it is readily available, and it has the functionality for Wi-Fi. Better embedded systems could have been considered, but we implemented ADP within the design constraints of the Raspberry Pi 3. More information on the Raspberry Pi 3 can be found at https://www.raspberrypi.org/products/ raspberry-pi-3-model-b/.

6.3.2 Required Software

Associated with the Raspberry Pi is the difficulty in generating code for ADP. We basically had two options on how to implement ADP on the Raspberry Pi using the QFLEX 2 Embedded panel. The first option was to write code from scratch that was compatible with the Raspberry Pi and the SPI protocol. This could either be C, C++, Python, etc. The difficulty with this approach would be converting all of the MATLAB code into another language less "math-friendly." This could be considered a project all in its own.

Luckily for us, we found a solution that would make the task of generating the code much easier. We considered using Simulink to generate the C-code which we could then use on the Raspberry Pi. We would then have to manipulate the C-code such that the SPI protocol could be observed. This would, again, be an extreme task. After some research, we found the solution. We found two support packages that help with code generation for the Raspberry Pi family. The Raspberry Pi Support Package for MAT-LAB (https://www.mathworks.com/hardware-support/raspberry-pi-matlab.html) and the Raspberry Pi Support Package for Simulink (https://www.mathworks.com/hardware-support/raspberry-pi-matlab.html) and the Raspberry-pi-simulink.html) make communication between MATLAB/Simulink and a Raspberry Pi possible. The Raspberry Support Package for MATLAB is more intended for running code using MATLAB on the Raspberry Pi. On the other hand, the Raspberry Support Package for Simulink is intended to generate code to be ran solely on the Raspberry Pi.

These packages can be obtained by going to one's MATLAB. In the Home tab, click on Add-Ons. You can either use Manage Add-Ons or Get Add-Ons and search for Raspberry Pi. These support packages are free, but they do require some setup. The setup includes configuring an SD card for the Raspberry Pi to be used. The SD card is configured with a Linux version specified by MATLAB. To complete the setup just follow the steps provided in the prompts; It is rather simple. Once the setup is complete, you can create a Simulink model and run the generated code on a Raspberry Pi. A tutorial of this can be found in Appendix D.1.

It should also be noted that we were using MATLAB 2016b for the extent of this project. Newer versions of MATLAB correspond to newer versions of the support packages which may have additional features.

6.3.3 SPI Communication

In order to utilize the QFLEX 2 Embedded panel, transmission between an embedded system and the Quanser AERO must be achieved via SPI communication. SPI stands for serial peripheral interface. In a nutshell, SPI is the process of sending data between two devices one bit at a time. These bits can be combined to form data needed by each device. An overview of the SPI communication process can be seen in Figure 6.21.

There are four main signals to consider in SPI communication: SS (slave-select), clock, MOSI (master-in slave-out), and MISO (master-in slave-out). One of the devices is the considered the master, and the other device is considered the slave. The master generates a clock signal. This clock signal is used to synchronize communication between the master and slave devices. Essentially, one bit is passed for each period of the clock cycle. Different variations of SPI specify when a bit is starting to send and when a bit is supposed to be stable, but these specifications will not be discussed in detail in this report. The master also controls the value of the slave-select signal. This signal is active-low to specify to the slave that transmission is required. The last two signals are MOSI and MISO. MOSI is the signal generated by the master to be sent to the slave. For our case, we are only concerned with the sending the data for the motor voltages and the base colors. MISO, on the other hand, is the signal generated by the slave to be sent to the master. For our case, we are only concerned with receiving the data from the Quanser AERO corresponding to the pitch and yaw measurements.



Figure 6.21: Overview of the SPI protocol. Image provided by https://www.mathworks.com/help/supportpkg/raspberrypi/ug/support-spi-communication.html.

There are actually multiple difficulties associated with SPI communication that Quanser failed to realize. To begin, SPI communication is intended for short information; that is low numbers of bits to be sent and received. We want to keep the information short, so we can update the information much faster. This is not the case with the Quanser AERO. Appendix F.3 specifies the data sent and received between the Quanser AERO and an embedded device. We can see that for each communication we are sending 51 bytes, or 408 bits. That is a lot of information to send via SPI communication. In our opinion, this is a major design flaw of the Quanser AERO. Specifically, we are only concerned with a few bytes, but we need to send everything for the system function properly.

The only way to make our system fast enough to send all of the bytes in a reasonable time is to increase the SPI clock frequency generated by the master. This is where you can run into hardware limitations. In the next section, we will discuss that Simulink can be used to generate signals on pins of the Raspberry Pi, but this can cause some difficulties. When we generated a clock signal on the Raspberry Pi 3, we were not getting any communication between the Quanser AERO and the Raspberry Pi. We set the clock period at 2 microseconds. With this period, we can send the 51 bytes in between each data collection in ADP. This seemed reasonable, but there was no communication.

After much thought, we decided to connect an oscilloscope to the Raspberry Pi pin that was supposed to generate the clock signal. There was no clock signal. Why is that? We decided to try larger clock periods. When we reach a clock period of 100 microseconds, we actually begin to see a clock signal that is accurate. It appears that the Raspberry Pi is incapable of switching its GPIO pins at very fast rates. This seemed very odd given that the built in SPI clock of the Raspberry Pi can run in the mHz range. We, however, could not access this built in SPI clock, so we had to generate our own clock signal which was not very fast. So, we have a 10 kHz clock signal. That means we will update the motor voltages every 0.051 seconds. This is not the value of τ we anticipated in the ADP. With this type of τ , the gain of the state-feedback system is updated roughly every 1 second which is at all not ideal. This is, however, what we are dealt because of hardware limitations. This could either make or break the implementation of ADP.

There is also one more issue we found with Quanser's implementation of SPI communication. The standard is to send the least-significant bit first in the communication. By trial-and-error and some research, we determined that Quanser sends the most-significant bit first. This design is not at all conventional. This can make troubleshooting very difficult if this is not known.

6.3.4 SPI Communication in Simulink

We already knew the capabilities of Simulink in generating C-code for the QFLEX 2 USB panel. We also found a Simulink support package for the Raspberry Pi as discussed earlier. Our thought was can we combine these two aspects in Simulink to generate C-code to be used on the Raspberry Pi? The answer is yes we can. To do this, we need to take our ADP Simulink model and convert it to SPI communication that can be used on the Quanser AERO. The model used for this whole process can be seen in Figure 6.22.

There are two basic subsystem blocks in Figure 6.22. The subsystem block on the left is ADP. This subsystem block has the exact same structure used in the previous models. The structure can be seen in Figure 6.23. We kept the same structure, but we made ADP its own entity. Remember, we needed an initialization function to determine the gain for the first T seconds. We still use this approach, but we also add commands to set the timing of the SPI communication which will be discussed next. We also secure a connection to the Raspberry Pi through MATLAB and enable the GPIO pins of the Raspberry Pi. This initialization code can be seen in Appendix D.2.

The other subsystem block seen in Figure 6.22 is the SPI communication block which is zoomed in on in Figure 6.24. In Figure 6.24, we can see that the SPI communication block is only triggered on the rising-edge of a clock signal generated by a GPIO pin. We already discussed the hardware limitations of this pin earlier. The inputs to the subsystem are the calculated motor voltages and the color values for the base which were found in ADP. The outputs of the subsystem are the current pitch and yaw values.



Figure 6.22: Simulink model used to generate the code for the Raspberry Pi 3.

Inside the SPI communication subsystem block, we can see what is really happening in Figure 6.25. Figure 6.25 shows how the information to be sent and received in the SPI communication is handled. Determining what bit to send is handled by a MATLAB function which can be found in Appendix D.3. The MATLAB function follows the transmission information provided in the QFLEX 2 Embedded datasheet found in Appendix F.3. The difficulty with this is we need to keep track of what bit and byte we are currently sending and receiving. This is why we have Data Store blocks in Figure 6.25. These Data Store blocks act as global variables for the MATLAB function. The MATLAB function in Appendix D.3 takes the information about the bit and byte we are currently sending along with the voltage and color information and sets the GPIO pins of the Raspberry Pi. Please refer to Appendix D.3 for the MATLAB code and to Appendix F.3 for how we thought



Figure 6.23: Simulink subsystem used for the ADP algorithm.

out the MATLAB code.

6.3.5 Results

As for the results with the Raspberry Pi, it is quite difficult to get a numerical analysis. We uses Simulink to generate the C-code onto the Raspberry Pi. With that method, there is no way to record data that we know of. This will require further research. The only result we can get is physically looking at the motions to determine if the Quanser AERO is in the correct position.

When we generated the C-code using Simulink, we kept getting one persistent error. We had a size mismatch because the code generation made the current pitch and yaw values 1×2 matrices. We have no idea as to why these values are matrices; they should only be single values. There are two options as to why this happened. The first option is the C-code generation is just wrong. The second option is the way we wrote the MATLAB function. We could have wrote the MATLAB function in such a way that the C-code generator interpreted it as a matrix instead of a single value. This might have worked and been undetectable in MATLAB, but the C-code generator experienced problems. Either way, we do not know the cause of the error which will require further research.

To overcome the error and proceed as scheduled, we decided to take the first element of the matrix we were supposedly generating. We know this is not the best approach, but at



Figure 6.24: Simulink subsystem used specifically by the Raspberry Pi 3.

this point, we were behind schedule, and we needed some results. With that fix, Simulink was able to successfully generate the code on the Raspberry Pi. A full tutorial of how to execute the code is given in Appendix D.1. When we executed the code, we could see the results. It appears that the Quanser AERO goes to the position specified in the code. Because of our error and the hardware limitations discussed, the position is not perfect. The Quanser AERO appears to oscillate around the desired position. We believe, because of the timing, that the state-feedback gain is not being updated fast enough. The Quanser AERO overshoots the desired position, so it is forced to go back on the next update. There is really nothing else we can do given the hardware and algorithm constraints.

Even though ADP works but not very well in this case, we did show a proof of concept. The proof of concept is that we are able to control the Quanser AERO using a Raspberry Pi



Figure 6.25: Simulink subsystem used to establish communication between the Raspberry Pi 3 and the Quanser AERO.

running ADP. This means ADP should be portable to other physical systems and embedded systems. The one drawback is that the desired position of the Quanser AERO has to be preprogrammed on the Raspberry Pi.

6.4 Raspberry Pi/Android

Our original objective of the project was to control the Quanser AERO via a Raspberry Pi and smart phone. Due to the difficulties with the C-code generation for the Raspberry Pi, we were unsure we would be able to complete the smart phone objective.

It turns out that by using Simulink, we actually made the process a lot easier. Simulink is actually able to create smart phone applications that can communicate with a Raspberry Pi. We literally had everything ready except for the application on the smart phone.

But, what is the point of the smart phone application? The development of a smart phone application can show another proof of concept. What if your system to be controlled is remote or your are actually flying an unmanned aerial vehicle? You cannot preprogram the embedded system like we did in the previous section; you need some sort of communication such as a smart phone.

6.4.1 Required Software

Because we are using the same concepts that we used for generating the C-code for the Raspberry Pi, we do not need much additional software. We do, however, need something to develop the smart phone application. Just as there was a Raspberry Pi Support Package for Simulink, there is an Android Support Package for Simulink (https://www.mathworks.com/hardware-support/android-programming-simulink.html). A helpful reference is located at https://www.mathworks.com/videos/control-raspberry-pi-from-your-android-device.html. This package can be obtained by going to one's MATLAB. In the Home tab, click on Add-Ons. You can either use Manage Add-Ons or Get Add-Ons and search for Android. This support packages is free, but it does require some setup. The setup includes setting your smart phone to allow you to develop applications. Follow the prompts in the setup, and the configuration is simple.

Because we will be using Wi-Fi as our communication medium, we also need a network. We need a local network connection for both the smart phone and the Raspberry Pi. We need a local network because it is much easier than the encrypted school network.

We also need to know the IP addresses of the smart phone and the Raspberry Pi once they are connected to the network. This task can be made simpler by downloading a device manager for your smart phone. We downloaded the IP Tools application from the Google Play Store. To manage the IP addresses of the devices on the network, we set the IP addresses as static using the router settings.

6.4.2 Simulink Model for Android Applications

Once we have all of the software set up for the Android smart phone application, we can begin updating the Simulink models. The Simulink model for the Raspberry Pi does not need to be edited significantly. We only need to change how we receive the desired pitch and yaw and also what we do with the current pitch and yaw. The updated Simulink model can be seen in Figure 6.26. This is the exact same Simulink model that was previously implemented. The only difference is how we are receiving the desired pitch and yaw.

The Simulink model in Figure 6.26 makes use of the UDP Send and Receive blocks. The UDP Receive block is specific for the Raspberry Pi. The only information needed in this block is the local port of the wireless network. This port number is specified in the UDP Send block. On the outputs of the current pitch and yaw, we see the UDP Send block. We are sending the UDP packet to an Android device, but we are doing it from a Raspberry Pi, so we need to use the Raspberry Pi UDP Send block. In the UDP Send block, we specify the local port number of the wireless network we want to connect through. We also need to specify the IP address of the receiver of the UDP packet. In our case, this will be the Android smart phone's IP address.



Figure 6.26: Simulink model used to communicate with the Android smart phone.

On the other end of the Wi-Fi communication is the Android smart phone. As discussed earlier, there is an Android Support Package for Simulink. With this package, we can create Simulink models that run on Android devices. The Simulink model can essentially be an application on the smart phone. Just as we did with the Raspberry Pi, we create the model for the Android smart phone, and then we generate the code on the physical device. A tutorial of how to accomplish this is given in Appendix E.1.

In order to generate the application, we need to create the Simulink model. This Simulink model can be seen in Figure 6.27. The Android Support Package for Simulink library has the necessary blocks to create the application. In our case, we used the Android Slider block and the Android Display block. These blocks basically imply what they mean, but they do it in the Android application. The Android Slider block is used to set the desired pitch and yaw. This value is then sent from the Android UDP Send block. The Android UDP Send block specifies the port number to be used on the wireless network. The block also specifies the IP address to where to send the UDP packet. In our case, this will be the Raspberry Pi's IP address. In this Simulink model, we also utilize the Android UDP Receive block. This block receives the UDP packet of the actual pitch and yaw measurements from the Raspberry Pi model discussed earlier. This value can then be displayed on the Android application with the use of the Android Display block. This model can be generated on the Android smart phone and then ready to use with the Raspberry Pi.



Figure 6.27: Simulink model used to create the Android smart phone application.

6.4.3 Results

The results for the smart phone application are more a matter of seeing the application work correctly. The model builds correctly for the Raspberry Pi 3. The model also builds correctly for the Android smart phone, but the build process takes a while longer. It turns out that you need to be connected to a network with Internet for the build process to complete. We are unsure why this needs to be true, but it does. A tutorial of how to run the application and get everything set up can be found in Appendix E.1.

When we run the Android smart phone application, we actually see communication to the Quanser AERO. We see the desired pitch and yaw changing as the sliders are adjusted. We also see the actual pitch and yaw readings on the application as well. Overall, this accomplishes our objective.

Chapter 7

Conclusion and Future Work

The work presented in this project successfully followed electrical engineering methodologies. We were given a system and method of control for that system; our job was to make it work. Following electrical engineering methodologies, we started with the system. We derived a highly assumed and linearized model. With the system model in hand, we analyzed ADP modifying it so it would work correctly with the new model. Once we completed the mathematical analysis and modeling, we could begin simulations to see if ADP actually worked for our system.

Simulations were conducted in both MATLAB and V-REP, but linear system models limited the simulation results. For our linear system, simulations showed proper trajectory tracking. V-REP simulations took much longer to complete due to no available Quanser AERO platforms. With successful simulations, we could begin implementation.

For the implementation process, we started simple. We created ADP in the form a Simulink model which we could then apply directly to the Quanser AERO via the QFLEX 2 USB panel. With successful control of the Quanser AERO, we modified the Simulink model slightly for interfacing with the QFLEX 2 Embedded panel. We utilized Simulink support packages to expedite the process. Using Simulink, we also were able to generate C-code that functioned correctly on the Raspberry Pi and Android smart phone.

7.1 Conclusion

This project encompassed a large amount of work for only being a year long, but several conclusions can be drawn from that work. It is possible to use ADP on the Quanser AERO, but there are some drawbacks. The Quanser AERO behaves, for the most part, as a linear system; the coupling is not too significant. ADP, thus, could not be tested accurately for this system. It would be just as beneficial to use a conventional linear control technique such as LQR. The results shown in this report show similar results for both methods. In our opinion, it would be more beneficial to use LQR for this system to limit control

complexity and computational requirements. There really is no advantage to using ADP for this particular system.

Even though ADP shows no advantage for this particular system, simulations and implementation were successfully completed for the Quanser AERO. We were able to simulate the Quanser AERO in MATLAB using the system model. The MATLAB simulations were limited in the fact that it is difficult to simulate the nonlinear coupling. V-REP also provided accurate simulation results for LQR, but the complexity of ADP made real-time simulation almost impossible.

For implementation, we conclude that it is possible to apply ADP directly to a Quanser AERO using Simulink or a Raspberry Pi. Implementing the Raspberry Pi is a feat due to the fact that Quanser provides very limited documentation on the Raspberry Pi interfacing. We were able to successfully implement ADP on the Raspberry Pi with SPI communication to the Quanser QFLEX 2 Embedded panel. ADP tracks the desired trajectory reasonably well but limitations noted suggest why perfect tracking is not achieved. Extending our Simulink models and capabilities, we were also able to successfully implement an Android smart phone application that communicates with the Raspberry Pi.

Overall, we conclude that it is possible to modify, simulate, and implement ADP on the Quanser AERO.

7.2 Future Directions

Even though we accomplished many objectives in this project, there is still room for improvement and further research. Further research can be broken down into the following categories: system, ADP, and implementation. A better system can be used to test ADP. The Quanser AERO, for the most part, was a linear system. We were not able to see the true capabilities of ADP with such a linear system.

ADP itself can also be researched further. Because our project encompassed so many objective, we were unable to fully validate ADP. The use of the actor neural network can be researched such that the system can be truly learned by the algorithm. We needed an approximate system model in order to utilize ADP. We were also unable to research the optimal sampling times for ADP. We were given these sampling times at the beginning of the project, but are they truly optimal? Lastly, ADP is very slow because of the recursive nature of the algorithm. This was especially detrimental in the V-REP simulations. If ADP could be made faster or streamlined, simulations and implementation code could run much faster and more efficient.

Implementation can also be researched further. We were constrained to the Raspberry Pi, but can other, simpler embedded systems use ADP. With the Raspberry Pi, we were

limited in the switching frequency of the GPIO pins. Further research could determine if there are workarounds to this or if we are truly limited. Above all, further research is needed in determining the error we received from the C-code generator in Simulink. We outputted a matrix instead of a single value, but we do not know the cause of this error. Was it code or compiler error? Or both? We were also limited in our ability to switch the desired trajectory once the C-code was generated for the Raspberry Pi. Further Linux research can be performed to determine alternate methods to change this once the code has been generated.

Our research may have come to an end for this project, but there is plenty more to discover.

APPENDICES

Appendix A

MATLAB

A.1 ADP Simulation (MATLAB)

```
1 % Andrew Fandel
_{2} % Use exact the linearized model to find error data points
3 % Use the error data points directly in the neural network instead of
4 % randomizing the error
5 close all; clear; clc;
6
7 testName = 'heli_sineWave';
 \text{heli} = 0; 
9 halfq = 1;
11 [A, B] = getAeroAB(heli);
12 % 2–DOF QUANSER HELICOPTER PARAMETERS
13 % Maximum applied voltage for the rotor motors
_{14} \text{ maxVolt} = 18;
15 % Number of state variables – theta, psi, dtheta, dpsi
16 n = 4;
_{17} theta = 1;
                 % Pitch
                 % Yaw
18 \text{ psi} = 2;
19 thetaDot = 3; \% Pitch angular velocity
20 psiDot = 4; % Yaw angular velocity
21 % Number of inputs - Vp, Vy
22 m = 2;
23
24 % SIMULATION TIMING
25 % Initial and final simulation times [s]
26 t0 = 0; tf = 25;
27 % Sampling time [s]
_{28} tau = 0.01;
29 % Larger sampling time for updating the inputs [s]
_{30} T = 0.2;
31 % Number of time steps
_{32} tsteps = floor ((tf-t0)/tau);
33 % Discrete time vecotor of sampling time (tau)
_{34} dt = tau * (0: tsteps);
```

```
35 % Number of equations for actor-critic neural network
36 % Number of training samples per T
_{37} nbar = floor (T/tau);
38
39 % COST FUNCTION
40 \% Q and R matrices used in the cost function
^{41} %Q_Mat_ADP = diag([270 100 1 1]);
_{42} %R_Mat_ADP = 0.0001 * diag([1 1]);
43 % Use the matrices used in Quanser documentation except add ones for the
_{44}~\% velocities -> need Q to be positive definite
_{45} Q_Mat_ADP = diag([200 75 1 1]);
46 R_Mat_ADP = 0.005 * \text{diag}([1 \ 1]);
47 % Modified LQR cost matrices
<sup>48</sup> Q_Mat_LQR = diag([200 \ 75 \ 1 \ 1]);
49 R_Mat_LQR = 0.005 * \text{diag}([1 \ 1]);
50 nonLinearAdjustment = 0;
51 % Name of .mat file to save the data
53
54 % INITIAL STATES
xInit = [deg2rad(-87) deg2rad(-45) (pi/180)*10 (pi/180)*7]';
56 \% x Init = \begin{bmatrix} 0 & 0 & 0 & 0 \end{bmatrix}';
57
58 % MATRIX INITIALIZATION
59 % Actual states - state variable x
60 xADP = zeros(n, tsteps+1);
_{61} xLQR = zeros (n, tsteps+1);
62 % Initialize x with initial conditions
^{63} xADP(:,1) = xInit;
_{64} xLQR(:,1) = xInit;
65 % Desired states
66 % Uncomment for desired states of all zero
_{67} %xd = zeros(n, tsteps+1);
_{68} % Uncomment for desired states that switch constant values
^{69} %xd = [deg2rad(37)*ones(1,tsteps+1); deg2rad(80)*ones(1,tsteps+1); zeros(1,
           tsteps+1; zeros(1, tsteps+1)];
70 \% d = [deg2rad(10) * ones(1,1000), deg2rad(60) * ones(1,tsteps+1-1000); deg2rad(
           (80) * ones (1,1500), deg2rad (-120) * ones (1, tsteps+1-1500); zeros (1, tsteps+1)
            ; zeros(1, tsteps+1);
71 % Uncomment for desired states that are sine and cosine waves
pitchSine = deg2rad(60) * sin(0.5 * dt);
    yawCosine = deg2rad(90) * \cos(0.5 * dt);
73
74 % pitchSquare = deg2rad(57) * square(0.5 * dt);
_{75} % yawSquare = deg2rad(-36)*square(0.25*dt);
xd = [pitchSine; yawCosine; zeros(1, tsteps+1); zeros(1, tsteps+1)];
77 \% xd = [pitchSquare; yawSquare; zeros(1,tsteps+1); zeros(1,tsteps+1)];
78 % State errors for entire simulation
79 errorADP = zeros(n, tsteps+1);
so errorLQR = zeros(n, tsteps+1);
81 % Initialize the error
s2 errorADP(:,1) = xd(:,1) - xADP(:,1);
s errorLQR(:,1) = xd(:,1) - xLQR(:,1);
84 % Actual inputs
uADP = zeros(m, tsteps+1);
```

```
uLQR = zeros(m, tsteps+1);
87
88 % CALCULATE THE CONTROLLER GAIN MATRIX USING LQR
89 fprintf('WORKING ON LQR GAIN MATRIX...\n');
90 kLQR = lqr(A, B, Q_Mat_LQR, R_Mat_LQR);
91
92 % FIND THE CRITIC NEURAL NEIWORK WEIGHTS BEFORE APPLYING ANY INPUTS
93 % Calculate the critic weights
94 wcInit = quanserAEROCriticTuningInitial(A,B,((2*pi).*rand(4,nbar)-pi),tau,
      R_Mat_ADP, Q_Mat_ADP, xd(:, 1);
95 % Use the weights to determine the P matrix
96 P_{-}Mat = [wcInit(5) wcInit(6) wcInit(7) wcInit(8);
            wcInit(6) wcInit(9) wcInit(10) wcInit(11);
97
            wcInit(7) wcInit(10) wcInit(12) wcInit(13);
98
            wcInit(8) wcInit(11) wcInit(13) wcInit(14)];
99
100 % Use P to determine the state-feedback gain
101 kADP = 0.5 * (R_Mat_ADP^{-1}) *B' * P_Mat;
103 % RUN THE SIMULATION
104 % Start at the next time tau after time zero
  for k = 2:tsteps+1
      % Current time
106
       t = (k-1)*tau; % generate value discrete time index
       if (mod(t,5) == 0)
108
          \% Print the current time
           fprintf('Current time, t = \% g [s] \setminus n', t);
      end
111
      % EXACT 2-DOF HELICOPTER MODEL
      \% Take previous values to find the derivative of the exact model
114
      xdotNonLinearADP = A*xADP(:, k-1) + B*uADP(:, k-1);
      xdotNonLinearLQR = A*xLQR(:, k-1) + B*uLQR(:, k-1);
      % Update the states of the exact model
      \% Use Euler integration to estimate the current states of the nonlinear
118
      % model
      xADP(:,k) = xADP(:,k-1) + xdotNonLinearADP*tau + nonLinearAdjustment
120
      .*(2.*rand(4,1)-1);
      xLQR(:,k) = xLQR(:,k-1) + xdotNonLinearLQR*tau + nonLinearAdjustment
      .*(2.*rand(4,1)-1);
      % ADD SOME DISTURBANCE TO THE STATES
      % This forces the states to some other position
123
      % if ((t > 15) \&\& (t < 17))
124
          % ADP(:,k) = [rad2deg(-45) 0 0 0]';
           %xLQR(:,k) = [rad2deg(-45) 0 0 0]';
126
      %end
      % Force the pitch angle to be in the range [-pi/2, pi/2] because of
128
      % physical constraints
      xADP(theta, k) = angleLimiterPitch(xADP(theta, k));
130
      xLQR(theta, k) = angleLimiterPitch(xLQR(theta, k));
      % Force the yaw angle to be in the range [-pi, pi] because the yaw is
      % free to do a complete circle
133
      xADP(psi,k) = angleLimiterYaw(xADP(psi,k));
134
      xLQR(psi,k) = angleLimiterYaw(xLQR(psi,k));
136
```

```
% FIND THE ERROR BETWEEN THE EXACT STATES AND THE DESIRED STATES
       errorADP(:,k) = xd(:,k) - xADP(:,k);
138
       \operatorname{errorLQR}(:, k) = \operatorname{xd}(:, k) - \operatorname{xLQR}(:, k);
140
       % UPDATE THE CRITIC WEIGHTS EVERY T TIME
141
       % Determine if the time is a multiple of T
       if (mod(t,T) == 0)
143
           % Update the critic weights
144
           wc = quanserAEROCriticTuning(A, B, errorADP(:, (k-nbar):k), tau,
145
      R_Mat_ADP, Q_Mat_ADP, xd(:,k), wcInit);
           % Use the weights to determine the P matrix
146
           P_{-}Mat = [wc(5) wc(6) wc(7) wc(8);
147
                      wc(6) wc(9) wc(10) wc(11);
148
                      wc(7) wc(10) wc(12) wc(13);
149
                      wc(8) wc(11) wc(13) wc(14);
           % Use P to determine the state-feedback gain
           kADP = 0.5 * (R_Mat_ADP^--1) *B' * P_Mat;
       end
       % Update the inputs using ADP
154
       uNewADP = kADP * errorADP(:, k);
       % Limit the voltages
156
       uNewADP(1) = sign(uNewADP(1)) * min(abs(uNewADP(1)), maxVolt);
       uNewADP(2) = sign(uNewADP(2)) *min(abs(uNewADP(2)), maxVolt);
158
       % Update the input matrices
159
       uADP(1,k) = uNewADP(1);
160
       uADP(2,k) = uNewADP(2);
       % Update the inputs using LQR
162
       uNewLQR = kLQR * errorLQR(:, k);
163
       % Limit the voltages
164
       uNewLQR(1) = sign(uNewLQR(1)) * min(abs(uNewLQR(1)), maxVolt);
       uNewLQR(2) = sign(uNewLQR(2)) * min(abs(uNewLQR(2)), maxVolt);
166
       % Update the input matrices
       uLQR(1,k) = uNewLQR(1);
168
       uLQR(2,k) = uNewLQR(2);
170 end
171
172 % SAVE OUTPUTS FOR PLOTTING
173 % Save the time, error, input, states and desired states vectors
save ([testName, '.mat'], 'dt', 'errorADP', 'uADP', 'xADP', 'xd', 'tau', 'R_Mat_ADP',
       'Q_Mat_ADP', 'errorLQR', 'uLQR', 'xLQR', 'R_Mat_LQR', 'Q_Mat_LQR');
175 % Notify that the simulation is complete
176 disp ('SIMULATION COMPLETE');
177
178 % ANGLE LIMITER FOR YAW FUNCTION
  function angle = angleLimiterYaw(angle)
179
  % This function limits the angle between -pi and pi
180
       angle = mod(angle, 2*pi);
181
182
       i=find (angle>pi);
183
       angle(i) = angle(i) - 2*pi;
184
185
       i=find (angle <-pi);
186
       angle(i) = angle(i) + 2*pi;
187
188 end
```
```
189
190 % ANGLE LIMITER FOR PITCH FUNCTION
<sup>191</sup> function angle = angleLimiterPitch(angle)
192 % This function limits the physical constraint of the pitch measurement to
_{193} % 90 degrees
       if (angle < 0)
194
            angle = \max(\text{angle}, -\text{pi}/2);
195
       end
196
197
       if (angle > 0)
198
            angle = min(angle, pi/2);
199
       end
200
201 end
202
203 % CRITIC WEIGHT TUNING NEURAL NETWORK
204 % This function is specific to the helicopter because of the number of
205 % weights and error model
  function weights = quanserAEROCriticTuningInitial(A,B,e_vec,tau,R_Mat,Q_Mat,
206
      xd)
       % REFERENCE DR. MIAH'S PAPER FOR EQUATION NUMBERS
207
208
       % System dimensions specific for our model
209
       [n, \tilde{}] = size(B);
210
211
       % ERROR MODEL OF THE HELICOPTER
       \% fbar and gbar — EQ 8
213
       fbar = @(e) A * e;
214
       gbar = -B;
215
       hbar = -A*xd;
       % DISCRETE-TIME ERROR MODEL FOR TIME TAU
218
       \% f and g — EQ 10
219
       f = @(e) fbar(e) *tau + e;
       g = gbar * tau;
221
       h = hbar*tau;
222
223
       % COST FUNCTION PARAMETERS
224
       % State penalizing function in the continuous cost function
       Qbar = @(e) e' * Q_Mat * e;
226
       % Control penalizing matrix in the continuous cost function
       Rbar = R_Mat;
228
       % The discrete-time cost function will have terms:
229
       % Right after EQ 11 in paper
230
       % State penalizing function in the discretized cost function
231
       Q = @(e) Qbar(e) *tau;
232
       % Control penalizing matrix in the discretized cost function
233
       R = Rbar*tau;
234
       % NEURAL NEIWORK FUNCTIONS
236
       % Critic neural network activation functions
237
       rho = @(e) [e(1); e(2); e(3); e(4); ...
238
                     e(1)^{2}; e(1) * e(2); e(1) * e(3); e(1) * e(4); \dots
                     e(2)^{2}; e(2) * e(3); e(2) * e(4); e(3)^{2}; e(3) * e(4); e(4)^{2};
240
       % Partial derivative of rho with respect to e
241
```

```
drhode = @(e) [1, 0, 0;
242
                      0, 1, 0, 0;
243
                      0, 0, 1, 0;
244
                      0, 0, 0, 1;
245
                      2 * e(1), 0, 0, 0;
246
                      e(2), e(1), 0, 0;
                      e(3), 0, e(1), 0;
248
                      e(4), 0, 0, e(1);
249
                      0, 2 * e(2), 0, 0;
250
                      0, e(3), e(2), 0;
251
                      0, e(4), 0, e(2);
252
                      0, 0, 2*e(3), 0;
253
                      0, 0, e(4), e(3);
254
                      0, 0, 0, 2 * e(4)];
255
256
       % TOLERANCES
257
       % Convergence tolerance for control policy
258
       EpsilonPolicy = 0.1;
       % Convergence tolerance for critic neural network
260
       EpsilonWcritic = 0.1;
261
262
       % TRAINING PARAMETERS
263
       % Number of outer loop iterations
264
       outerLoopMax = 700;
265
       % Number of inner loop iterations
266
       innerLoopMax = 100;
267
       \% Number of equations needed for training, number of sub-intervals
268
       % Number of training samples
269
       [, nbar] = size(e_vec);
       % WEIGHT INITIALIZATION
272
       % Initialize the weights of the critic neural network to zero
       WcLast = zeros(length(rho(e_vec(:,1))),1);
274
275
       % LEAST-SQUARES COMPUTATION INITIALIZATION
276
       \% Matrices required for computing least squares weights of the critic
277
       % neural networks --- EQ 20
278
       V = zeros(nbar, 1);
       Lambda = zeros(nbar, length(rho(e_vec(:,1))));
280
281
       \% Matrix to hold the derivative of the error model during policy
282
       % updating
283
       e_k_plus_1 = zeros(n, nbar);
284
285
       % Product of the least squares matrices must be invertible
286
       % Logic flag indicating if the critic weights are unsolvable
287
       \% The weights are unsolvable because the least squares matrices have no
288
       % solution --- not invertible
289
       diverged = 0;
290
291
       % OUTER LOOP
292
       for i = 1:(outerLoopMax - 1)
293
       % Determine if the least squares matrices are invertible
294
295
       if diverged = 0
```

```
% For each of the data collection (discrete time index)
296
            for k = 1:nbar
297
                % Initialize the optimal inputs to zero
298
                uNew = [0; 0];
299
                % INNER LOOP
300
                for j = 1:(innerLoopMax - 1)
301
                    % Get the updated input value
302
                    uLast = uNew;
303
                    % Update the error model
304
                     e_k_plus_1(:,k) = f(e_vec(:,k)) + g*uLast + h;
305
                    % Compute the new optimal inputs
306
                    uNew = -0.5*(R^{(-1)})*g'*drhode(e_k_plus_1(:,k))'*WcLast;
307
308
                    % Check convergence of the optimal inputs
309
                     if norm(uNew - uLast) < EpsilonPolicy
                         break;
311
                    end
312
                end
313
314
                % Update the values for the least-squares computation
315
                V(k,:) = Q(e_vec(:,k)) + uNew'*R*uNew + WcLast'*rho(e_k_plus_1)
316
       (:, k));
                Lambda(k,:) = rho(e_vec(:,k))';
317
           end
318
       end
319
       \% Verify the least square solution exists for the critic's weights
321
       \% If the error data is consistent or there is no error, this will not
322
       % hold, so set the weights to zero
       if det (Lambda' * Lambda) == 0
324
            fprintf ('AWESOME...YOU HAVE NO ERROR...I''M GOING TO SET THE WEIGHTS
325
       TO ZERO(n');
            weights = zeros(length(rho(e_vec(:,1))),1);
326
           break:
327
       end:
328
       \% Calculate least squares solution of critic's weights — EQ 20
       WcNew = (Lambda'*Lambda)^{(-1)}Lambda'*V;
331
       % Make sure the weights did not diverge
332
       \% If the weights are diverging, just set them to a large number
333
       if isnan (WcNew)
334
            fprintf('OOPS...DIVERGING WEIGHTS...I''M GOING TO USE LARGE WEIGHTS\
335
      n');
            weights = 1000 * \text{ones}(\text{length}(\text{rho}(e_{\text{vec}}(:,1))), 1);
           break;
       end;
338
339
       % Check for convergence of the critic weights
340
       if norm(WcNew - WcLast) < EpsilonWcritic
341
            weights = WcNew;
342
            fprintf('GREAT...THE WEIGHTS CONVERGED\n');
343
            break;
344
       end
345
       % If the weights did not converge, repeat the loop
346
```

```
WcLast = WcNew;
347
348
       \% If we reached the last iteration of the loop, just use the last
349
       % weights found
350
       if (i = (outerLoopMax - 1))
351
            fprintf ('OOPS...YOU REACHED THE END OF THE OUTER LOOP...I''M GOING
352
      TO USE THE LAST VALUE\langle n' \rangle;
            weights = WcNew;
353
       end
354
       end
356 end
357
358 % CRITIC WEIGHT TUNING NEURAL NETWORK
  % This function is specific to the helicopter because of the number of
359
360 % weights and error model
<sup>361</sup> function weights = quanserAEROCriticTuning(A, B, e_vec, tau, R_Mat, Q_Mat, xd,
       wcInit)
       % REFERENCE DR. MIAH'S PAPER FOR EQUATION NUMBERS
362
363
       % System dimensions specific for our model
364
       [n, \tilde{}] = size(B);
365
366
       % ERROR MODEL OF THE HELICOPTER
367
       % fbar and gbar --- EQ 8
368
       fbar = @(e) A * e;
369
       gbar = -B;
       hbar = -A*xd;
371
372
       % DISCRETE-TIME ERROR MODEL FOR TIME TAU
       \% f and g — EQ 10
374
       f = @(e) fbar(e) *tau + e;
375
       g = gbar * tau;
       h = hbar * tau;
378
       % COST FUNCTION PARAMETERS
379
       \% State penalizing function in the continuous cost function
380
       Qbar = @(e) e' * Q_Mat * e;
381
       % Control penalizing matrix in the continuous cost function
382
       Rbar = R_Mat;
383
       % The discrete-time cost function will have terms:
384
       % Right after EQ 11 in paper
       % State penalizing function in the discretized cost function
386
       Q = @(e) Qbar(e) *tau;
387
       % Control penalizing matrix in the discretized cost function
388
       R = Rbar*tau:
389
390
       % NEURAL NETWORK FUNCTIONS
391
       % Critic neural network activation functions
392
       rho = @(e) [e(1); e(2); e(3); e(4);...
393
                     e(1)^{2}; e(1) * e(2); e(1) * e(3); e(1) * e(4); \dots
394
                     e(2)^{2}; e(2) * e(3); e(2) * e(4); e(3)^{2}; e(3) * e(4); e(4)^{2};
395
       % Partial derivative of rho with respect to e
396
       drhode = @(e) [1, 0, 0; 0;
397
398
                       0, 1, 0, 0;
```

```
0, 0, 1, 0;
399
                       0, 0, 0, 1;
400
                       2 * e(1), 0, 0, 0;
401
                       e(2), e(1), 0, 0;
402
                       e(3), 0, e(1), 0;
403
                       e(4), 0, 0, e(1);
404
                       0, 2*e(2), 0, 0;
405
                       0\,,\ e\,(3)\,,\ e\,(2)\,,\ 0\,;
406
                       0, e(4), 0, e(2);
407
                       0, 0, 2*e(3), 0;
408
                       0, 0, e(4), e(3);
409
                       0, 0, 0, 2 * e(4);
410
411
       % TOLERANCES
412
       % Convergence tolerance for control policy
413
       EpsilonPolicy = 0.1;
414
       % Convergence tolerance for critic neural network
415
       EpsilonWcritic = 0.1;
416
417
       % TRAINING PARAMETERS
418
       % Number of outer loop iterations
419
       outerLoopMax = 700;
420
       % Number of inner loop iterations
421
       innerLoopMax = 100;
422
       \% Number of equations needed for training, number of sub-intervals
423
       % Number of training samples
       [, nbar] = size(e_vec);
425
426
       % WEIGHT INITIALIZATION
427
       \% Initialize the weights of the critic neural network to zero
428
       WcLast = zeros(length(rho(e_vec(:,1))), 1);
429
430
       % LEAST-SQUARES COMPUTATION INITIALIZATION
       \% Matrices required for computing least squares weights of the critic
432
       % neural networks --- EQ 20
433
       V = zeros(nbar, 1);
434
       Lambda = zeros(nbar, length(rho(e_vec(:,1))));
435
436
       % Matrix to hold the derivative of the error model during policy
437
       % updating
438
       e_k_plus_1 = zeros(n, nbar);
439
440
       % Product of the least squares matrices must be invertible
441
       \% Logic flag indicating if the critic weights are unsolvable
442
       \% The weights are unsolvable because the least squares matrices have no
443
       % solution --- not invertible
444
       diverged = 0;
445
446
       % OUTER LOOP
447
       for i = 1:(outerLoopMax - 1)
448
       \% Determine if the least squares matrices are invertible
449
       if diverged = 0
450
           % For each of the data collection (discrete time index)
451
           for k = 1:nbar
452
```

```
% Initialize the optimal inputs to zero
453
                uNew = [0; 0];
454
               % INNER LOOP
455
                for j = 1:(innerLoopMax - 1)
456
                    % Get the updated input value
457
                    uLast = uNew;
458
                    % Update the error model
                    e_k_plus_1(:,k) = f(e_vec(:,k)) + g*uLast + h;
460
                    % Compute the new optimal inputs
461
                    uNew = -0.5*(R^{(-1)})*g'*drhode(e_k_plus_1(:,k))'*WcLast;
462
463
                    % Check convergence of the optimal inputs
464
                    if norm(uNew - uLast) < EpsilonPolicy
465
                         break;
466
                    end
467
468
                end
469
               % Update the values for the least-squares computation
470
                V(k,:) = Q(e_vec(:,k)) + uNew' * R*uNew + WcLast'* rho(e_k_plus_1)
471
       (:,k));
                Lambda(k,:) = rho(e_vec(:,k))';
472
           end
473
       end
474
475
       \% Verify the least square solution exists for the critic's weights
       \% If the error data is consistent or there is no error, this will not
       \% hold, so set the weights to what they were initially before the
478
       % simulation
479
       if det(Lambda'*Lambda) = 0
            fprintf ('AWESOME...YOU HAVE NO ERROR...I''M GOING TO USE THE
481
      ORIGINAL WEIGHTS\langle n' \rangle;
           weights = wcInit;
482
           break;
       end:
484
485
       \% Calulcate least squares solution of critic's weights --- EQ 20
486
       WcNew = (Lambda'*Lambda)^{(-1)}Lambda'*V;
487
       % Make sure the weights did not diverge
       \% If the weights diverged, set the weights to what they were initially
489
       % before the simulation
490
       if isnan (WcNew)
491
            fprintf('OOPS...DIVERGING WEIGHTS...I''M GOING TO USE THE ORIGINAL
492
      WEIGHTS\n');
            weights = wcInit;
493
           break;
494
       end;
495
496
       % Check for convergence of the critic weights
497
       if norm(WcNew - WcLast) < EpsilonWcritic
498
            fprintf('GREAT...THE WEIGHTS CONVERGED\n');
499
            weights = WcNew;
500
           break;
501
       end
502
       \% If the weights did not converge, do another iteration of the loop
503
```

```
WcLast = WcNew;
504
505
       \% If we reach the last iteration of the loop, just use the last weights
506
       if (i = (outerLoopMax - 1))
507
            fprintf ( 'OOPS...YOU REACHED THE END OF THE OUTER LOOP...I''M GOING
508
      TO USE THE LAST VALUE\langle n' \rangle;
            weights = WcNew;
509
       end
       end
512 end
_{514} function [A, B] = getAeroAB(choice)
515 % This function builds the A and B matrix for either configuration of the
      Quanser Aero
516 % choice 0 selects the helicopter mode
517 % choice 1 selects the halfquad
                                        mode
518
519 % Quanser Aero Parameters
_{520} % Moment of Inertia of helicopter body (kg-m^2)
_{521} L-body = 6.5*0.0254; % length of horizontal body (metal tube)
_{522} m_body = 0.094; % mass of horizontal body (metal tube)
523 J_body = m_body * L_body^2 / 12; % horizontal cylinder rotating about CM
524
525 % Moment of Inertia of yoke fork that rotates about yaw axis (kg-m<sup>2</sup>)
_{526} m_yoke = 0.526; % mass of entire yoke assembly (kg)
_{527} % h_yoke = 9*0.0254; % height of yoke assembly (m)
_{528} r_fork = 0.04/2; % radius of each fork (approximated as cylinder)
J_{yoke} = 0.5 * m_{yoke} * r_{fork}^{2};
530
531 % Moment of Inertia from motor + guard assembly about pivot (kg-m<sup>2</sup>)
_{532} m_prop = 0.43; % mass of dc motor + shield + propeller shield
_{533} % m_motor = 0.203; % mass of dc motor
_{534} r_prop = 6.25*0.0254; % distance from CM to center of pitch axis
535 J_prop = m_prop * r_prop ^2; % using parallel axis theorem
536
537 % Equivalent Moment of Inertia about Pitch and Yaw Axis (kg-m^2)
J_{338} Jp = J_body + 2*J_prop; % pitch: body and 2 props
J_{39} Jy = J_body + 2*J_prop + J_yoke; % yaw: body, 2 props, and yoke
540
541 % Thrust-torque constant (N-m/V) [found experimentally]
_{542} Kpp = 0.0011; % (pre-production unit: 0.0015)
543 Kyy = 0.0022; % (pre-production unit: 0.0040)
544 Kpy = 0.0021; % thrust acting on pitch from yaw (pre-production unit:
      0.0020)
545 Kyp = -0.0027; % thrust acting on yaw from pitch (pre-production unit:
       -0.0017)
546
547 % Stiffness (N-m/rad) [found experimentally]
_{548} Ksp = 0.037463;
549
550 % Viscous damping (N-m-s/rad) [found experimentally]
_{551} Dp = 0.0071116; % pitch axis (pre-production unit: Dp = 0.0226)
_{552} Dy = 0.0220; % yaw axis (pre-production unit: Dy = 0.0211)
```

```
if choice = 0
554
             disp ('helicopter mode');
             A = \begin{bmatrix} 0 & 0 & 1 & 0 \end{bmatrix};
556
                       0 \ 0 \ 0 \ 1;
557
                    -\text{Ksp}/\text{Jp} 0 -\text{Dp}/\text{Jp} 0;
558
                       0 \ 0 \ 0 \ -Dy/Jy];
             \mathbf{B} = \begin{bmatrix} 0 & 0 \end{bmatrix};
560
                       0 \ 0;
561
                       Kpp/Jp Kpy/Jp;
562
                       Kyp/Jy Kyy/Jy];
563
564
             disp('A - Matrix');
565
             \operatorname{disp}\left(\mathbf{A}\right);
566
             \operatorname{disp}('B - \operatorname{Matrix}');
567
             \operatorname{disp}(B);
568
     else
569
             disp ('halfquad mode');
570
             A = \begin{bmatrix} 0 & 0 & 1 & 0 \end{bmatrix};
571
                       0 \ 0 \ 0 \ 1;
572
                    -\text{Ksp}/\text{Jp} \ 0 \ -\text{Dp}/\text{Jp} \ 0;
573
                       0 \ 0 \ 0 \ -Dy/Jy];
574
             B = \begin{bmatrix} 0 & 0 \end{bmatrix};
575
                       0 \ 0;
576
                      -Kpp/Jy Kpp/Jy;
577
                      -Kyy/Jy -Kyy/Jy];
578
             \operatorname{disp}(A - \operatorname{Matrix});
579
             \operatorname{disp}(A);
580
             disp('B - Matrix');
581
             \operatorname{disp}(B);
582
583 end
584
585 end
```

Appendix B

Running V-REP

B.1 Running a Simulation

B.1.1 File Setup

Setup the API Files

First, unpack the vrepAeroADP file to a spot on the computer. To use the following three MATLAB files, the version of them must be selected to reflect the 32/64 bit-architecture of the host system.

- remApi.m
- remoteApiProto.m
- remoteApi.dll (optional MATLAB will compile)

Make sure your MATLAB uses the same bit-architecture as the remoteApi library: 64bit MATLAB with 32bit remoteApi library will not work, and vice-versa! The above files are located in V-REP's installation directory at:

../programming/remoteApiBindings/matlab

The Vortex Physics Engine

The Vortex Physics Engine is a unified real-time simulation platform that V-REP uses to calculate the dynamics of the scene. To obtain Vortex, an account at CM-Labs.com must be made, and a free license acquired.

- 1. CM Labs Account and Vortex License
- 2. Vortex Download

B.1.2 Running the Simulation

To run a V-REP scene, controlled by MATLAB:

- 1. VREP: Select the Vortex physics engine.
- 2. VREP: Select dt = 50 ms.
- 3. VREP: Click the clock icon, to the right of the play button. Make sure the output console reads: Toggled to non real-time simulation mode.
- 4. V-REP: Press the Play button in the V-REP scene.
- 5. MATLAB: Press the Play Button/Run icon under the **Editor** tab.

B.2 Skeletal Framework

B.2.1 MATLAB Framework

```
2 % 2-DOF QUANSER HELICOPTER PARAMETERS
3 % Maximum applied voltage for the rotor motors
4 \text{ maxVolt} = 18;
5 % Number of state variables - theta, psi, dtheta, dpsi
6 n = 4;
                 % Pitch
_{7} \text{ theta} = 1;
s psi = 2;
                % Yaw
9 thetaDot = 3; \% Pitch angular velocity
10 psiDot = 4; % Yaw angular velocity
11 % Number of inputs – Vp, Vy
12 m = 2;
13
14 % SIMULATION TIMING
15 % Initial and final simulation times [s]
16 t0 = 0; tf = 30;
17 % Sampling time [s]
18 \text{ tau} = 0.05;
19 % Number of time steps
_{20} tsteps = floor ((tf-t0)/tau);
21 % Discrete time vecotor of sampling time (tau)
dt = tau * (0: tsteps);
23
state = zeros(n, tsteps+1);
_{25} u = zeros (m, tsteps+1);
26
27 %% connection routine
28
29 ip = '127.0.0.1';
_{30} port = 19999;
31
```

```
32 % create a handle to remote API
33 vrep = remApi('remoteApi'); % using the prototype file (remoteApiProto.m)
_{34} vrep.simxFinish(-1); % just in case, close all opened connections
35 clientID = vrep.simxStart(ip, port, true, true, 5000, 5);
36
37 %% simulation
  if (clientID > -1) % if clientID exists
38
       disp('Connected to remote API server');
39
40
      %Set up the handles
41
       [rtn, aero] = vrep.simxGetObjectHandle(clientID,...
42
                                                   'Aero',...
43
                                                    vrep.simx_opmode_blocking)
44
       [rtn, yJ] = vrep.simxGetObjectHandle(clientID,...
45
                                               yawJoint',...
46
                                               vrep.simx_opmode_blocking)
47
       [rtn, pJ] = vrep.simxGetObjectHandle(clientID,...
48
                                               pitchJoint',...
49
                                              vrep.simx_opmode_blocking)
50
51
      % setup joint position streaming
      vrep.simxGetJointPosition(clientID,...
53
                                     уЈ , . . .
54
                                     vrep.simx_opmode_streaming);
      vrep.simxGetJointPosition(clientID,...
56
                                     pJ , . . .
                                     vrep.simx_opmode_streaming);
58
59
      % add a delay to let the streaming operations initialize
60
      pause(1);
61
62
       [ret, state(1,1)] = vrep.simxGetJointPosition(clientID,...
63
                                                        pJ , . . .
64
                                                        vrep.simx_opmode_buffer);
65
66
       [ret, state(2,1)] = vrep.simxGetJointPosition(clientID,...
67
68
                                                        yJ , . . .
                                                        vrep.simx_opmode_buffer);
69
70
71
       state(3,1) = 0;
72
       state(4, 1) = 0;
73
74
      % start a timer
      tic:
76
      % RUN THE SIMULATION
77
      % Start at the next time tau after time zero
78
       for k = 2:tsteps+1
79
           % Current time
80
           t = (k-1)*tau; % generate value discrete time index
81
82
          \% Step 1 – get/store the state of the quadcopter from VREP
83
           [ret, state(1,k)] = vrep.simxGetJointPosition(clientID,...
84
                                                             pJ , . . .
85
```

```
vrep.simx_opmode_buffer
86
       );
87
             [ret, state(2,k)] = vrep.simxGetJointPosition(clientID,...
88
                                                                 yJ , . . .
89
                                                                 vrep.simx_opmode_buffer
90
       );
91
92
93
       % algorithm goes here
94
95
96
            u(1,k) = 5; \% motor 0
97
            u(2,k) = 5.01; \% \text{ motor } 1
98
99
            \% Step 3 – Update Inputs
100
            V0V1 = [u(1,k) u(2,k)];
            packedData=vrep.simxPackFloats(V0V1);
102
            vrep.simxSetStringSignal(clientID,...
103
                                         'Vin_matlab',...
104
                                         packedData , . . .
                                         vrep.simx_opmode_oneshot);
106
107
           pause(tau);
108
           t = t + tau;
110
       end
113
        toc; % end the timer
114
115
        ret = 1;
116
        while (ret \tilde{} = 0)
117
            [ret]=vrep.simxClearStringSignal(clientID,...
118
                                                 'Vin_matlab',...
119
                                                 vrep.simx_opmode_blocking);
120
       end
   else
123
        disp('Failed connecting to remote API server');
124
  end % if (clientID >-1)
125
126
127 disp('Sim ended');
```

B.2.2 LUA Framework

```
function sysCall_init()
    --- port to talk to MATLAB
    simRemoteApi.start(19999)
    --- Assign handles to Propeller Scripts
    propellerScripts={-1,-1}
    propellerScripts[1] = sim.getScriptHandle('propeller')
```

```
propellerScripts[2] = sim.getScriptHandle('propeller#0')
8
9
  end
10
  function sysCall_actuation()
       -- Receive velocities from server
       local packedData = sim.getStringSignal('Vin_matlab')
14
       if packedData then
16
            - Clear the signal
           sim.clearStringSignal('Vin_matlab')
18
           -- unpack 2 floats, save in array "a={V0, V1}"
19
           local a = sim.unpackFloatTable(packedData, 0, 2, 0)
20
           for i = 1, 2, 1 do
               -- set script variable Vin associated
               -- with the propellers
24
         sim.setScriptSimulationParameter(propellerScripts[i], 'Vin', a[i])
           end
26
      end
27
  end
28
  function sysCall_sensing()
30
31
  end
33
  function sysCall_cleanup()
34
35
36 end
```

B.3 Calculate State-Space

```
1 function [A, B] = getAeroAB( choice )
2 %% this function builds the A and B matrix for either configuration of the
     Quanser Aero
_3 % choice 0 selects the helicopter mode
4 % choice 1 selects the halfquad
                                     mode
5
      %% Quanser Aero Parameters
6
      % Moment of Inertia of helicopter body (kg-m^2)
      L_{body} = 6.5 * 0.0254; % length of horizontal body (metal tube)
8
      m_body = 0.094; % mass of horizontal body (metal tube)
9
      J_body = m_body * L_body^2 / 12; \% horizontal cylinder rotating about CM
11
      \% Moment of Inertia of yoke fork that rotates about yaw axis (kg-m^2)
      m_yoke = 0.526; % mass of entire yoke assembly (kg)
      \% h_yoke = 9*0.0254; % height of yoke assembly (m)
14
      r_{fork} = 0.04/2; % radius of each fork (approximated as cylinder)
      J_yoke = 0.5 * m_yoke * r_fork^2;
      % Moment of Inertia from motor + guard assembly about pivot (kg-m^2)
18
      m_{prop} = 0.43; % mass of dc motor + shield + propeller shield
      \% m_motor = 0.203; \% mass of dc motor
```

```
r_{prop} = 6.25*0.0254; % distance from CM to center of pitch axis
21
        J_prop = m_prop * r_prop^2; % using parallel axis theorem
2.2
23
        % Equivalent Moment of Inertia about Pitch and Yaw Axis (kg-m^2)
24
        Jp = J_body + 2*J_prop; % pitch: body and 2 props
25
        Jy = J_body + 2*J_prop + J_yoke; \% yaw: body, 2 props, and yoke
26
        % Thrust-torque constant (N-m/V) [found experimentally]
28
        Kpp = 0.0011; \% (pre-production unit: 0.0015)
29
        Kyy = 0.0022; \% (pre-production unit: 0.0040)
30
        Kpy = 0.0021; % thrust acting on pitch from yaw (pre-production unit:
       0.0020)
        Kyp = -0.0027; % thrust acting on yaw from pitch (pre-production unit:
       -0.0017)
       % Stiffness (N-m/rad) [found experimentally]
34
        Ksp = 0.037463;
35
36
        % Viscous damping (N-m-s/rad) [found experimentally]
37
        Dp = 0.0226; % pitch axis (pre-production unit: Dp = 0.0226)
38
        Dy = 0.0220; % yaw axis (pre-production unit: Dy = 0.0211)
39
40
        if choice = 0
41
             disp ('helicopter mode');
42
             A = \begin{bmatrix} 0 & 0 & 1 & 0 \end{bmatrix};
43
                    0 \ 0 \ 0 \ 1;
44
                  -\text{Ksp}/\text{Jp} \ 0 \ -\text{Dp}/\text{Jp} \ 0;
45
                    0 \ 0 \ 0 \ -Dy/Jy];
46
             B = \begin{bmatrix} 0 & 0 \end{bmatrix}
47
                    0 \ 0;
48
                    Kpp/Jp Kpy/Jp;
49
                    Kyp/Jy Kyy/Jy];
50
             \operatorname{disp}(A - \operatorname{Matrix});
             \operatorname{disp}(A);
             disp('B - Matrix');
54
             \operatorname{disp}(B);
        else
56
             disp ('halfquad mode');
             A = \begin{bmatrix} 0 & 0 & 1 & 0 \end{bmatrix};
58
                    0 \ 0 \ 0 \ 1;
59
                  -\text{Ksp}/\text{Jp} 0 -\text{Dp}/\text{Jp} 0;
60
                    0 \ 0 \ 0 \ -Dy/Jy];
61
             B = \begin{bmatrix} 0 & 0 \end{bmatrix}
62
                    0 0:
63
                    -Kpp/Jy Kpp/Jy;
64
                    -Kyy/Jy -Kyy/Jy];
65
             \operatorname{disp}(A - \operatorname{Matrix});
66
             \operatorname{disp}(A);
67
             disp('B - Matrix');
68
             \operatorname{disp}(B);
69
        end
70
71
72 end
```

```
1 function [A, B] = getVREPAeroAB( choice )
2 % this function builds the A and B matrix for either configuration of the
      Quanser Aero
_3 % choice 0 selects the helicopter mode
4 % choice 1 selects the halfquad
                                         mode
5
      %% VREP parameters
6
      % Moment of Inertia of pitcharm
7
8 %
         M_{fuselage} = 3.375 e^{-1};
9 %
         M_{\text{sideblock}} = 4.275 e - 2;
       M_{\text{-}}fuselage = 0.85;
       M_{\text{-sideblock}} = 0.4;
11
       M_motor
                  = 0.225;
13
       x_fuselage = 3.0000e - 2;
14
       y_fuselage = 3.75000e - 1;
       z_{fuselage} = 3.0000e - 2;
16
       x_{sideblock} = 4.7500 e^{-2};
18
       y_{-sideblock} = 3.0000e - 2;
19
       z_{sideblock} = 3.0000e - 2;
20
       r_{sideblock} = 3.8750e - 2;
21
22
       r_{-motor} = 1.5875 e^{-1};
23
24
       % MoI of yoke
       side_m = 0.18;
26
       cyl_{-}mass = 0.05;
       cross_m = .1;
28
       x_side = 5.0000e - 2;
30
       y_side = 5.0000e - 2;
31
       z_side = 2.0000e - 1;
       r_side
               = 8.7500 e - 2;
33
34
       x_{-}cross = 1.2500e-1;
35
       y_{-}cross = 5.0000e-2;
36
       z_{-}cross = 5.0000e-2;
37
38
       dia_{cyl} = 5.0000e-2;
39
       h_cyl
                 = 2.0000 e - 2;
40
41
       J_body = getrectPrismI(y_fuselage, z_fuselage, M_fuselage)
42
                + getrectPrismI(y_sideblock, z_sideblock, M_sideblock)
43
                + getrectPrismI(y_sideblock, z_sideblock, M_sideblock);
44
45
       \% Moment of Inertia of yoke fork that rotates about yaw axis (kg-m^2)
46
       J_support_on_axis = getrectPrismI(x_side, y_side, side_m);
47
       J_{support} = parAxisTheorem(J_{support_on_axis,side_m,r_side});
48
       J_cross = getrectPrismI(x_cross, y_cross, cross_m);
49
       J_{cyl} = 0.5 * h_{cyl} * (dia_{cyl}/2) .^{2};
50
       J_yoke = 2*J_support + J_cross + J_cyl;
52
```

```
% Moment of Inertia from motor + guard assembly about pivot (kg-m<sup>2</sup>)
54
        J_prop = M_motor * r_motor ^2; % using parallel axis theorem
56
        \% Equivalent Moment of Inertia about Pitch and Yaw Axis (kg-m^2)
        Jp = J_body + 2*J_prop \% pitch: body and 2 props
58
        Jy = J_body + 2*J_prop + J_yoke \% yaw: body, 2 props, and yoke
59
        % Thrust-torque constant (N-m/V) [found experimentally]
61
        Kpp = 0.03; \% (pre-production unit: 0.0015)
62
        Kyy = 0.0022; \% (pre-production unit: 0.0040)
63
        Kpy = 0.0021; % thrust acting on pitch from yaw (pre-production unit:
64
        0.0020)
        Kyp = -0.0027; % thrust acting on yaw from pitch (pre-production unit:
65
        -0.0017)
66
        % Stiffness (N-m/rad) [found experimentally]
67
         Ksp = 0.037463;
68
69
        % Viscous damping (N-m-s/rad) [found experimentally]
70
           Dp = 0.0071116; % pitch axis (pre-production unit: Dp = 0.0226)
71
           Dy = 0.0220; % yaw axis (pre-production unit: Dy = 0.0211)
73 %
           Ksp = 0;
74 %
           Dp = 0;
75 %
           Dy = 0;
76
        if choice == 0
             disp ('helicopter mode');
78
             A = \begin{bmatrix} 0 & 0 & 1 & 0 \end{bmatrix};
79
                    0 \ 0 \ 0 \ 1;
80
                   -\mathrm{Ksp}/\mathrm{Jp} 0 -\mathrm{Dp}/\mathrm{Jp} 0;
81
                    0 \ 0 \ 0 \ -Dy/Jy];
82
             B = \begin{bmatrix} 0 & 0 \end{bmatrix};
                    0 \ 0;
84
                    Kpp/Jp Kpy/Jp;
85
                    Kyp/Jy Kyy/Jy];
86
87
              disp('A - Matrix');
              \operatorname{disp}(A);
89
              disp('B - Matrix');
90
              \operatorname{disp}(B);
91
        else
92
             disp ('halfquad mode');
93
             A = \begin{bmatrix} 0 & 0 & 1 & 0 \end{bmatrix};
94
                    0 \ 0 \ 0 \ 1;
95
                   -\mathrm{Ksp}/\mathrm{Jp} 0 -\mathrm{Dp}/\mathrm{Jp} 0;
96
                    0 \ 0 \ 0 \ -Dy/Jy];
97
             B = \begin{bmatrix} 0 & 0 \end{bmatrix}
98
                    0 \ 0;
99
                    -Kpp/Jy Kpp/Jy;
                    -Kyy/Jy Kyy/Jy];
              disp('A - Matrix');
              \operatorname{disp}(A);
103
              disp('B - Matrix');
104
              \operatorname{disp}(B);
```

```
106
        end
107
108
   end
109
   function [I] = getrectPrismI(a, b, m)
110
        I = m*(a.^2 + b.^2)/12;
   end
112
113
   function [pax] = parAxisTheorem(I,m,r)
114
        pax = I + m * r * r;
115
116 end
```

B.4 Joint Control

B.4.1 LUA 2-DOF Helicopter

```
-- Initialization
```

```
if (sim_call_type=sim.syscb_init) then
2
      simRemoteApi.start(19999)
3
4
      -- Get Handles for Propeller Objects and simulation timestep
5
    pJ = sim.getObjectHandle('pitchJoint');
6
    yJ = sim.getObjectHandle('yawJoint');
7
      prop0 = sim.getObjectHandle('motor0');
8
      prop1 = sim.getObjectHandle('motor1');
9
      timestep = sim.getSimulationTimeStep()
      print('timestep is : ', timestep)
11
      -- Get parameter values to work with
      Jy = sim.getScriptSimulationParameter(sim.handle_self, 'Jy')
13
      print ('Jy is : ', Jy)
14
      Jp = sim.getScriptSimulationParameter(sim.handle_self, 'Jp')
      print('Jp is : ', Jp)
16
      Dp = sim.getScriptSimulationParameter(sim.handle_self, 'Dp')
      print('Dp is : ', Dp)
18
      Dy = sim.getScriptSimulationParameter(sim.handle_self, 'Dy')
      print('Dy is : ', Dy)
20
      Kpp = sim.getScriptSimulationParameter(sim.handle_self, 'Kpp')
      print('Kpp is : ', Kpp)
      Kpy = sim.getScriptSimulationParameter(sim.handle_self, 'Kpy')
23
      print('Kpy is : ', Kpy)
24
      Kyp = sim.getScriptSimulationParameter(sim.handle_self, 'Kyp')
      print('Kyp is : ', Kyp)
26
      Kyy = sim.getScriptSimulationParameter(sim.handle_self, 'Kyy')
27
      print ('Kyy is : ', Kyy)
28
    Ksp = sim.getScriptSimulationParameter(sim.handle_self, 'Ksp')
      print('Ksp is : ', Ksp)
30
31
      -- Store current measured angular velocity to calculate angular
      acceleration
    prev_pJ_pose = sim.getJointPosition(pJ)
33
    prev_yJ_pose = sim.getJointPosition(yJ)
34
```

```
35
      sim.setJointTargetVelocity(pJ, 0)
36
      sim.setJointTargetVelocity(yJ, 0)
      sim.setJointTargetVelocity(prop0,0)
38
      sim.setJointTargetVelocity(prop1,0)
39
      pitchTarVel = 0
40
      yawTarVel = 0
41
42
     graphHandle=sim.getObjectHandle("Graph")
43
44
      PitchDesired = sim.getIntegerSignal("PitchDesired")
45
      YawDesired = sim.getIntegerSignal("YawDesired")
46
      lastPAcc = 0
47
      lastYAcc = 0
48
49
      sim.setJointTargetVelocity(prop0,1*0.8)
50
      sim.setJointTargetVelocity(prop1,-1*0.8)
51
  end
53
54
   - Looping code
55
     (sim_call_type=sim.syscb_actuation) then
  i f
56
58
          local t = sim.getSimulationTime()
59
60
       -- Receive velocities from server
61
      local packedData=sim.getStringSignal('MATLAB_SIG')
62
      if packedData then
63
           sim.clearStringSignal('MATLAB_SIG') -- Clear the signal
64
           local V=sim.unpackFloatTable(packedData, 0, 4, 0)
65
66
      pJ_pose = sim_getJointPosition(pJ)
67
      yJ_{pose} = sim.getJointPosition(yJ)
68
           pJ_vel = (pJ_pose - prev_pJ_pose)/timestep
           yJ_vel = (yJ_pose - prev_yJ_pose)/timestep
71
           pAcc = -(Ksp/Jp)*math.sin(pJ_pose) + -(Dy/Jp)*pJ_vel + (Kpp/Jp)*V[1]
      + (Kpy/Jp)*V[2]
           yAcc = -(Dp/Jy) * yJ_vel + (Kyp/Jy) * V[1] + (Kyy/Jy) * V[2]
73
74
           pitchTarVel = pitchTarVel + ((pAcc+lastPAcc)/2)*timestep
76
           yawTarVel = yawTarVel + ((yAcc+lastYAcc)/2)*timestep
77
78
           sim.setJointTargetVelocity(pJ, pitchTarVel)
           sim.setJointTargetVelocity(yJ, yawTarVel)
80
81
           prev_pJ_pose = pJ_pose
82
           prev_yJ_pose = yJ_pose
83
           lastPAcc = pAcc
84
85
           lastYAcc = yAcc
```

```
sim.setGraphUserData(graphHandle, 'PitchDesired', V[3])
86
           sim.setGraphUserData(graphHandle, 'YawDesired', V[4])
87
          sim.setJointTargetVelocity(prop0,V[1]*0.8)
          sim.setJointTargetVelocity(prop1, -V[2]*0.8)
90
91
       end
92
93
  end
94
95
  function sysCall_sensing()
96

    put your sensing code here

97
  end
98
99
  if (sim_call_type=sim.syscb_cleanup) then
100
102 end
```

B.4.2 LUA 2-DOF Half-Quadcopter

```
1 --- Initialization
```

```
if (sim_call_type=sim.syscb_init) then
2
      simRemoteApi.start(19999)
3
4
    pJ = sim.getObjectHandle('pitchJoint');
5
    yJ = sim.getObjectHandle('yawJoint');
6
      prop0 = sim.getObjectHandle('motor0');
      prop1 = sim.getObjectHandle('motor1');
8
      timestep = sim.getSimulationTimeStep()
9
      print('timestep is : ', timestep)
        - Get parameter values to work with
      Jy = sim.getScriptSimulationParameter(sim.handle_self, 'Jy')
12
      print('Jy is : ', Jy)
13
      Jp = sim.getScriptSimulationParameter(sim.handle_self, 'Jp')
14
      print('Jp is : ', Jp)
      Dp = sim.getScriptSimulationParameter(sim.handle_self, 'Dp')
      print ('Dp is : ', Dp)
17
      Dy = sim.getScriptSimulationParameter(sim.handle_self, 'Dy')
18
      print('Dy is : ', Dy)
19
      Kpp = sim.getScriptSimulationParameter(sim.handle_self, 'Kpp')
20
      print('Kpp is : ', Kpp)
      Kyy = sim.getScriptSimulationParameter(sim.handle_self, 'Kyy')
      print('Kyy is : ', Kyy)
23
    Ksp = sim.getScriptSimulationParameter(sim.handle_self, 'Ksp')
24
      print('Ksp is : ', Ksp)
      -- Store current measured angular velocity to calculate angular
27
      acceleration
    prev_pJ_pose = sim_getJointPosition(pJ)
28
    prev_yJ_pose = sim.getJointPosition(yJ)
29
30
      sim.setJointTargetVelocity(pJ, 0)
31
      sim.setJointTargetVelocity(yJ, 0)
```

```
sim.setJointTargetVelocity(prop0,0)
33
       sim.setJointTargetVelocity(prop1,0)
34
       pitchTarVel = 0
35
      vawTarVel = 0
36
37
      graphHandle=sim.getObjectHandle("Graph")
38
39
       PitchDesired = sim.getIntegerSignal("PitchDesired")
40
      lastPAcc = 0
41
      lastYAcc = 0
43
       sim.setJointTargetVelocity(prop0,1*0.8)
44
       sim.setJointTargetVelocity(prop1,-1*0.8)
45
46
  end
47
48
     Looping code
49
```

```
if (sim_call_type=sim.syscb_actuation) then
50
         local t = sim.getSimulationTime()
       -- Receive velocities from server
53
      local packedData=sim.getStringSignal('MATLAB_SIG')
54
      if packedData then
           sim.clearStringSignal('MATLAB_SIG') -- Clear the signal
           local V=sim.unpackFloatTable(packedData,0,4,0)
58
          --- Measure pitch and yaw
      pJ_pose = sim_getJointPosition(pJ)
60
      yJ_pose = sim_getJointPosition(yJ)
61
62
            - Calculate Velocites
63
          pJ_vel = (pJ_pose - prev_pJ_pose) / timestep
64
           yJ_vel = (yJ_pose - prev_yJ_pose)/timestep
65
66
          pAcc = -(Ksp/Jy)*math.sin(pJ_pose) + -(Dy/Jy)*pJ_vel + -(Kpp/Jy)*V
67
      [1] + (Kpp/Jy) *V[2]
          yAcc = -(Dp/Jp) * yJ_vel + -(Kyy/Jy) *V[1] + -(Kyy/Jy) *V[2]
68
           pitchTarVel = pitchTarVel + ((pAcc+lastPAcc)/2)*timestep
71
          yawTarVel = yawTarVel + ((yAcc+lastYAcc)/2)*timestep
72
           sim.setJointTargetVelocity(pJ, pitchTarVel)
74
           sim.setJointTargetVelocity(yJ, yawTarVel)
75
76
           prev_pJ_pose = pJ_pose
           prev_yJ_pose = yJ_pose
78
          lastPAcc = pAcc
79
          lastYAcc = vAcc
80
          sim.setGraphUserData(graphHandle, 'PitchDesired', V[3])
81
          sim.setGraphUserData(graphHandle, 'YawDesired', V[4])
82
83
```

```
sim.setJointTargetVelocity(prop0,V[1]*0.8)
84
           sim.setJointTargetVelocity(prop1,-V[2]*0.8)
85
       end
87
88
89
  \operatorname{end}
90
  function sysCall_sensing()
91
  end
92
93
  if (sim_call_type=sim.syscb_cleanup) then
94
95
96 end
```

B.4.3 LQR Half-Quadcopter

```
1\% Use exact model of the 2-DOF helicopter and the linearized model to find
2 % error data points
<sup>3</sup>% Use the error data points directly in the neural network instead of
4 % randomizing the error
5 close all; clear; clc;
6 % Name of .mat file to save the data
7 testName = 'half_joint_sinesine';
  [A, B] = getAeroAB(1)
8
9
11 % 2–DOF QUANSER HELICOPTER PARAMETERS
12 % Maximum applied voltage for the rotor motors
_{13} maxVolt = 18;
14 % Number of state variables – theta, psi, dtheta, dpsi
15 n = 4;
_{16} theta = 1;
                 % Pitch
                 % Yaw
17 \text{ psi} = 2;
18 thetaDot = 3; \% Pitch angular velocity
19 psiDot = 4; % Yaw angular velocity
20 % Number of inputs – Vp, Vy
_{21} m = 2;
22
23 % SIMULATION TIMING
24 % Initial and final simulation times [s]
_{25} t0 = 0; tf = 30;
26 % Sampling time [s]
_{27} tau = 0.05;
28 % Larger sampling time for updating the inputs [s]
29 T = 0.2;
30 % Number of time steps
31 \text{ tsteps} = \text{floor}((\text{tf}-\text{t0})/\text{tau});
32 % Discrete time vecotor of sampling time (tau)
dt = tau * (0: tsteps);
34
35 % COST FUNCTION
_{36}\ \%\ Q and R matrices used in the cost function
_{37} % Use the matrices used in Quanser documentation except add ones for the
38 % velocities -> need Q to be positive definite
```

```
39 % Modified LQR cost matrices
40 Q_Mat_LQR = diag([250 \ 30 \ 1 \ 1]);
41 R_Mat_LQR = 0.005 * \text{diag}([1 \ 1]);
_{42} nonLinearAdjustment = 0.01;
43
44
45 % INITIAL STATES
_{46} \% \text{ xInit} = [\deg 2 \operatorname{rad}(0) \ \deg 2 \operatorname{rad}(0) \ (0 * \operatorname{pi}/180) * 10 \ (0 * \operatorname{pi}/180) * 7]';
_{47} %xInit = \begin{bmatrix} 0 & 0 & 0 & 0 \end{bmatrix}';
48
49 % MATRIX INITIALIZATION
50 % Actual states – state variable x
51  % xADP = zeros (n, tsteps+1);
_{52} xLQR = _{zeros}(n, tsteps+1);
_{53} pose = zeros (n, tsteps+1);
_{54} % xLQR(:,1) = xInit;
55 % State errors for entire simulation
56 \operatorname{errorLQR} = \operatorname{zeros}(n, \operatorname{tsteps}+1);
57 % Actual inputs
_{58} uLQR = zeros (m, tsteps+1);
59
60 % Desired states
61 % Uncomment for desired states of all zero
_{62} %xd = zeros(n, tsteps+1);
_{63} % Uncomment for desired states that switch constant values
_{64} %xd = [deg2rad(37)*ones(1,tsteps+1); deg2rad(80)*ones(1,tsteps+1); zeros(1,
      tsteps+1; zeros(1, tsteps+1);
_{65} %xd = [\deg 2rad(10) * ones(1,1000), \deg 2rad(60) * ones(1,tsteps+1-1000); \deg 2rad(60) * ones(1,tsteps+1)]
      (80) * ones (1,1500), deg2rad (-120) * ones (1,tsteps+1-1500); zeros (1,tsteps+1)
       ; \operatorname{zeros}(1, \operatorname{tsteps}+1)];
66 % Uncomment for desired states that are sine and cosine waves
   pitchSine = deg2rad(30) * sin(0.5 * dt);
67
   yawCosine = deg2rad(60) * \cos(0.5 * dt);
69 % pitchSquare = deg2rad(25) * square(0.5 * dt);
70 % yawSquare = deg2rad(-36)*square(0.25*dt);
71 \% xd = [pitchSine; yawCosine; zeros(1,tsteps+1); zeros(1,tsteps+1)];
_{72} % xd = [pitchSquare; yawSquare; zeros(1,tsteps+1); zeros(1,tsteps+1)];
73
74
75
76 %% CALCULATE THE CONTROLLER GAIN MATRIX USING LQR
77 fprintf('WORKING ON LQR GAIN MATRIX... \ n');
78 kLQR = lqr(A, B, Q_Mat_LQR, R_Mat_LQR);
79
  pause(1);
80
81 %% connection routine
82
ip = '127.0.0.1';
_{84} port = 19999;
85
86 % create a handle to remote API
87 vrep = remApi('remoteApi'); % using the prototype file (remoteApiProto.m)
ss vrep.simxFinish(-1); % just in case, close all opened connections
se clientID = vrep.simxStart(ip, port, true, true, 5000, 5);
```

```
90
91 %% simulation
   if (clientID >-1) \% if clientID exists
92
       disp('Connected to remote API server');
93
94
95 %Set up the handles
   [rtn, aero] = vrep.simxGetObjectHandle(clientID, ...
96
                                                'Aero ' ,...
97
                                                 vrep.simx_opmode_blocking)
98
   [rtn, yJ] = vrep.simxGetObjectHandle(clientID,...
99
                                            yawJoint',...
100
                                            vrep.simx_opmode_blocking)
   [rtn, pJ] = vrep.simxGetObjectHandle(clientID,...
                                            pitchJoint',...
103
                                           vrep.simx_opmode_blocking)
104
106 % setup joint position streaming
  vrep.simxGetJointPosition(clientID,...
107
                                  yJ , . . .
108
                                  vrep.simx_opmode_streaming);
109
  vrep.simxGetJointPosition(clientID,...
110
                                  pJ , . . .
                                  vrep.simx_opmode_streaming);
114 % add a delay to let the streaming operations initialize
  pause(1);
115
116
   [ret, pose(1,1)] = vrep.simxGetJointPosition(clientID,...
117
118
                                                         pJ . . . .
                                                         vrep.simx_opmode_buffer);
119
120
   [ret, pose(2,1)] = vrep.simxGetJointPosition(clientID,...
                                                         vJ , . . .
                                                         vrep.simx_opmode_buffer);
123
124
126 pose (3, 1) = 0;
127 pose (4, 1) = 0;
128 \text{ xLQR}(:, 1) = \text{pose}(:, 1);
129 % Initialize the error
130 errorLQR(:,1) = xd(:,1) - xLQR(:,1);
133 tic;
134 % RUN THE SIMULATION
135 % Start at the next time tau after time zero
   for k = 2:tsteps+1
136
       % Current time
       t = (k-1)*tau; % generate value discrete time index
139
       % Step 1 - get/store the state of the quadcopter from VREP
140
       [ret, pose(1,k)] = vrep.simxGetJointPosition(clientID, ...)
141
                                                             pJ , . . .
142
143
                                                             vrep.simx_opmode_buffer)
```

```
144
        [ret, pose(2,k)] = vrep.simxGetJointPosition(clientID,...
145
                                                               vJ , . . .
146
                                                                vrep.simx_opmode_buffer)
147
       ;
148
       pose(3,k) = (pose(1,k) - pose(1,k-1)) \setminus tau;
149
       pose(4,k) = (pose(2,k) - pose(2,k-1)) \setminus tau;
       xLQR(:,k) = pose(:,k);
153
154
       % EXACT 2-DOF MODEL
       % Take previous values to find the derivative of the exact model
       xdotNonLinearLQR = A*xLQR(:, k-1) + B*uLQR(:, k-1);
158
       % Update the states of the exact model
       \% Use Euler integration to estimate the current states of the nonlinear
160
       % model
161
        xLQR(:,k) = xLQR(:,k-1) + xdotNonLinearLQR*tau; \% + nonLinearAdjustment
162
       .*(2.*rand(4,1)-1);
163
164 %
           xLQR(theta, k) = angleLimiterPitch(xLQR(theta, k));
       % Force the yaw angle to be in the range [-pi,pi]
165
        xLQR(psi,k) = angleLimiterYaw(xLQR(psi,k));
167
       % FIND THE ERROR BETWEEN THE EXACT STATES AND THE DESIRED STATES
168
       \operatorname{errorLQR}(:, k) = \operatorname{xd}(:, k) - \operatorname{xLQR}(:, k);
169
       % Update the inputs using LQR
171
       uNewLQR = kLQR * errorLQR(:, k);
       % Limit the voltages
       uNewLQR(1) = sign(uNewLQR(1)) * min(abs(uNewLQR(1)), maxVolt);
174
       uNewLQR(2) = sign(uNewLQR(2)) * min(abs(uNewLQR(2)), maxVolt);
       % Update the input matrices
176
       uLQR(1,k) = uNewLQR(1);
       uLQR(2,k) = uNewLQR(2);
178
179
       \% Step 3 – Update Inputs
180
       VREP_PACKET = [uNewLQR(1) \ uNewLQR(2) \ rad2deg(xd(1,k)) \ rad2deg(xd(2,k))];
181
    %
        motorSpeeds = \begin{bmatrix} 20 & 0 \end{bmatrix};
182
       packedData=vrep.simxPackFloats(VREP_PACKET);
183
       vrep.simxSetStringSignal(clientID,...
184
                                    'MATLAB_SIG ' ....
185
                                   packedData , ...
186
                                    vrep.simx_opmode_oneshot);
187
188
      pause(tau);
189
190
      t = t + tau;
191
193
194 end
```

```
195 toc;
196
197 \%ret = 1;
198 while (ret \tilde{} = 0)
        [ret]=vrep.simxClearStringSignal(clientID,...
199
                                             'MATLAB_SIG', ...
200
                                             vrep.simx_opmode_blocking);
201
  \operatorname{end}
202
203
   else
204
        disp('Failed connecting to remote API server');
205
  end % if (clientID >-1)
206
207
   disp('Sim ended');
208
209
210 figure
plot (dt, rad2deg(pose(1, :))); hold on;
plot (dt, rad2deg(xd(1, :))); hold off;
213 title('pitch')
214
215 figure
plot (dt, rad2deg(pose(2, :))); hold on;
plot (dt, rad2deg(xd(2, :))); hold off;
218 title ('yaw')
219
220 figure
plot (dt, uLQR(1, :)); hold on;
222 \operatorname{plot}(\operatorname{dt},\operatorname{uLQR}(2,:)); hold off;
223 title('Inputs')
224
225 % SAVE OUTPUTS FOR PLOTTING
226 % Save the time, error, input, states and desired states vectors
227 save ([testName, '.mat'], 'dt', 'xd', 'tau', 'errorLQR', 'uLQR', 'xLQR', 'R_Mat_LQR'
       , 'Q_Mat_LQR ');
228 % Notify that the simulation is complete
229 disp ('SIMULATION COMPLETE');
230
231 % ANGLE LIMITER FOR YAW FUNCTION
_{232} function angle = angleLimiterYaw(angle)
   % This function limits the angle between -pi and pi
233
        angle = mod(angle, 2*pi);
234
235
        i=find (angle>pi);
236
        angle(i) = angle(i) - 2*pi;
237
238
        i=find(angle<-pi);
239
        angle(i) = angle(i) + 2*pi;
240
241 end
```

B.4.4 LQR 2-DOF Helicopter

```
    % Use exact model of the 2-DOF helicopter and the linearized model to find
    % error data points
    % Use the error data points directly in the neural network instead of
```

```
4 % randomizing the error
5 close all; clear; clc;
7 % Name of .mat file to save the data
s testName = 'heli_joint_sqsq';
9
10 % LINEAR MODEL MATRIX PARAMETERS (2DoF Helicopter)
[A, B] = getAeroAB(0);
13 % 2–DOF QUANSER HELICOPTER PARAMETERS
14 % Maximum applied voltage for the rotor motors
_{15} \max \text{Volt} = 18;
16 % Number of state variables - theta, psi, dtheta, dpsi
17 n = 4;
_{18} theta = 1;
                  % Pitch
                  % Yaw
19 psi = 2;
20 thetaDot = 3; \% Pitch angular velocity
21 psiDot = 4; % Yaw angular velocity
_{22} % Number of inputs - Vp, Vy
_{23} m = 2;
24
25 % SIMULATION TIMING
26 % Initial and final simulation times [s]
27 t0 = 0; tf = 30;
28 % Sampling time [s]
29 tau = 0.05;
30 % Larger sampling time for updating the inputs [s]
_{31} T = 0.2;
32 % Number of time steps
_{33} tsteps = floor ((tf-t0)/tau);
34 % Discrete time vecotor of sampling time (tau)
_{35} dt = tau * (0: tsteps);
36
37 % COST FUNCTION
38 % Q and R matrices used in the cost function
_{39} % Use the matrices used in Quanser documentation except add ones for the
40 % velocities -> need Q to be positive definite
41 % Modified LQR cost matrices
_{42} Q_Mat_LQR = diag([250 30 1 1]);
43 R_Mat_LQR = 0.005 * \text{diag}([1 \ 1]);
44 nonLinearAdjustment = 0.01;
45
46 % INITIAL STATES
47 \% \text{ xInit} = [\deg 2 \operatorname{rad}(0) \ \deg 2 \operatorname{rad}(0) \ (0 * \operatorname{pi}/180) * 10 \ (0 * \operatorname{pi}/180) * 7]';
_{48} %xInit = \begin{bmatrix} 0 & 0 & 0 \end{bmatrix}';
49
50 % MATRIX INITIALIZATION
51 % Actual states - state variable x
_{52} xLQR = zeros (n, tsteps+1);
_{53} pose = zeros (n, tsteps+1);
54
55 % State errors for entire simulation
56 errorLQR = zeros(n, tsteps+1);
57 % Actual inputs
```

```
_{58} uLQR = zeros (m, tsteps+1);
 59
 60 % Desired states
 61 % Uncomment for desired states of all zero
 _{62} %xd = zeros (n, tsteps+1);
 _{63} % Uncomment for desired states that switch constant values
 _{64} %xd = [deg2rad(37)*ones(1,tsteps+1); deg2rad(80)*ones(1,tsteps+1); zeros(1,
             tsteps+1; zeros(1, tsteps+1);
 _{65} %xd = [\deg 2rad(10) * ones(1,1000), \deg 2rad(60) * ones(1,tsteps+1-1000); \deg 2rad(60) * ones(1,tsteps+1-1000); deg 2rad(60) * ones(1,tsteps+1-1000) * ones(1,tsteps+1-1000) * ones(1,tsteps+1-1000) * ones(1,tsteps+1-100
             (80) * ones (1, 1500), deg2rad (-120) * ones (1, tsteps + 1 - 1500); zeros (1, tsteps + 1)
              ; \operatorname{zeros}(1, \operatorname{tsteps}+1)];
 66 % Uncomment for desired states that are sine and cosine waves
 _{67} % pitchSine = deg2rad (30) * sin (0.5 * dt);
 _{68} % yawCosine = deg2rad(60) * cos(0.5 * dt);
          pitchSquare = deg2rad(25) * square(0.5 * dt);
 69
          yawSquare = deg2rad(-36) * square(0.25 * dt);
 70
       %xd = [pitchSine; yawCosine; zeros(1, tsteps+1); zeros(1, tsteps+1)];
 71
        xd = [pitchSquare; yawSquare; zeros(1, tsteps+1); zeros(1, tsteps+1)];
 72
 73
 74
 75 % CALCULATE THE CONTROLLER GAIN MATRIX USING LQR
     fprintf ('WORKING ON LQR GAIN MATRIX... \langle n' \rangle;
 76
 77 kLQR = lqr(A, B, Q_Mat_LQR, R_Mat_LQR);
 78
 79 %% connection routine
 so ip = '127.0.0.1';
 _{81} port = 19999;
 82
 83 % create a handle to remote API
 84 vrep = remApi('remoteApi'); % using the prototype file (remoteApiProto.m)
 vrep.simxFinish(-1); % just in case, close all opened connections
 se clientID = vrep.simxStart(ip, port, true, true, 5000, 5);
 87
 88 % simulation
     if (clientID >-1) % if clientID exists
 89
               disp('Connected to remote API server');
 90
 91
 92 % Set up the handles
      [rtn, aero] = vrep.simxGetObjectHandle(clientID,...
 93
                                                                                               'Aero ',...
 94
                                                                                                vrep.simx_opmode_blocking)
 95
      [rtn, yJ] = vrep.simxGetObjectHandle(clientID,...
 96
 97
                                                                                       yawJoint',...
                                                                                        vrep.simx_opmode_blocking)
 98
      [rtn, pJ] = vrep.simxGetObjectHandle(clientID,...
 99
                                                                                       pitchJoint',...
100
                                                                                     vrep.simx_opmode_blocking)
101
103 % setup joint position streaming
104 vrep.simxGetJointPosition(clientID,...
                                                                   yJ , . . .
105
                                                                   vrep.simx_opmode_streaming);
106
vrep.simxGetJointPosition(clientID,...
108
                                                                   рЈ , . . .
```

```
vrep.simx_opmode_streaming);
111 % add a delay to let the streaming operations initialize
112 pause (1);
   [ret, pose(1,1)] = vrep.simxGetJointPosition(clientID,...
114
                                                         pJ , . . .
                                                         vrep.simx_opmode_buffer);
117
    ret, pose(2,1)] = vrep.simxGetJointPosition(clientID,...
118
119
                                                         yJ , . . .
                                                         vrep.simx_opmode_buffer);
120
122
123 pose (3, 1) = 0;
124 pose (4, 1) = 0;
_{125} \text{ xLQR}(:, 1) = \text{pose}(:, 1);
126 % Initialize the error
127 errorLQR(:,1) = xd(:,1) - xLQR(:,1);
128
130
  tic:
131 % RUN THE SIMULATION
132 % Start at the next time tau after time zero
   for k = 2:tsteps+1
133
       % Current time
134
       t = (k-1)*tau; % generate value discrete time index
135
136
       \% Step 1 - get/store the state of the quadcopter from VREP
       [ret, pose(1,k)] = vrep.simxGetJointPosition(clientID,...
                                                             pJ , . . .
139
                                                             vrep.simx_opmode_buffer)
140
       ;
141
       [ret, pose(2,k)] = vrep.simxGetJointPosition(clientID, ...)
142
                                                             yJ , . . .
143
                                                             vrep.simx_opmode_buffer)
144
       ;
145
       pose(3,k) = (pose(1,k) - pose(1,k-1)) \setminus tau;
146
       pose(4,k) = (pose(2,k) - pose(2,k-1)) \setminus tau;
147
148
149
       xLQR(:,k) = pose(:,k);
       % EXACT 2-DOF MODEL
153
       \% Take previous values to find the derivative of the exact model
154
       xdotNonLinearLQR = A*xLQR(:, k-1) + B*uLQR(:, k-1);
       % Update the states of the exact model
156
       % Use Euler integration to estimate the current states of the nonlinear
157
       \% model
158
        xLQR(:,k) = xLQR(:,k-1) + xdotNonLinearLQR*tau; \% + nonLinearAdjustment
       .*(2.*rand(4,1)-1);
```

160

```
%
           xLQR(theta, k) = angleLimiterPitch(xLQR(theta, k));
161
       % Force the yaw angle to be in the range [-pi,pi]
        xLQR(psi,k) = angleLimiterYaw(xLQR(psi,k));
163
164
        \% FIND THE ERROR BETWEEN THE EXACT STATES AND THE DESIRED STATES
165
        \operatorname{errorLQR}(:, k) = \operatorname{xd}(:, k) - \operatorname{xLQR}(:, k);
166
167
       % Update the inputs using LQR
168
        uNewLQR = kLQR * errorLQR(:, k);
        % Limit the voltages
        uNewLQR(1) = sign(uNewLQR(1)) * min(abs(uNewLQR(1)), maxVolt);
        uNewLQR(2) = sign(uNewLQR(2)) * min(abs(uNewLQR(2)), maxVolt);
172
        % Update the input matrices
        uLQR(1,k) = uNewLQR(1);
174
        uLQR(2,k) = uNewLQR(2);
176
       % Step 3 - Update Inputs
        Vin=[uNewLQR(1) \ uNewLQR(2) \ rad2deg(xd(1,k)) \ rad2deg(xd(2,k))];
178
    %
        motorSpeeds = \begin{bmatrix} 20 & 0 \end{bmatrix};
179
        packedData=vrep.simxPackFloats(Vin);
180
        vrep.simxSetStringSignal(clientID,...
181
                                     'MATLAB_SIG', ....
182
                                    packedData , ...
183
                                    vrep.simx_opmode_oneshot);
184
       pause(tau);
186
187
       t = t + tau;
188
189
190
191 end
   toc;
192
193
194 \,\% ret = 1;
   while (ret \tilde{} = 0)
195
        [ret]=vrep.simxClearStringSignal(clientID,...
196
                                             'MATLAB_SIG',...
197
                                             vrep.simx_opmode_blocking);
198
   end
199
200
   else
201
        disp('Failed connecting to remote API server');
202
   end % if (clientID >-1)
203
204
  disp('Sim ended');
205
206
207 figure
plot (dt, rad2deg(pose(1, :))); hold on;
plot (dt, rad2deg(xd(1, :))); hold off;
210 title('pitch')
211
212 figure
plot (dt, rad2deg(pose(2, :))); hold on;
```

```
plot (dt, rad2deg(xd(2, :))); hold off;
215 title('yaw')
216
217 figure
218 \operatorname{plot}(\operatorname{dt},\operatorname{uLQR}(1,:)); hold on;
plot (dt, uLQR(2, :)); hold off;
220 title ('Inputs')
221
222 % SAVE OUTPUTS FOR PLOTTING
223 % Save the time, error, input, states and desired states vectors
224 save ([testName, '.mat'], 'dt', 'xd', 'tau', 'errorLQR', 'uLQR', 'xLQR', 'R_Mat_LQR'
       , 'Q_Mat_LQR');
225 % Notify that the simulation is complete
226 disp ('SIMULATION COMPLETE');
227
228 % ANGLE LIMITER FOR YAW FUNCTION
_{229} function angle = angleLimiterYaw(angle)
230 % This function limits the angle between -pi and pi
        angle = mod(angle, 2*pi);
232
        i=find (angle>pi);
234
        angle(i) = angle(i) - 2*pi;
235
        i=find (angle <-pi);
236
        angle(i) = angle(i) + 2*pi;
237
238 end
```

B.4.5 ADP Half-Quadcopter

```
1 % Use exact model of the 2-DOF helicopter and the linearized model to find
2 % error data points
_3 % Use the error data points directly in the neural network instead of
4 % randomizing the error
5 close all; clear; clc;
6 testName = 'half_joint_adp_sinesine';
8 % LINEAR MODEL MATRIX PARAMETERS
9 [A, B] = getAeroAB(1)
10 % 2-DOF QUANSER HELICOPTER PARAMETERS
11 % Maximum applied voltage for the rotor motors
_{12} \max Volt = 18;
13 % Number of state variables – theta, psi, dtheta, dpsi
14 n = 4;
                 % Pitch
_{15} \text{ theta} = 1;
                % Yaw
16 psi = 2;
17 thetaDot = 3; \% Pitch angular velocity
18 psiDot = 4; % Yaw angular velocity
19 % Number of inputs – Vp, Vy
20 m = 2;
21
22 % SIMULATION TIMING
23 % Initial and final simulation times [s]
t_{24} t_{0} = 0; t_{f} = 25;
25 % Sampling time [s]
```

```
_{26} tau = 0.05;
27 % Larger sampling time for updating the inputs [s]
_{28} T = 0.75;
29 % Number of time steps
30 \text{ tsteps} = \text{floor}((\text{tf}-\text{t0})/\text{tau});
31 % Discrete time vecotor of sampling time (tau)
dt = tau * (0: tsteps);
33 % Number of equations for actor-critic neural network
34 % Number of training samples per T
_{35} nbar = floor (T/tau);
36
37 % COST FUNCTION
_{38}~\%~Q and R matrices used in the cost function
39 %Q_Mat_ADP = diag([270 \ 100 \ 1 \ 1]);
40 %R_Mat_ADP = 0.0001 * diag([1 1]);
41 % Use the matrices used in Quanser documentation except add ones for the
42 % velocities -> need Q to be positive definite
<sup>43</sup> Q_Mat_ADP = diag([200 \ 75 \ 1 \ 1]);
44 R_Mat_ADP = 0.005 * \text{diag}([1 \ 1]);
_{45} % Modified LQR cost matrices
46 Q_Mat_LQR = diag([200 \ 75 \ 1 \ 1]);
47 R_Mat_LQR = 0.005 * \text{diag}([1 \ 1]);
48 nonLinearAdjustment = 0.03;
49 % Name of .mat file to save the data
50
54 % MATRIX INITIALIZATION
55 % Actual states – state variable x
_{56} xADP = zeros (n, tsteps+1);
57 % State errors for entire simulation
serrorADP = zeros(n, tsteps+1);
59 % Actual inputs
60 uADP = zeros(m, tsteps+1);
61
62 % Desired states
63 % Uncomment for desired states of all zero
_{64} %xd = zeros(n, tsteps+1);
_{65} % Uncomment for desired states that switch constant values
66 \ \% xd = [deg2rad(37) * ones(1, tsteps+1); deg2rad(80) * ones(1, tsteps+1); zeros(1, tsteps+1); zeros(1, tsteps+1); deg2rad(80) * ones(1, tsteps+1); zeros(1, tst
                 tsteps+1; zeros(1, tsteps+1);
67 \%xd = [\deg 2rad(10) * ones(1,1000), \deg 2rad(60) * ones(1,tsteps+1-1000); \deg 2rad(60) * ones(1,tsteps+1-1000); deg 2rad(10) * ones(1,1000), deg 2rad(10) * ones(1,1000), deg 2rad(10) * ones(1,1000) * 
                   (80) * ones(1, 1500), deg2rad(-120) * ones(1, tsteps+1-1500); zeros(1, tsteps+1)
                   ; \operatorname{zeros}(1, \operatorname{tsteps}+1);
_{68} % Uncomment for desired states that are sine and cosine waves
69 pitchSine = deg2rad(30) * sin(0.5 * dt);
70 yawCosine = deg2rad (90) *\cos(0.5*dt);
71 % pitchSquare = deg2rad(57) * square(0.5 * dt);
_{72} % yawSquare = deg2rad(-36)*square(0.25*dt);
_{73} xd = [pitchSine; yawCosine; zeros(1,tsteps+1); zeros(1,tsteps+1)];
74 \%xd = [pitchSquare; yawSquare; zeros (1, tsteps+1); zeros (1, tsteps+1)];
75
76
```

77

```
78 % FIND THE CRITIC NEURAL NEIWORK WEIGHTS BEFORE APPLYING ANY INPUTS
79 % Calculate the critic weights
80 wcInit = quanserAEROCriticTuningInitial(A, B, ((2*pi).*rand(4,nbar)-pi),tau,
      R_Mat_ADP, Q_Mat_ADP, xd(:, 1));
81 % Use the weights to determine the P matrix
  P_Mat = [wcInit(5) wcInit(6) wcInit(7) wcInit(8);
82
            wcInit(6) wcInit(9) wcInit(10) wcInit(11);
83
            wcInit(7) wcInit(10) wcInit(12) wcInit(13);
84
            wcInit(8) wcInit(11) wcInit(13) wcInit(14)];
85
86 % Use P to determine the state-feedback gain
^{87} kADP = 0.5 * (R_Mat_ADP^--1) *B' * P_Mat;
88
89
90
91
92
  %% connection routine
93
94
_{95} ip = '127.0.0.1';
_{96} port = 19999;
97
98 % create a handle to remote API
99 vrep = remApi('remoteApi'); % using the prototype file (remoteApiProto.m)
100 vrep.simxFinish(-1); % just in case, close all opened connections
  clientID = vrep.simxStart(ip, port, true, true, 5000, 5);
103 %% simulation
  if (clientID >-1) % if clientID exists
104
       disp('Connected to remote API server');
105
106
  %Set up the handles
107
   [rtn, aero] = vrep.simxGetObjectHandle(clientID,...
108
                                               'Aero ',...
109
                                               vrep.simx_opmode_blocking)
   [rtn, yJ] = vrep.simxGetObjectHandle(clientID,...
                                           yawJoint',...
                                           vrep.simx_opmode_blocking)
   [rtn, pJ] = vrep.simxGetObjectHandle(clientID,...
114
                                           pitchJoint',...
                                          vrep.simx_opmode_blocking)
116
117
118 % setup joint position streaming
vrep.simxGetJointPosition(clientID,...
                                 yJ , . . .
120
                                 vrep.simx_opmode_streaming);
  vrep.simxGetJointPosition(clientID,...
122
123
                                 pJ , . . .
                                 vrep.simx_opmode_streaming);
124
126 % add a delay to let the streaming operations initialize
  pause(1);
127
128
129 [ret, pose(1,1)] = vrep.simxGetJointPosition(clientID,...
```

```
130
                                                        pJ , . . .
                                                        vrep.simx_opmode_buffer);
   [ret, pose(2,1)] = vrep.simxGetJointPosition(clientID,...
133
134
                                                        yJ , . . .
                                                        vrep.simx_opmode_buffer);
136
137
138 pose (3, 1) = 0;
139 pose (4, 1) = 0;
_{140} \text{ xADP}(:, 1) = \text{pose}(:, 1);
141 % Initialize the error
142 errorADP(:,1) = xd(:,1) - xADP(:,1);
143
144
145
146 % RUN THE SIMULATION
147 % Start at the next time tau after time zero
148
149
150 % RUN THE SIMULATION
  % Start at the next time tau after time zero
152
  tic:
   for k = 2:tsteps+1
       % Current time
154
       t = (k-1)*tau; % generate value discrete time index
156
       \% Step 1 - get/store the state of the quadcopter from VREP
       [ret, pose(1,k)] = vrep.simxGetJointPosition(clientID,...
158
                                                            pJ , . . .
                                                            vrep.simx_opmode_buffer)
160
      ;
161
       [ret, pose(2,k)] = vrep.simxGetJointPosition(clientID,...
162
                                                            yJ , . . .
163
                                                            vrep.simx_opmode_buffer)
164
      ;
165
       pose(3,k) = (pose(1,k) - pose(1,k-1))/tau;
166
       pose(4,k) = (pose(2,k) - pose(2,k-1))/tau;
167
168
169
       xADP(:,k) = pose(:,k);
170
       % EXACT 2-DOF HELICOPTER MODEL
       \% Take previous values to find the derivative of the exact model
       xdotNonLinearADP = A*xADP(:, k-1) + B*uADP(:, k-1);
       % Update the states of the exact model
174
       \% Use Euler integration to estimate the current states of the nonlinear
       % model
       xADP(:,k) = xADP(:,k-1) + xdotNonLinearADP*tau + nonLinearAdjustment
       .*(2.*rand(4,1)-1);
       % ADD SOME DISTURBANCE TO THE STATES
178
       % This forces the states to some other position
179
180
       % if ((t > 15) \&\& (t < 17))
```

```
% ADP(:,k) = [rad2deg(-45) 0 0 0]';
181
           %xLQR(:,k) = [rad2deg(-45) 0 0 0]';
182
       %end
183
       % Force the pitch angle to be in the range [-pi/2, pi/2] because of
184
       % physical constraints
185
       xADP(theta, k) = angleLimiterPitch(xADP(theta, k));
186
       \% Force the yaw angle to be in the range [-pi, pi] because the yaw is
       % free to do a complete circle
188
       xADP(psi,k) = angleLimiterYaw(xADP(psi,k));
189
190
       % FIND THE ERROR BETWEEN THE EXACT STATES AND THE DESIRED STATES
       \operatorname{errorADP}(:, k) = \operatorname{xd}(:, k) - \operatorname{xADP}(:, k);
192
193
       % UPDATE THE CRITIC WEIGHTS EVERY T TIME
194
       % Determine if the time is a multiple of T
       if (mod(t,T) == 0)
196
           % Update the critic weights
197
            wc = quanserAEROCriticTuning(A, B, errorADP(:, (k-nbar):k), tau,
198
      R_Mat_ADP, Q_Mat_ADP, xd(:,k), wcInit);
           % Use the weights to determine the P matrix
199
            P_{Mat} = [wc(5) wc(6) wc(7) wc(8);
200
                      wc(6) wc(9) wc(10) wc(11);
201
                      wc(7) wc(10) wc(12) wc(13);
202
                      wc(8) wc(11) wc(13) wc(14);
203
           % Use P to determine the state-feedback gain
204
           kADP = 0.5 * (R_Mat_ADP^--1) *B' * P_Mat;
205
       end
206
       % Update the inputs using ADP
207
       uNewADP = kADP * errorADP(:, k);
208
       % Limit the voltages
209
       uNewADP(1) = sign(uNewADP(1)) * min(abs(uNewADP(1)), maxVolt);
210
       uNewADP(2) = sign(uNewADP(2)) * min(abs(uNewADP(2)), maxVolt);
       % Update the input matrices
212
       uADP(1,k) = uNewADP(1);
213
       uADP(2,k) = uNewADP(2);
214
       Vin=[uNewADP(1) uNewADP(2) rad2deg(xd(1,k)) rad2deg(xd(2,k))];
216
       packedData=vrep.simxPackFloats(Vin);
       vrep.simxSetStringSignal(clientID,...
218
                                   'MATLAB_SIG',...
219
                                   packedData , ...
                                   vrep.simx_opmode_oneshot);
221
      pause(tau);
223
224
      t = t + tau;
225
226
227
228 end
229 toc;
230
_{231} % ret = 1;
while (ret \tilde{} = 0)
   [ret]=vrep.simxClearStringSignal(clientID,...
233
```

```
'MATLAB_SIG', ...
234
                                            vrep.simx_opmode_blocking);
   end
236
237
   else
238
        disp('Failed connecting to remote API server');
239
   end % if (clientID >-1)
240
241
   disp('Sim ended');
242
243
244
245 figure
plot (dt, rad2deg(pose(1, :))); hold on;
plot (dt, rad2deg(xd(1,:))); hold off;
   title ('pitch')
248
249
250 figure
<sup>251</sup> plot (dt, rad2deg(pose(2,:))); hold on;
<sup>252</sup> plot (dt, rad2deg (xd (2,:))); hold off;
253 title ('yaw')
254
255 figure
_{256} plot (dt, uADP(1,:)); hold on;
_{257} plot (dt, uADP(2,:)); hold off;
  title('Inputs')
258
259
260 % SAVE OUTPUTS FOR PLOTTING
261 % Save the time, error, input, states and desired states vectors
  save ([testName, '.mat'], 'dt', 'errorADP', 'uADP', 'xADP', 'xd', 'tau', 'R_Mat_ADP',
262
        'Q_Mat_ADP ');
263 % Notify that the simulation is complete
<sup>264</sup> disp ('SIMULATION COMPLETE');
265
266 % ANGLE LIMITER FOR YAW FUNCTION
_{267} function angle = angleLimiterYaw(angle)
  % This function limits the angle between -pi and pi
268
        angle = mod(angle, 2*pi);
269
        i=find (angle>pi);
271
        angle(i) = angle(i) - 2*pi;
272
273
274
        i=find (angle <- pi);
        angle(i) = angle(i) + 2*pi;
275
276 end
277
278 % ANGLE LIMITER FOR PITCH FUNCTION
_{279} function angle = angleLimiterPitch (angle)
280 % This function limits the physical constraint of the pitch measurement to
  % 90 degrees
281
        if (angle < 0)
282
            angle = \max(\text{angle}, -\text{pi}/2);
283
       end
284
285
286
        if (angle > 0)
```

```
angle = min(angle, pi/2);
287
        end
288
   end
289
290
291 % CRITIC WEIGHT TUNING NEURAL NEIWORK
292 % This function is specific to the helicopter because of the number of
293 % weights and error model
<sup>294</sup> function weights = quanserAEROCriticTuningInitial(A, B, e_vec, tau, R_Mat,
       Q_Mat, xd)
       % REFERENCE DR. MIAH'S PAPER FOR EQUATION NUMBERS
295
296
       % SYSTEM PARAMETERS SPECIFIC TO THE 2-DOF QUANSER AERO
297
298 %
          A = \begin{bmatrix} 0 & 0 & 1 & 0; & 0 & 0 & 0 & 1; & -1.7442 & 0 & -0.3307 & 0; & 0 & 0 & 0 & -0.9283 \end{bmatrix};
          B = \begin{bmatrix} 0 & 0; & 0 & 0; & 0.0512 & 0.0977; & -0.1139 & 0.0928 \end{bmatrix};
  %
299
       % System dimensions specific for our model
300
        [n, \tilde{}] = size(B);
301
302
       % ERROR MODEL OF THE HELICOPTER
303
        \% fbar and gbar — EQ 8
304
        fbar = @(e) A * e;
305
        gbar = -B;
306
        hbar = -A*xd;
307
308
        % DISCRETE-TIME ERROR MODEL FOR TIME TAU
309
        \% f and g — EQ 10
310
        f = @(e) fbar(e) *tau + e;
311
        g = gbar * tau;
312
        h = hbar*tau;
313
314
        % COST FUNCTION PARAMETERS
315
       % State penalizing function in the continuous cost function
316
        Qbar = @(e) e'*Q_Mat*e;
317
       % Control penalizing matrix in the continuous cost function
318
        Rbar = R_Mat:
319
       \% The discrete-time cost function will have terms:
320
        % Right after EQ 11 in paper
321
        % State penalizing function in the discretized cost function
322
        Q = @(e) \quad Qbar(e) *tau;
323
       % Control penalizing matrix in the discretized cost function
324
       R = Rbar*tau;
325
326
       % NEURAL NEIWORK FUNCTIONS
       % Critic neural network activation functions
328
        rho = @(e) [e(1); e(2); e(3); e(4);...
                      e(1)^{2}; e(1) * e(2); e(1) * e(3); e(1) * e(4); \dots
330
                      e(2)^{2}; e(2) * e(3); e(2) * e(4); e(3)^{2}; e(3) * e(4); e(4)^{2};
331
       % Partial derivative of rho with respect to e
332
        drhode = @(e) [1, 0, 0, 0;
                        0, 1, 0, 0;
334
                        0, 0, 1, 0;
335
                        0, 0, 0, 1;
336
                        2 * e(1), 0, 0, 0;
337
                        e(2), e(1), 0, 0;
338
339
                        e(3), 0, e(1), 0;
```
```
e(4), 0, 0, e(1);
340
                      0, 2*e(2), 0, 0;
341
                      0, e(3), e(2), 0;
342
                      0, e(4), 0, e(2);
343
                      0, 0, 2*e(3), 0;
344
                      0, 0, e(4), e(3);
345
                      0, 0, 0, 2 * e(4)];
346
       % TOLERANCES
348
       % Convergence tolerance for control policy
       EpsilonPolicy = 0.1;
350
       % Convergence tolerance for critic neural network
351
       EpsilonWcritic = 0.1;
352
       % TRAINING PARAMETERS
354
       % Number of outer loop iterations
355
       outerLoopMax = 700;
356
       % Number of inner loop iterations
357
       innerLoopMax = 100;
358
       \% Number of equations needed for training, number of sub-intervals
359
       % Number of training samples
360
       [~, nbar] = size(e_vec);
361
362
       % WEIGHT INITIALIZATION
363
       % Initialize the weights of the critic neural network to zero
364
       WcLast = zeros(length(rho(e_vec(:,1))), 1);
365
366
       % LEAST-SQUARES COMPUTATION INITIALIZATION
367
       \% Matrices required for computing least squares weights of the critic
368
       % neural networks --- EQ 20
369
       V = zeros(nbar, 1);
370
       Lambda = zeros(nbar, length(rho(e_vec(:,1))));
       \% Matrix to hold the derivative of the error model during policy
373
       % updating
374
       e_k_plus_1 = zeros(n, nbar);
375
       % Product of the least squares matrices must be invertible
377
       % Logic flag indicating if the critic weights are unsolvable
378
       \% The weights are unsolvable because the least squares matrices have no
379
       % solution -- not invertible
       diverged = 0;
381
382
       % OUTER LOOP
383
       for i = 1:(outerLoopMax - 1)
384
       % Determine if the least squares matrices are invertible
385
       if diverged = 0
386
           % For each of the data collection (discrete time index)
387
           for k = 1:nbar
388
               % Initialize the optimal inputs to zero
389
               uNew = [0; 0];
390
               % INNER LOOP
391
                for j = 1:(innerLoopMax - 1)
392
                    % Get the updated input value
393
```

```
uLast = uNew;
394
                     % Update the error model
395
                      e_k_plus_1(:,k) = f(e_vec(:,k)) + g*uLast + h;
396
                     % Compute the new optimal inputs
397
                     uNew = -0.5*(R^{(-1)})*g'*drhode(e_k_plus_1(:,k))'*WcLast;
398
300
                     % Check convergence of the optimal inputs
400
                     if norm(uNew - uLast) < EpsilonPolicy
401
                          break;
402
                     end
403
                 end
404
405
                % Update the values for the least-squares computation
406
                 V(k,:) = Q(e_vec(:,k)) + uNew' * R * uNew + WcLast' * rho(e_k_plus_1)
407
       (:,k));
                 Lambda(k,:) = rho(e_vec(:,k))';
408
            end
409
       end
410
411
       \% Verify the least square solution exists for the critic's weights
412
       \% If the error data is consistent or there is no error, this will not
413
       % hold, so set the weights to zero
414
       if det(Lambda'*Lambda) = 0
415
            fprintf ('AWESOME...YOU HAVE NO ERROR...I'''M GOING TO SET THE WEIGHTS
416
       TO ZEROn');
            weights = \operatorname{zeros}(\operatorname{length}(\operatorname{rho}(\operatorname{e_vec}(:,1))),1);
417
            break;
418
       end:
419
       \% Calculate least squares solution of critic's weights --- EQ 20
421
       WcNew = (Lambda' * Lambda)^{(-1)} * Lambda' * V;
422
       % Make sure the weights did not diverge
423
       \% If the weights are diverging, just set them to a large number
424
       if isnan (WcNew)
425
            fprintf('OOPS...DIVERGING WEIGHTS...I''M GOING TO USE LARGE WEIGHTS\
426
       n');
            weights = 1000 * \text{ones} (\text{length}(\text{rho}(e_{\text{vec}}(:,1))), 1);
427
            break;
428
       end;
429
430
       % Check for convergence of the critic weights
        if norm(WcNew - WcLast) < EpsilonWcritic
432
            weights = WcNew;
433
            fprintf('GREAT...THE WEIGHTS CONVERGED\n');
434
            break:
435
       end
436
       % If the weights did not converge, repeat the loop
437
       WcLast = WcNew;
438
439
       \% If we reached the last iteration of the loop, just use the last
440
       % weights found
441
        if (i = (outerLoopMax - 1))
442
            fprintf ( 'OOPS...YOU REACHED THE END OF THE OUTER LOOP...I''M GOING
443
      TO USE THE LAST VALUE\langle n' \rangle;
```

```
weights = WcNew;
444
       end
445
446
       end
447 end
448
449 % CRITIC WEIGHT TUNING NEURAL NETWORK
450 % This function is specific to the helicopter because of the number of
451 % weights and error model
452 function weights = quanserAEROCriticTuning(A, B, e_vec, tau, R_Mat, Q_Mat, xd,
       wcInit)
       % REFERENCE DR. MIAH'S PAPER FOR EQUATION NUMBERS
453
454
       % SYSTEM PARAMETERS SPECIFIC TO THE 2-DOF QUANSER AERO
455
456 %
         A = \begin{bmatrix} 0 & 0 & 1 & 0; & 0 & 0 & 0 & 1; & -1.7442 & 0 & -0.3307 & 0; & 0 & 0 & 0 & -0.9283 \end{bmatrix};
          B = \begin{bmatrix} 0 & 0; & 0 & 0; & 0.0512 & 0.0977; & -0.1139 & 0.0928 \end{bmatrix};
457
       % System dimensions specific for our model
458
        [n, \tilde{}] = size(B);
459
460
       % ERROR MODEL OF THE HELICOPTER
461
       % fbar and gbar — EQ 8
462
       fbar = @(e) A * e;
463
        gbar = -B;
464
465
       hbar = -A*xd;
466
       % DISCRETE-TIME ERROR MODEL FOR TIME TAU
467
       \% f and g — EQ 10
        f = @(e) fbar(e) *tau + e;
469
       g = gbar * tau;
470
       h = hbar*tau;
472
       % COST FUNCTION PARAMETERS
473
       % State penalizing function in the continuous cost function
474
       Qbar = @(e) e' * Q_Mat * e;
       % Control penalizing matrix in the continuous cost function
476
       Rbar = R_Mat;
477
       % The discrete-time cost function will have terms:
478
       % Right after EQ 11 in paper
479
       % State penalizing function in the discretized cost function
480
       Q = @(e) Qbar(e) *tau;
481
       % Control penalizing matrix in the discretized cost function
482
       R = Rbar*tau:
484
       % NEURAL NETWORK FUNCTIONS
485
       % Critic neural network activation functions
486
       rho = @(e) [e(1); e(2); e(3); e(4);...
487
                      e(1)^{2}; e(1) * e(2); e(1) * e(3); e(1) * e(4); \dots
488
                      e(2)^{2}; e(2) * e(3); e(2) * e(4); e(3)^{2}; e(3) * e(4); e(4)^{2};
489
       % Partial derivative of rho with respect to e
490
        drhode = @(e) [1, 0, 0; 0;
491
                        0, 1, 0, 0;
492
                        0, 0, 1, 0;
493
                        0, 0, 0, 1;
494
                        2 * e(1), 0, 0;
495
496
                        e(2), e(1), 0, 0;
```

```
e(3), 0, e(1), 0;
497
                      e(4), 0, 0, e(1);
498
                      0, 2*e(2), 0, 0;
499
                      0, e(3), e(2), 0;
500
                      0, e(4), 0, e(2);
501
                      0, 0, 2*e(3), 0;
                      0, 0, e(4), e(3);
503
                      0, 0, 0, 2 * e(4)];
504
505
       % TOLERANCES
506
       % Convergence tolerance for control policy
507
       EpsilonPolicy = 0.1;
508
       % Convergence tolerance for critic neural network
509
       EpsilonWcritic = 0.1;
       % TRAINING PARAMETERS
512
       % Number of outer loop iterations
513
       outerLoopMax = 100;
514
       % Number of inner loop iterations
       innerLoopMax = 100;
       \% Number of equations needed for training, number of sub-intervals
       % Number of training samples
518
       [, nbar] = size(e_vec);
519
520
       % WEIGHT INITIALIZATION
       \% Initialize the weights of the critic neural network to zero
       WcLast = zeros(length(rho(e_vec(:,1))), 1);
523
524
       % LEAST-SQUARES COMPUTATION INITIALIZATION
       \% Matrices required for computing least squares weights of the critic
526
       % neural networks --- EQ 20
527
       V = zeros(nbar, 1);
528
       Lambda = zeros(nbar, length(rho(e_vec(:,1))));
530
       \% Matrix to hold the derivative of the error model during policy
       % updating
       e_k_plus_1 = zeros(n, nbar);
534
       % Product of the least squares matrices must be invertible
       \% Logic flag indicating if the critic weights are unsolvable
536
       \% The weights are unsolvable because the least squares matrices have no
       % solution --- not invertible
538
       diverged = 0;
539
540
       % OUTER LOOP
541
       for i = 1:(outerLoopMax - 1)
       \% Determine if the least squares matrices are invertible
543
       if diverged = 0
544
           % For each of the data collection (discrete time index)
545
           for k = 1:nbar
546
               % Initialize the optimal inputs to zero
547
               uNew = [0; 0];
548
               % INNER LOOP
549
                for j = 1:(innerLoopMax - 1)
```

```
% Get the updated input value
                    uLast = uNew;
                    % Update the error model
553
                    e_{k_{p}lus_{1}}(:,k) = f(e_{vec}(:,k)) + g*uLast + h;
554
                    % Compute the new optimal inputs
                    uNew = -0.5*(R^{(-1)})*g'*drhode(e_k_plus_1(:,k))'*WcLast;
556
                    % Check convergence of the optimal inputs
558
                    if norm(uNew - uLast) < EpsilonPolicy
559
                         break;
560
                    end
561
                end
562
563
                % Update the values for the least-squares computation
564
                V(k,:) = Q(e_vec(:,k)) + uNew' * R*uNew + WcLast'* rho(e_k_plus_1)
565
       (:,k));
                Lambda(k,:) = rho(e_vec(:,k))';
566
           end
567
       end
568
569
       \% Verify the least square solution exists for the critic's weights
       \% If the error data is consistent or there is no error, this will not
       \% hold, so set the weights to what they were initially before the
       % simulation
573
       if det (Lambda' \ast Lambda) == 0
574
           %fprintf('AWESOME...YOU HAVE NO ERROR...I''M GOING TO USE THE
      ORIGINAL WEIGHTS\langle n' \rangle;
            weights = wcInit:
576
            break;
       end;
578
579
       \% Calulcate least squares solution of critic's weights --- EQ 20
580
       WcNew = (Lambda'*Lambda)^{(-1)}Lambda'*V;
581
       % Make sure the weights did not diverge
582
       \% If the weights diverged, set the weights to what they were initially
583
       % before the simulation
584
       if isnan (WcNew)
           %fprintf('OOPS...DIVERGING WEIGHTS...I''M GOING TO USE THE ORIGINAL
586
      WEIGHTS\n');
            weights = wcInit;
587
           break;
588
       end;
589
590
       % Check for convergence of the critic weights
       if norm(WcNew - WcLast) < EpsilonWcritic
           %fprintf('GREAT...THE WEIGHTS CONVERGED\n');
593
           weights = WcNew;
594
            break;
       end
596
       \% If the weights did not converge, do another iteration of the loop
597
       WcLast = WcNew;
598
599
       \% If we reach the last iteration of the loop, just use the last weights
600
       if (i = (outerLoopMax - 1))
601
```

```
602 %fprintf('OOPS...YOU REACHED THE END OF THE OUTER LOOP...I''M GOING
TO USE THE LAST VALUE\n');
603 weights = WcNew;
604 end
605 end
606 end
```

B.4.6 ADP 2-DOF Helicopter

```
1\% Use exact model of the 2-DOF helicopter and the linearized model to find
2 % error data points
_3 % Use the error data points directly in the neural network instead of
4 % randomizing the error
5 close all; clear; clc;
6 testName = 'heli_joint_adp_sinesine';
8 % LINEAR MODEL MATRIX PARAMETERS
9 [A, B] = getAeroAB(0)
10 % 2-DOF QUANSER HELICOPTER PARAMETERS
11 % Maximum applied voltage for the rotor motors
_{12} \text{ maxVolt} = 18;
13 % Number of state variables – theta, psi, dtheta, dpsi
_{14} n = 4;
_{15} \text{ theta} = 1;
                 % Pitch
16 psi = 2;
                  % Yaw
17 thetaDot = 3; % Pitch angular velocity
18 psiDot = 4; % Yaw angular velocity
19 % Number of inputs - Vp, Vy
20 m = 2;
21
22 % SIMULATION TIMING
23 % Initial and final simulation times [s]
_{24} t0 = 0; tf = 25;
25 % Sampling time [s]
_{26} tau = 0.05;
27 % Larger sampling time for updating the inputs [s]
_{28} T = 0.75;
29 % Number of time steps
30 \text{ tsteps} = \text{floor}((\text{tf}-\text{t0})/\text{tau});
31 % Discrete time vecotor of sampling time (tau)
_{32} dt = tau * (0: tsteps);
33 % Number of equations for actor-critic neural network
34 % Number of training samples per T
_{35} nbar = floor (T/tau);
37 % COST FUNCTION
38 % Q and R matrices used in the cost function
39 %Q_Mat_ADP = diag([270 \ 100 \ 1 \ 1]);
40 %R_Mat_ADP = 0.0001 * diag ([1 1]);
_{41} % Use the matrices used in Quanser documentation except add ones for the
_{42} % velocities -> need Q to be positive definite
<sup>43</sup> Q_Mat_ADP = diag([200 \ 75 \ 1 \ 1]);
44 R_Mat_ADP = 0.005 * \text{diag}([1 \ 1]);
45 % Modified LQR cost matrices
```

```
46 Q_Mat_LQR = diag([200 \ 75 \ 1 \ 1]);
47 R_Mat_LQR = 0.005 * \text{diag}([1 \ 1]);
48 nonLinearAdjustment = 0.03;
49 % Name of .mat file to save the data
50
52
53
54 % MATRIX INITIALIZATION
55 % Actual states – state variable x
_{56} xADP = zeros (n, tsteps+1);
57 % State errors for entire simulation
58 errorADP = zeros(n, tsteps+1);
59 % Actual inputs
60 uADP = zeros(m, tsteps+1);
61
62 % Desired states
63 % Uncomment for desired states of all zero
_{64} %xd = zeros(n, tsteps+1);
_{65} % Uncomment for desired states that switch constant values
66 \ \% xd = [deg2rad(37) * ones(1, tsteps+1); deg2rad(80) * ones(1, tsteps+1); zeros(1, tsteps+1); zeros(1, tsteps+1); deg2rad(80) * ones(1, tsteps+1); zeros(1, tst
                   tsteps+1; zeros(1, tsteps+1);
67 \% d = [deg2rad(10) * ones(1,1000), deg2rad(60) * ones(1,tsteps+1-1000); deg2rad(
                   (80) * ones (1,1500), deg2rad (-120) * ones (1, tsteps+1-1500); zeros (1, tsteps+1)
                   ; \operatorname{zeros}(1, \operatorname{tsteps}+1)];
68 % Uncomment for desired states that are sine and cosine waves
69 pitchSine = deg2rad(60) * sin(0.5 * dt);
yawCosine = deg2rad (90) * \cos(0.5 * dt);
71 % pitchSquare = deg2rad(57) * square(0.5 * dt);
_{72} % yawSquare = deg2rad(-36)*square(0.25*dt);
_{73} xd = [pitchSine; yawCosine; zeros(1,tsteps+1); zeros(1,tsteps+1)];
74 \%xd = [pitchSquare; yawSquare; zeros(1,tsteps+1); zeros(1,tsteps+1)];
76
77
78 % FIND THE CRITIC NEURAL NEIWORK WEIGHTS BEFORE APPLYING ANY INPUTS
79 % Calculate the critic weights
      wcInit = quanserAEROCriticTuningInitial(A, B, ((2*pi).*rand(4,nbar)-pi),tau,
80
                  R_Mat_ADP, Q_Mat_ADP, xd(:, 1));
81 % Use the weights to determine the P matrix
      P_Mat = [wcInit(5) wcInit(6) wcInit(7) wcInit(8);
82
                                      wcInit(6) wcInit(9) wcInit(10) wcInit(11);
83
                                      wcInit(7) wcInit(10) wcInit(12) wcInit(13);
84
                                      wcInit(8) wcInit(11) wcInit(13) wcInit(14)];
85
86 % Use P to determine the state-feedback gain
^{87} kADP = 0.5 * (R_Mat_ADP^--1) *B' * P_Mat;
88
89
90
91
92
93 %% connection routine
94
95 ip = '127.0.0.1';
```

```
96 port = 19999;
97
98 % create a handle to remote API
99 vrep = remApi('remoteApi'); % using the prototype file (remoteApiProto.m)
100 vrep.simxFinish(-1); % just in case, close all opened connections
clientID = vrep.simxStart(ip, port, true, true, 5000, 5);
103 %% simulation
104 if (clientID >-1) % if clientID exists
       disp('Connected to remote API server');
105
106
107 % Set up the handles
   [rtn, aero] = vrep.simxGetObjectHandle(clientID, ...
108
                                                'Aero ',...
109
                                                 vrep.simx_opmode_blocking)
   [rtn, yJ] = vrep.simxGetObjectHandle(clientID,...
111
                                            yawJoint',...
112
                                            vrep.simx_opmode_blocking)
113
   [rtn, pJ] = vrep.simxGetObjectHandle(clientID,...
114
                                            pitchJoint',...
115
                                           vrep.simx_opmode_blocking)
116
117
118 % setup joint position streaming
  vrep.simxGetJointPosition(clientID,...
119
                                  yJ , . . .
120
                                  vrep.simx_opmode_streaming);
   vrep.simxGetJointPosition(clientID,...
                                  pJ , . . .
123
                                  vrep.simx_opmode_streaming);
124
125
126 % add a delay to let the streaming operations initialize
  pause(1);
127
128
   [ret, pose(1,1)] = vrep.simxGetJointPosition(clientID,...
                                                         pJ , . . .
130
                                                         vrep.simx_opmode_buffer);
   [ret, pose(2,1)] = vrep.simxGetJointPosition(clientID,...
134
                                                        yJ , . . .
                                                         vrep.simx_opmode_buffer);
136
137
138 pose (3, 1) = 0;
139 pose (4, 1) = 0;
_{140} \text{ xADP}(:, 1) = \text{pose}(:, 1);
141 % Initialize the error
142 errorADP(:,1) = xd(:,1) - xADP(:,1);
143
144
145
146 % RUN THE SIMULATION
  % Start at the next time tau after time zero
147
148
149
```

```
150 % RUN THE SIMULATION
151 % Start at the next time tau after time zero
152 tic;
153 for k = 2:tsteps+1
      % Current time
154
       t = (k-1)*tau; % generate value discrete time index
156
       % Step 1 - get/store the state of the quadcopter from VREP
       [ret, pose(1,k)] = vrep.simxGetJointPosition(clientID,...
158
                                                            pJ , . . .
                                                            vrep.simx_opmode_buffer)
160
      ;
161
       [ret, pose(2,k)] = vrep.simxGetJointPosition(clientID,...
                                                            yJ , . . .
                                                            vrep.simx_opmode_buffer)
164
      ;
165
       pose(3,k) = (pose(1,k) - pose(1,k-1)) \setminus tau;
166
       pose(4,k) = (pose(2,k) - pose(2,k-1)) \setminus tau;
167
168
170
       xADP(:,k) = pose(:,k);
       % EXACT 2-DOF HELICOPTER MODEL
171
       \% Take previous values to find the derivative of the exact model
       xdotNonLinearADP = A*xADP(:, k-1) + B*uADP(:, k-1);
       % Update the states of the exact model
174
       \% Use Euler integration to estimate the current states of the nonlinear
       % model
176
       xADP(:,k) = xADP(:,k-1) + xdotNonLinearADP*tau + nonLinearAdjustment
       .*(2.*rand(4,1)-1);
       % ADD SOME DISTURBANCE TO THE STATES
178
       % This forces the states to some other position
       % if ((t > 15) \&\& (t < 17))
180
           % ADP(:,k) = [rad2deg(-45) 0 0 0]';
181
           %xLQR(:,k) = [rad2deg(-45) 0 0 0]';
182
       %end
183
       % Force the pitch angle to be in the range [-pi/2, pi/2] because of
184
       % physical constraints
185
       xADP(theta, k) = angleLimiterPitch(xADP(theta, k));
186
       \% Force the yaw angle to be in the range [-pi, pi] because the yaw is
       % free to do a complete circle
188
       xADP(psi,k) = angleLimiterYaw(xADP(psi,k));
189
190
       % FIND THE ERROR BETWEEN THE EXACT STATES AND THE DESIRED STATES
       \operatorname{errorADP}(:, k) = \operatorname{xd}(:, k) - \operatorname{xADP}(:, k);
192
193
       % UPDATE THE CRITIC WEIGHTS EVERY T TIME
194
       % Determine if the time is a multiple of T
       if (mod(t,T) == 0)
196
           % Update the critic weights
197
           wc = quanserAEROCriticTuning(A, B, errorADP(:,(k-nbar):k),tau,
198
      R_Mat_ADP, Q_Mat_ADP, xd(:,k), wcInit);
199
           % Use the weights to determine the P matrix
```

```
P_{Mat} = [wc(5) wc(6) wc(7) wc(8);
200
                      wc(6) wc(9) wc(10) wc(11);
201
                      wc(7) wc(10) wc(12) wc(13);
202
                      wc(8) wc(11) wc(13) wc(14);
203
            % Use P to determine the state-feedback gain
204
            kADP = 0.5 * (R_Mat_ADP^--1) * B' * P_Mat;
205
       end
206
       % Update the inputs using ADP
207
       uNewADP = kADP * errorADP(:, k);
208
       % Limit the voltages
209
       uNewADP(1) = sign(uNewADP(1)) * min(abs(uNewADP(1)), maxVolt);
       uNewADP(2) = sign(uNewADP(2)) *min(abs(uNewADP(2)), maxVolt);
211
       % Update the input matrices
212
       uADP(1,k) = uNewADP(1);
213
       uADP(2,k) = uNewADP(2);
214
215
       Vin=[uNewADP(1) uNewADP(2) rad2deg(xd(1,k)) rad2deg(xd(2,k))];
216
        packedData=vrep.simxPackFloats(Vin);
217
        vrep.simxSetStringSignal(clientID,...
218
                                    'MATLAB_SIG',...
219
                                   packedData , ...
                                   vrep.simx_opmode_oneshot);
221
222
      pause(tau);
223
224
      t = t + tau;
225
226
227
   end
228
   toc;
229
230
_{231} % ret = 1;
   while (ret \tilde{} = 0)
232
        [ret]=vrep.simxClearStringSignal(clientID,...
233
                                            'MATLAB_SIG',...
234
                                           vrep.simx_opmode_blocking);
   end
236
   else
238
       disp('Failed connecting to remote API server');
239
   end % if (clientID >-1)
240
241
   disp('Sim ended');
242
243
244
245 figure
plot (dt, rad2deg(pose(1, :))); hold on;
plot (dt, rad2deg(xd(1,:))); hold off;
   title('pitch')
248
249
250 figure
<sup>251</sup> plot (dt, rad2deg(pose(2,:))); hold on;
<sup>252</sup> plot (dt, rad2deg (xd (2,:))); hold off;
253 title ('yaw')
```

```
254
255 figure
256 plot (dt, uADP(1,:)); hold on;
_{257} plot (dt, uADP(2,:)); hold off;
258 title ('Inputs')
259
260 % SAVE OUTPUTS FOR PLOTTING
261 % Save the time, error, input, states and desired states vectors
262 save ([testName, '.mat'], 'dt', 'errorADP', 'uADP', 'xADP', 'xd', 'tau', 'R_Mat_ADP',
       'Q_Mat_ADP');
263 % Notify that the simulation is complete
<sup>264</sup> disp ('SIMULATION COMPLETE');
265
266 % ANGLE LIMITER FOR YAW FUNCTION
   function angle = angleLimiterYaw(angle)
267
   % This function limits the angle between -pi and pi
268
        angle = mod(angle, 2*pi);
269
270
        i=find (angle>pi);
271
        angle(i) = angle(i) - 2*pi;
272
273
274
        i=find (angle <- pi);
        angle(i) = angle(i) + 2*pi;
275
276 end
277
278 % ANGLE LIMITER FOR PITCH FUNCTION
_{279} function angle = angleLimiterPitch (angle)
280 % This function limits the physical constraint of the pitch measurement to
   \% 90 degrees
281
        if (angle < 0)
282
            angle = max(angle, -pi/2);
283
        end
284
285
        if (angle > 0)
286
             angle = min(angle, pi/2);
287
        end
288
   end
289
290
291 % CRITIC WEIGHT TUNING NEURAL NETWORK
292 % This function is specific to the helicopter because of the number of
293 % weights and error model
<sup>294</sup> function weights = quanserAEROCriticTuningInitial(A, B, e_vec, tau, R_Mat,
       Q_Mat, xd)
       % REFERENCE DR. MIAH'S PAPER FOR EQUATION NUMBERS
295
296
       % SYSTEM PARAMETERS SPECIFIC TO THE 2-DOF QUANSER AERO
297
298 %
          A = \begin{bmatrix} 0 & 0 & 1 & 0; & 0 & 0 & 0 & 1; & -1.7442 & 0 & -0.3307 & 0; & 0 & 0 & 0 & -0.9283 \end{bmatrix};
          B = \begin{bmatrix} 0 & 0; & 0 & 0; & 0.0512 & 0.0977; & -0.1139 & 0.0928 \end{bmatrix};
299
        % System dimensions specific for our model
300
        [n, \tilde{}] = size(B);
301
302
        % ERROR MODEL OF THE HELICOPTER
303
        \% fbar and gbar --- EQ 8
304
305
        fbar = @(e) A * e;
```

```
306
       gbar = -B;
       hbar = -A*xd;
307
308
       % DISCRETE-TIME ERROR MODEL FOR TIME TAU
309
       % f and g — EQ 10
310
       f = @(e) fbar(e) *tau + e;
311
       g = gbar * tau;
312
       h = hbar*tau;
313
314
       % COST FUNCTION PARAMETERS
       \% State penalizing function in the continuous cost function
316
       Qbar = @(e) e' * Q_Mat * e;
317
       % Control penalizing matrix in the continuous cost function
318
       Rbar = R_Mat;
       % The discrete-time cost function will have terms:
       % Right after EQ 11 in paper
321
       % State penalizing function in the discretized cost function
322
       Q = @(e) \quad Qbar(e) *tau;
323
       % Control penalizing matrix in the discretized cost function
324
       R = Rbar*tau;
325
326
       % NEURAL NEIWORK FUNCTIONS
328
       % Critic neural network activation functions
       rho = @(e) [e(1); e(2); e(3); e(4);...
329
                     e(1)^{2}; e(1) * e(2); e(1) * e(3); e(1) * e(4); \dots
330
                     e(2)^{2}; e(2) * e(3); e(2) * e(4); e(3)^{2}; e(3) * e(4); e(4)^{2};
       % Partial derivative of rho with respect to e
332
       drhode = @(e) [1, 0, 0; 0];
333
                       0, 1, 0, 0;
334
                       0, 0, 1, 0;
335
                       0, 0, 0, 1;
336
                       2 * e(1), 0, 0, 0;
337
                       e(2), e(1), 0, 0;
338
                       e(3), 0, e(1), 0;
339
                       e(4), 0, 0, e(1);
340
                       0, 2*e(2), 0, 0;
341
                       0, e(3), e(2), 0;
342
                       0, e(4), 0, e(2);
343
                       0, 0, 2*e(3), 0;
344
                       0, 0, e(4), e(3);
345
                       0, 0, 0, 2*e(4)];
346
347
       % TOLERANCES
348
       % Convergence tolerance for control policy
       EpsilonPolicy = 0.1;
350
       % Convergence tolerance for critic neural network
351
       EpsilonWcritic = 0.1;
352
353
       % TRAINING PARAMETERS
354
       % Number of outer loop iterations
355
       outerLoopMax = 700;
356
       % Number of inner loop iterations
357
       innerLoopMax = 100;
358
       % Number of equations needed for training, number of sub-intervals
359
```

```
% Number of training samples
360
       [, nbar] = size(e_vec);
361
362
       % WEIGHT INITIALIZATION
363
       % Initialize the weights of the critic neural network to zero
364
       WcLast = zeros(length(rho(e_vec(:,1))), 1);
365
366
       % LEAST-SQUARES COMPUTATION INITIALIZATION
367
       % Matrices required for computing least squares weights of the critic
368
       % neural networks --- EQ 20
369
       V = zeros(nbar, 1);
       Lambda = zeros(nbar, length(rho(e_vec(:,1))));
371
372
       \% Matrix to hold the derivative of the error model during policy
       % updating
374
       e_k_plus_1 = zeros(n, nbar);
375
376
       \% Product of the least squares matrices must be invertible
377
       % Logic flag indicating if the critic weights are unsolvable
378
       \% The weights are unsolvable because the least squares matrices have no
379
       % solution -- not invertible
380
       diverged = 0;
381
382
       % OUTER LOOP
383
       for i = 1:(outerLoopMax - 1)
384
       % Determine if the least squares matrices are invertible
       if diverged = 0
386
           \% For each of the data collection (discrete time index)
387
           for k = 1:nbar
388
                % Initialize the optimal inputs to zero
389
                uNew = [0; 0];
390
               % INNER LOOP
391
                for j = 1:(innerLoopMax - 1)
392
                    % Get the updated input value
393
                    uLast = uNew;
394
                    % Update the error model
395
                    e_{k_{plus_{1}}(:,k)} = f(e_{vec}(:,k)) + g*uLast + h;
396
                    % Compute the new optimal inputs
397
                    uNew = -0.5*(R^{(-1)})*g'*drhode(e_k_plus_1(:,k))'*WcLast;
398
399
                    % Check convergence of the optimal inputs
400
                    if norm(uNew - uLast) < EpsilonPolicy
401
                         break;
402
                    end
403
                end
404
405
               \% Update the values for the least-squares computation
406
                V(k,:) = Q(e_vec(:,k)) + uNew' * R * uNew + WcLast' * rho(e_k_plus_1)
407
       (:,k));
                Lambda(k,:) = rho(e_vec(:,k))';
408
           end
409
       end
410
411
       \% Verify the least square solution exists for the critic's weights
412
```

```
\% If the error data is consistent or there is no error, this will not
413
        % hold, so set the weights to zero
414
        if det (Lambda' * Lambda) == 0
415
             fprintf ('AWESOME...YOU HAVE NO ERROR...I'''M GOING TO SET THE WEIGHTS
416
        TO ZERO\langle n' \rangle;
             weights = \operatorname{zeros}(\operatorname{length}(\operatorname{rho}(\operatorname{e_vec}(:,1))),1);
417
             break;
        end;
419
        \% Calculate least squares solution of critic's weights — EQ 20
421
        WcNew = (Lambda' * Lambda)^{(-1)} * Lambda' * V;
422
        % Make sure the weights did not diverge
423
        \% If the weights are diverging, just set them to a large number
424
        if isnan (WcNew)
425
             fprintf('OOPS...DIVERGING WEIGHTS...I''M GOING TO USE LARGE WEIGHTS\
426
       n ');
             weights = 1000 * \text{ones}(\text{length}(\text{rho}(\text{e_vec}(:,1))), 1);
427
             break;
428
        end;
429
430
        % Check for convergence of the critic weights
431
        if norm(WcNew - WcLast) < EpsilonWcritic
432
             weights = WcNew;
433
             fprintf('GREAT...THE WEIGHTS CONVERGED\n');
434
             break;
435
        end
        % If the weights did not converge, repeat the loop
437
        WcLast = WcNew:
438
439
        \% If we reached the last iteration of the loop, just use the last
440
       % weights found
441
        if (i = (outerLoopMax - 1))
442
             fprintf ('OOPS...YOU REACHED THE END OF THE OUTER LOOP...I''M GOING
443
       TO USE THE LAST VALUE\langle n' \rangle;
             weights = WcNew;
444
        end
445
        end
446
447 end
448
449 % CRITIC WEIGHT TUNING NEURAL NETWORK
450 % This function is specific to the helicopter because of the number of
451 % weights and error model
   function weights = quanserAEROCriticTuning(A, B, e_vec, tau, R_Mat, Q_Mat, xd,
452
       wcInit)
       % REFERENCE DR. MIAH'S PAPER FOR EQUATION NUMBERS
453
454
       % SYSTEM PARAMETERS SPECIFIC TO THE 2–DOF QUANSER AERO
455
456 %
          A = \begin{bmatrix} 0 & 0 & 1 & 0; & 0 & 0 & 0 & 1; & -1.7442 & 0 & -0.3307 & 0; & 0 & 0 & 0 & -0.9283 \end{bmatrix};
          B = \begin{bmatrix} 0 & 0; & 0 & 0; & 0.0512 & 0.0977; & -0.1139 & 0.0928 \end{bmatrix};
457
  %
       % System dimensions specific for our model
458
        [n, \tilde{}] = size(B);
459
460
        % ERROR MODEL OF THE HELICOPTER
461
       % fbar and gbar — EQ 8
462
```

```
fbar = @(e) A * e;
463
       gbar = -B;
464
       hbar = -A*xd;
465
466
       % DISCRETE-TIME ERROR MODEL FOR TIME TAU
467
       % f and g — EQ 10
468
       f = @(e) fbar(e) *tau + e;
469
       g = gbar * tau;
470
       h = hbar*tau;
471
472
       % COST FUNCTION PARAMETERS
473
       % State penalizing function in the continuous cost function
474
       Qbar = @(e) e' * Q_Mat * e;
475
       % Control penalizing matrix in the continuous cost function
476
       Rbar = R_Mat;
477
       % The discrete-time cost function will have terms:
478
       % Right after EQ 11 in paper
479
       % State penalizing function in the discretized cost function
480
       Q = @(e) \quad Qbar(e) *tau;
481
       % Control penalizing matrix in the discretized cost function
482
       R = Rbar*tau;
483
484
       % NEURAL NEIWORK FUNCTIONS
485
       % Critic neural network activation functions
486
       rho = @(e) [e(1); e(2); e(3); e(4);...
487
                     e(1)^{2}; e(1) * e(2); e(1) * e(3); e(1) * e(4); \dots
                     e(2)^{2}; e(2) * e(3); e(2) * e(4); e(3)^{2}; e(3) * e(4); e(4)^{2};
489
       % Partial derivative of rho with respect to e
490
       drhode = @(e) [1, 0, 0; 0;
491
                       0, 1, 0, 0;
492
                       0, 0, 1, 0;
493
                       0, 0, 0, 1;
494
                       2 * e(1), 0, 0, 0;
495
                       e(2), e(1), 0, 0;
496
                       e(3), 0, e(1), 0;
497
                       e(4), 0, 0, e(1);
498
                       0, 2 * e(2), 0, 0;
499
                       0, e(3), e(2), 0;
500
                       0, e(4), 0, e(2);
501
                       0, 0, 2*e(3), 0;
502
                       0, 0, e(4), e(3);
503
                       0, 0, 0, 2 * e(4)];
504
505
       % TOLERANCES
506
       % Convergence tolerance for control policy
       EpsilonPolicy = 0.1;
508
       % Convergence tolerance for critic neural network
509
       EpsilonWcritic = 0.1;
       % TRAINING PARAMETERS
512
       % Number of outer loop iterations
       outerLoopMax = 100;
514
       % Number of inner loop iterations
       innerLoopMax = 100;
```

```
% Number of equations needed for training, number of sub-intervals
       % Number of training samples
518
       [, nbar] = size(e_vec);
519
       % WEIGHT INITIALIZATION
       \% Initialize the weights of the critic neural network to zero
       WcLast = zeros(length(rho(e_vec(:,1))), 1);
       % LEAST-SQUARES COMPUTATION INITIALIZATION
       \% Matrices required for computing least squares weights of the critic
       % neural networks --- EQ 20
       V = zeros(nbar, 1);
528
       Lambda = zeros(nbar, length(rho(e_vec(:,1))));
529
530
       \% Matrix to hold the derivative of the error model during policy
       % updating
       e_k_plus_1 = zeros(n, nbar);
533
534
       % Product of the least squares matrices must be invertible
       % Logic flag indicating if the critic weights are unsolvable
536
       \% The weights are unsolvable because the least squares matrices have no
       % solution -- not invertible
538
       diverged = 0;
540
       % OUTER LOOP
541
       for i = 1:(outerLoopMax - 1)
       % Determine if the least squares matrices are invertible
543
       if diverged = 0
544
           \% For each of the data collection (discrete time index)
545
           for k = 1:nbar
546
               % Initialize the optimal inputs to zero
547
               uNew = [0; 0];
548
               % INNER LOOP
549
                for j = 1:(innerLoopMax - 1)
                   % Get the updated input value
                    uLast = uNew;
                    % Update the error model
553
                    e_k_plus_1(:,k) = f(e_vec(:,k)) + g*uLast + h;
554
                    % Compute the new optimal inputs
                    uNew = -0.5*(R^{(-1)})*g'*drhode(e_k_plus_1(:,k))'*WcLast;
556
                    % Check convergence of the optimal inputs
558
                    if norm(uNew - uLast) < EpsilonPolicy
559
                        break;
560
                    end
561
               end
562
563
               % Update the values for the least-squares computation
564
               V(k,:) = Q(e_vec(:,k)) + uNew' * R * uNew + WcLast' * rho(e_k_plus_1)
565
       (:, k));
               Lambda(k,:) = rho(e_vec(:,k))';
566
           end
567
       end
568
569
```

```
\% Verify the least square solution exists for the critic's weights
570
       \% If the error data is consistent or there is no error, this will not
       \% hold, so set the weights to what they were initially before the
572
       % simulation
573
       if det (Lambda' \ast Lambda) == 0
574
           %fprintf('AWESOME...YOU HAVE NO ERROR...I''M GOING TO USE THE
      ORIGINAL WEIGHTS\langle n' \rangle;
            weights = wcInit;
           break;
       end;
578
       \% Calulcate least squares solution of critic's weights — EQ 20
580
       WcNew = (Lambda'*Lambda)^{(-1)}Lambda'*V;
581
       % Make sure the weights did not diverge
       \% If the weights diverged, set the weights to what they were initially
       % before the simulation
584
       if isnan (WcNew)
585
           %fprintf('OOPS...DIVERGING WEIGHTS...I''M GOING TO USE THE ORIGINAL
      WEIGHTS\n');
           weights = wcInit;
587
           break;
588
       end;
589
590
       % Check for convergence of the critic weights
591
       if norm(WcNew - WcLast) < EpsilonWcritic
592
           %fprintf('GREAT...THE WEIGHTS CONVERGED\n');
            weights = WcNew;
594
           break;
595
       end
596
       \% If the weights did not converge, do another iteration of the loop
       WcLast = WcNew;
598
       \% If we reach the last iteration of the loop, just use the last weights
600
       if (i == (outerLoopMax - 1))
601
           %fprintf('OOPS...YOU REACHED THE END OF THE OUTER LOOP...I''M GOING
602
      TO USE THE LAST VALUE\langle n' \rangle;
            weights = WcNew;
603
       end
604
       end
605
606 end
```

B.5 Motor Approach

B.5.1 Quanser AERO Base

```
1 -- Initialization
2 if (sim_call_type=sim.syscb_init) then
3 simRemoteApi.start(19999)
4
5 pJ = sim.getObjectHandle('pitchJoint');
6 yJ = sim.getObjectHandle('yawJoint');
```

```
prop0 = sim.getObjectHandle('motor0');
      prop1 = sim.getObjectHandle('motor1');
8
      timestep = sim.getSimulationTimeStep()
9
      print('timestep is : ', timestep)
      -- Get parameter values to work with
11
      Jy = sim.getScriptSimulationParameter(sim.handle_self, 'Jy')
      print ('Jy is : ', Jy)
13
      Jp = sim.getScriptSimulationParameter(sim.handle_self, 'Jp')
14
      print('Jp is : ', Jp)
      Dp = sim.getScriptSimulationParameter(sim.handle_self, 'Dp')
      print('Dp is : ', Dp)
      Dy = sim.getScriptSimulationParameter(sim.handle_self, 'Dy')
18
      print('Dy is : ', Dy)
19
      Kpp = sim.getScriptSimulationParameter(sim.handle_self, 'Kpp')
      print('Kpp is : ', Kpp)
21
      Kyy = sim.getScriptSimulationParameter(sim.handle_self, 'Kyy')
22
      print ('Kyy is : ', Kyy)
23
    Ksp = sim.getScriptSimulationParameter(sim.handle_self, 'Ksp')
^{24}
      print('Ksp is : ', Ksp)
25
26
      -- Store current measured angular velocity to calculate angular
      acceleration
    prev_pJ_pose = sim.getJointPosition(pJ)
28
    prev_yJ_pose = sim.getJointPosition(yJ)
29
30
      sim.setJointTargetVelocity(pJ, 0)
      sim.setJointTargetVelocity(yJ, 0)
32
      sim.setJointTargetVelocity(prop0,0)
33
      sim.setJointTargetVelocity(prop1,0)
34
      pitchTarVel = 0
      yawTarVel = 0
36
      graphHandle=sim.getObjectHandle("Graph")
39
      PitchDesired = sim.getIntegerSignal("PitchDesired")
40
      lastPAcc = 0
41
      lastYAcc = 0
42
43
      sim.setJointTargetVelocity(prop0,1*0.8)
44
      sim.setJointTargetVelocity(prop1,-1*0.8)
45
46
  end
47
48

    Looping code

49
  if
     (sim_call_type=sim.syscb_actuation) then
50
         local t = sim.getSimulationTime()
         - Receive velocities from server
      local packedData=sim.getStringSignal('MATLAB_SIG')
54
      if packedData then
           sim.clearStringSignal('MATLAB_SIG') -- Clear the signal
56
```

A. Birge & A. Fandel (Bradley University)

local V=sim.unpackFloatTable(packedData, 0, 4, 0)

```
58
          --- Measure pitch and yaw
59
       pJ_pose = sim_getJointPosition(pJ)
60
      yJ_{pose} = sim_{get}JointPosition(yJ)
61
62
           --- Calculate Velocites
63
           pJ_vel = (pJ_pose - prev_pJ_pose)/timestep
64
           yJ_vel = (yJ_pose - prev_yJ_pose)/timestep
65
66
           pAcc = -(Ksp/Jy)*math.sin(pJ_pose) + -(Dy/Jy)*pJ_vel + -(Kpp/Jy)*V
67
      [1] + (Kpp/Jy) *V[2]
           yAcc = -(Dp/Jp) * yJ_vel + -(Kyy/Jy) * V[1] + -(Kyy/Jy) * V[2]
68
69
70
           pitchTarVel = pitchTarVel + ((pAcc+lastPAcc)/2)*timestep
           yawTarVel = yawTarVel + ((yAcc+lastYAcc)/2)*timestep
72
73
           sim.setJointTargetVelocity(pJ, pitchTarVel)
74
           sim.setJointTargetVelocity(yJ, yawTarVel)
75
76
           prev_pJ_pose = pJ_pose
           prev_yJ_pose = yJ_pose
78
           lastPAcc = pAcc
79
           lastYAcc = yAcc
80
           sim.setGraphUserData(graphHandle, 'PitchDesired', V[3])
81
           sim.setGraphUserData(graphHandle, 'YawDesired', V[4])
82
83
          sim.setJointTargetVelocity(prop0,V[1]*0.8)
84
          sim.setJointTargetVelocity(prop1,-V[2]*0.8)
85
86
      end
87
88
  end
89
90
91 function sysCall_sensing()
92 end
93
  if (sim_call_type=sim.syscb_cleanup) then
94
95
96 end
```

B.5.2 Quanser AERO Motor#0

```
1 function getI(V)
2
3 a = 0.0000277
4 b = -0.00000966
5 c = 0.0213
6 d = -0.00319
7 I = a*V*V*V + b*V*V + c*V + d;
8 return I
9 end
10
11 if (sim_call_type=sim.syscb_init) then
```

12

```
-- Get Handles for Propeller Objects and simulation timestep
      propeller_obj=sim.getObjectAssociatedWithScript(sim.handle_self)
14
      propeller_respondable=sim.getObjectHandle('propeller_respondable');
      timestep = sim.getSimulationTimeStep()
16
    -- Input voltage [Control]
18
      Vin = sim.getScriptSimulationParameter(sim.handle_self, 'Vin')
19
     - Constants from Quanser Model
20
    -- Thrust Coefficient in pitch
21
    Kpp = sim.getScriptSimulationParameter(sim.handle_self, 'Kpp')
     - Drag coefficient [m/(rad/s)]
23
    Kd = sim.getScriptSimulationParameter(sim.handle_self, 'Kd')
24
    - Motor internal resistance [ohms]
      Rm = sim.getScriptSimulationParameter(sim.handle_self, 'Rm')
26
     - Motor torque constant [same as Km] [N*m/A]
27
      Kt = sim.getScriptSimulationParameter(sim.handle_self, 'Kt')
28
      - Motor back EMF constant [same as Kt] [V/(rad/s)]
      Km = sim.getScriptSimulationParameter(sim.handle_self, 'Km')
30
     − Rotor Inertia [kg*m<sup>2</sup>]
31
      Jm = sim.getScriptSimulationParameter(sim.handle_self, 'Jm')
32
     − Hub Inertia [kg*m<sup>2</sup>]
    Jh = sim.getScriptSimulationParameter(sim.handle_self, 'Jh')
34
    - Propeller Inertia [kg*m<sup>2</sup>]
35
    Jp = sim.getScriptSimulationParameter(sim.handle_self, 'Jp')
36
      - Inductance [H]
    L = sim.getScriptSimulationParameter(sim.handle_self, 'L')
38
      ---Propeller Center Radius
39
      R = sim.getScriptSimulationParameter(sim.handle_self, 'R')
40
      momentOfInertia = sim.getScriptSimulationParameter(sim.handle_self,'
42
      motorMomentOfInertia')
43
       - Store current measured angular velocity to calculate angular
44
      acceleration
    prev_i_m = 0
45
46
  end
47
48
  if (sim_call_type=sim.syscb_actuation) then
49
50
      -- Get the transformation matrix relative to global coordinates
      m=sim.getObjectMatrix(propeller_respondable,-1)
         assign Vin (this is set by the Aero script) [Volts]
54
      Vin=sim.getScriptSimulationParameter(sim.handle_self,'Vin')
56
     - calc motor current [Amps]
    i_m = getI(Vin)
59
     - estimate di_dt for inductance calc in KVL loop
60
    di_dt = (i_m - prev_i_m)/timestep
61
62
     - estimate motor rpm [rad/s]
63
```

```
omega\_motor = (Vin - Rm*i\_m - L*di\_dt)/(Km)
64
65
      - set the thrust equal to Kpp*Vin
66
     thrust = -Kpp * Vin/R
67
68
     ---assign 3 dimensional force
69
       force = \{0.0, 0.0, thrust\}
70
71
       -- Get rotation matrix and its inverse
72
       m_{rotation} = m
73
74
       -- remove translational component from
75
       -- function return
76
       m_{rotation}[4]=0
       m_{rotation}[8]=0
78
       m_{rotation}[12]=0
79
80

    Calculate the coupling torque

81
     torque_m = Km*i_m
82
83
       -- Calculate Rotational Drag Torque (negative if clockwise and positive
84
       if cclockwise)
       torque_d = Kd*omega_motor
85
86
       -- Calculate total Torque
       torque = \{0.0, 0.0, 0.0\}
88
       torque[3] = torque_m + torque_d
89
90
       if (mccw) then
91
            torque[3] = -torque[3]
92
93
       end
94
95
       -- Apply force and torque onto the body
96
       force=sim.multiplyVector(m_rotation, force)
97
       torque=sim.multiplyVector(m_rotation, torque)
98
99
       sim.addForceAndTorque(propeller_respondable, force, torque)
100
       --- Store current
     prev_i_m = i_m
103
104 end
106 if (sim_call_type=sim.syscb_cleanup) then
       -- Do nothing
107
108 end
```

B.5.3 Quanser AERO Motor#1

```
1 function getI(V)

2 a = 0.0000277

4 b = -0.00000966

5 c = 0.0213
```

```
d = -0.00319
    I = a * V * V * V + b * V * V + c * V + d;
    return I
8
9 end
  if (sim_call_type=sim.syscb_init) then
      -- Get Handles for Propeller Objects and simulation timestep
13
      propeller_obj=sim.getObjectAssociatedWithScript(sim.handle_self)
14
      propeller_respondable=sim.getObjectHandle('propeller_respondable');
      timestep = sim.getSimulationTimeStep()
16
      - Input voltage [Control]
18
      Vin = sim.getScriptSimulationParameter(sim.handle_self, 'Vin')
19
20
     - Constants from Quanser Model
21
      - Thrust Coefficient in pitch
2.2
    Kpp = sim.getScriptSimulationParameter(sim.handle_self, 'Kpp')
23
     - Drag coefficient [m/(rad/s)]
24
    Kd = sim.getScriptSimulationParameter(sim.handle_self, 'Kd')
25
    - Motor internal resistance [ohms]
26
      Rm = sim.getScriptSimulationParameter(sim.handle_self, 'Rm')
27
    -- Motor torque constant [same as Km] [N*m/A]
28
      Kt = sim.getScriptSimulationParameter(sim.handle_self, 'Kt')
29
      - Motor back EMF constant [same as Kt] [V/(rad/s)]
30
      Km = sim.getScriptSimulationParameter(sim.handle_self, 'Km')
      - Rotor Inertia [kg*m^2]
32
      Jm = sim.getScriptSimulationParameter(sim.handle_self, 'Jm')
33
     - Hub Inertia [kg*m^2]
34
    Jh = sim.getScriptSimulationParameter(sim.handle_self, 'Jh')
35
    -- Propeller Inertia [kg*m<sup>2</sup>]
36
    Jp = sim.getScriptSimulationParameter(sim.handle_self, 'Jp')
     - Inductance [H]
38
    L = sim.getScriptSimulationParameter(sim.handle_self, 'L')
39
       ---Propeller Center Radius
40
      R = sim.getScriptSimulationParameter(sim.handle_self, 'R')
41
42
      momentOfInertia = sim.getScriptSimulationParameter(sim.handle_self,'
43
      motorMomentOfInertia')
44
      -- Store current measured angular velocity to calculate angular
45
      acceleration
    prev_i_m = 0
46
47
  \operatorname{end}
48
49
  if
     (sim_call_type=sim.syscb_actuation) then
50
      spinDir = mccw
      --- Get the transformation matrix relative to global coordinates
53
      m=sim.getObjectMatrix(propeller_respondable, -1)
54
      -- assign Vin (this is set by the Aero script) [Volts]
56
      Vin=sim.getScriptSimulationParameter(sim.handle_self, 'Vin')
57
```

58

```
- calc motor current [Amps]
59
     i_{-m} = getI(Vin)
60
61
     --- estimate di_dt for inductance calc in KVL loop
62
     di_dt = (i_m - prev_i_m)/timestep
63
64
     - estimate motor rpm [rad/s]
65
     omega_motor = (Vin - Rm*i_m - L*di_dt)/(Km)
66
67
      - set the thrust equal to Kpp*Vin
68
     thrust = -Kpp * Vin/R
69
70
     ---assign 3 dimensional force
       force = \{0.0, 0.0, thrust\}
73
       -- Get rotation matrix and its inverse
74
       m_{rotation} = m
76
        - remove translational component from
77
       -- function return
78
       m_{rotation}[4]=0
79
       m_{rotation}[8]=0
80
       m_{rotation}[12]=0
81
82
      - Calculate the coupling torque
83
     torque_m = -Km*i_m
84
85
       -- Calculate Rotational Drag Torque (negative if clockwise and positive
86
      if cclockwise)
       torque_d = -Kd*omega_motor
87
88
       -- Calculate total Torque
89
       torque = \{0.0, 0.0, 0.0\}
90
       torque[3] = torque_m + torque_d
91
92
93
94
       - Apply force and torque onto the body
95
       force=sim.multiplyVector(m_rotation, force)
96
       torque=sim.multiplyVector(m_rotation, torque)
97
98
       sim.addForceAndTorque(propeller_respondable, force, torque)
99
100
       -- Store current
     prev_i_m = i_m
103 end
104
105 if (sim_call_type=sim.syscb_cleanup) then
       — Do nothing
106
107 end
```

B.5.4 Quanser AERO Motor PID Control

```
1 % Use exact model of the 2-DOF helicopter and the linearized model to find
2 % error data points
3 % Use the error data points directly in the neural network instead of
4 % randomizing the error
5 close all; clear; clc;
6
7 % Name of .mat file to save the data
s testName = 'heli_joint_sqsq';
9
10 % LINEAR MODEL MATRIX PARAMETERS (2DoF Helicopter)
11 [A, B] = getAeroAB(0);
12
13 % 2–DOF QUANSER HELICOPTER PARAMETERS
14 % Maximum applied voltage for the rotor motors
_{15} maxVolt = 24;
16 % Number of state variables - theta, psi, dtheta, dpsi
17 n = 4;
_{18} \text{ theta} = 1;
                  % Pitch
19 psi = 2;
                  % Yaw
20 thetaDot = 3; \% Pitch angular velocity
21 psiDot = 4; % Yaw angular velocity
_{22}\ \% Number of inputs – Vp, Vy
_{23} m = 2;
24
25 % SIMULATION TIMING
26 % Initial and final simulation times [s]
27 t0 = 0; tf = 30;
28 % Sampling time [s]
_{29} tau = 0.05;
30 % Larger sampling time for updating the inputs [s]
_{31} T = 0.2;
32 % Number of time steps
_{33} tsteps = floor((tf-t0)/tau);
34 % Discrete time vecotor of sampling time (tau)
dt = tau * (0: tsteps);
36
37
38 % INITIAL STATES
39 \% \text{ xInit} = [\deg 2 \operatorname{rad}(0) \ \deg 2 \operatorname{rad}(0) \ (0 * \operatorname{pi}/180) * 10 \ (0 * \operatorname{pi}/180) * 7]';
40 \%xInit = \begin{bmatrix} 0 & 0 & 0 \end{bmatrix}';
41
42 % MATRIX INITIALIZATION
43 % Actual states – state variable x
44 pose = zeros(n, tsteps+1);
45
46 % State errors for entire simulation
47 error = zeros(n, tsteps+1);
48 % Actual inputs
49 u = zeros(m, tsteps+1);
50
51 % Desired states
52 % Uncomment for desired states of all zero
53 \%xd = zeros(n, tsteps+1);
_{54} % Uncomment for desired states that switch constant values
```

```
55 \%xd = [\deg 2rad(-30) * ones(1, tsteps+1); \deg 2rad(-80) * ones(1, tsteps+1); zeros
             (1, tsteps+1); zeros(1, tsteps+1)];
 56 \%xd = [deg2rad(10)*ones(1,1000), deg2rad(60)*ones(1,tsteps+1-1000); deg2rad(10)*ones(1,tsteps+1-1000); deg2rad(10)*ones(1,1000); deg2rad(10)*ones(10)*ones(1,1000); deg2rad(10)*ones(1,1000); deg2rad(10)*ones(1,1000); deg2rad(10)*ones(1,1000); deg2rad(10)*ones(10)*ones(10)*ones(10)*ones(10)*ones(10)*ones(10)*ones(10)*ones(1
             (80) * ones (1,1500), deg2rad (-120) * ones (1, tsteps+1-1500); zeros (1, tsteps+1)
             ; zeros(1, tsteps+1)];
 57 % Uncomment for desired states that are sine and cosine waves
        pitchSine = deg2rad(30) * sin(0.5 * dt);
 58
       yawCosine = deg2rad(60) * cos(0.5 * dt);
 59
 60 % pitchSquare = deg2rad(25) * square(0.5 * dt);
 61 % yawSquare = deg2rad(-36)*square(0.25*dt);
 _{62} xd = [pitchSine; yawCosine; zeros(1,tsteps+1); zeros(1,tsteps+1)];
 _{63} % xd = [pitchSquare; yawSquare; zeros(1,tsteps+1); zeros(1,tsteps+1)];
 64
 65
 66 %% connection routine
 ip = '127.0.0.1';
 _{68} port = 19999;
 69
 70 % create a handle to remote API
 71 vrep = remApi('remoteApi'); % using the prototype file (remoteApiProto.m)
 _{72} vrep.simxFinish(-1); % just in case, close all opened connections
 73
     clientID = vrep.simxStart(ip, port, true, true, 5000, 5);
 74
 75 %% simulation
     if (clientID >-1) % if clientID exists
 76
               disp('Connected to remote API server');
 77
 78
 79 % Set up the handles
      |rtn, aero| = vrep.simxGetObjectHandle(clientID,...
 80
                                                                                                'Aero', \dots
 81
                                                                                                 vrep.simx_opmode_blocking)
 82
      [rtn, yJ] = vrep.simxGetObjectHandle(clientID,...
 83
                                                                                        yawJoint',...
 84
                                                                                        vrep.simx_opmode_blocking)
 85
      [rtn, pJ] = vrep.simxGetObjectHandle(clientID,...
 86
                                                                                        pitchJoint',...
 87
                                                                                      vrep.simx_opmode_blocking)
 88
 89
 90 % setup joint position streaming
     vrep.simxGetJointPosition(clientID,...
 91
                                                                   yJ , . . .
 92
                                                                    vrep.simx_opmode_streaming);
 93
     vrep.simxGetJointPosition(clientID,...
 94
                                                                   pJ , . . .
 95
                                                                    vrep.simx_opmode_streaming);
 96
 97
 _{98}% add a delay to let the streaming operations initialize
      pause(1);
 99
      [ret, pose(1,1)] = vrep.simxGetJointPosition(clientID,...
                                                                                                                pJ , . . .
102
                                                                                                                vrep.simx_opmode_buffer);
103
104
105 [ret, pose(2,1)] = vrep.simxGetJointPosition(clientID,...
```

```
106
                                                            yJ , . . .
                                                            vrep.simx_opmode_buffer);
107
108
109
110 pose(3,1) = 0;
111 pose (4, 1) = 0;
112 % Initialize the error
<sup>113</sup> error (:, 1) = xd(:, 1) - pose(:, 1);
114
115 k_d = 3;
116 k_i = 10;
_{117} k_{-}p = 8;
118
119 k_1 = 4;
120 k_2 = 4;
121
_{122} e_theta_sum = 0;
123 e_{psi_{sum}} = 0;
124
125 tic;
126 % RUN THE SIMULATION
  % Start at the next time tau after time zero
127
   for k = 2:tsteps+1
128
       % Current time
129
       t = (k-1)*tau; % generate value discrete time index
130
       \% Step 1 - get/store the state of the quadcopter from VREP
        [ret, pose(1,k)] = vrep.simxGetJointPosition(clientID,...
133
                                                                 pJ , . . .
134
                                                                 vrep.simx_opmode_buffer)
135
       ;
136
        [ret, pose(2,k)] = vrep.simxGetJointPosition(clientID,...
                                                                yJ , . . .
138
                                                                 vrep.simx_opmode_buffer)
139
       ;
140
        pose(3,k) = (pose(1,k) - pose(1,k-1))/tau;
141
        pose(4,k) = (pose(2,k) - pose(2,k-1))/tau;
142
143
        if(k \ge 3)
144
              pose(3,k) = (pose(3,k) + pose(3,k-1) + pose(3,k-2))/3;
145
             pose(4,k) = (pose(4,k) + pose(4,k-1) + pose(4,k-2))/3;
146
       end
147
148
        error(:,k) = xd(:,k) - pose(:,k);
149
        e_theta_sum = e_theta_sum + error(1,k)*tau;
        e_{theta} = k_p * error(1,k) + k_i * e_{theta_sum} + k_d * error(3,k);
153
        e_{psi_sum} = e_{psi_sum} + error(2,k)*tau;
154
        e_{psi} = k_{p} \cdot e_{rror}(2,k) + k_{i} \cdot e_{psi} \cdot sum + k_{d} \cdot e_{rror}(4,k);
156
       u(1,k) = -k_1 * e_1 + e_2 * e_p si;
157
```

```
u(2,k) = -k_2 * e_p si + k_1 * e_t heta;
158
159
        u(1,k) = \operatorname{sign}(u(1,k)) * \min(\operatorname{abs}(u(1,k)), \max \operatorname{Volt});
160
        u(2,k) = sign(u(2,k)) * min(abs(u(2,k)), maxVolt);
161
162
        \% Step 3 – Update Inputs
163
        Vin = [u(1,k) \ u(2,k)];
164
    %
        motorSpeeds = \begin{bmatrix} 20 & 0 \end{bmatrix};
165
        packedData=vrep.simxPackFloats(Vin);
166
        vrep.simxSetStringSignal(clientID,...
                                        'MATLAB_SIG',...
168
                                       packedData , ...
169
                                       vrep.simx_opmode_oneshot);
170
171
       pause(tau);
172
173
       t = t + tau;
174
176
177 end
178 toc;
179
_{180} % ret = 1;
   while (ret \tilde{} = 0)
181
         [ret]=vrep.simxClearStringSignal(clientID,...
182
                                                 'MATLAB_SIG', ...
183
                                                vrep.simx_opmode_blocking);
184
185 end
186
   else
187
        disp('Failed connecting to remote API server');
188
   end % if (clientID >-1)
189
190
   disp('Sim ended');
191
193 figure
   plot(dt,rad2deg(pose(1,:))); hold on;
194
<sup>195</sup> plot (dt, rad2deg(xd(1, :))); hold off;
   title('pitch')
196
197
198 figure
<sup>199</sup> plot (dt, rad2deg(pose(2,:))); hold on;
plot (dt, rad2deg(xd(2, :))); hold off;
   title('yaw')
201
202
203 figure
204 plot(dt, u(1, :)); hold on;
205 plot (dt, u(2,:)); hold off;
   title('Inputs')
206
207
208
209
211
```

212 % SAVE OUTPUTS FOR PLOTTING
213 % Save the time, error, input, states and desired states vectors
214 save([testName,'.mat'],'dt', 'xd','tau','error','pose','u','k_p','k_i','k_d');
215 % Notify that the simulation is complete
216 disp('SIMULATION COMPLETE');

Appendix C QFLEX 2 USB

C.1 QFLEX 2 USB Panel (Tutorial)

1. Verify that all of the proper software is installed as mentioned in Section 6.2.

2. Install the QFLEX 2 USB panel to the Quanser AERO as seen in Figure C.1.



Figure C.1: QFLEX 2 USB panel installed on the Quanser AERO.

- 3. Open the Simulink model that you would like to use.
- 4. Power on the Quanser AERO, and connect your laptop to the QFLEX 2 USB panel with the provided cable. See Figure C.2.



Figure C.2: Connections from your laptop to the Quanser AERO.

5. Build the Simulink model. See Figure C.3.



Figure C.3: Build the Simulink model.

6. Connect to the target. This starts the communication process between Simulink and

the Quanser AERO. See Figure C.4.



Figure C.4: Connect to the target.

- 7. Run the Simulink model. See Figure C.5.
- 8. Observe the Quanser AERO's motion control. See Figure C.6.
- 9. When you are done with the Quanser AERO, stop the model. See Figure C.7.
- 10. You may disconnect the Quanser AERO and close Simulink when finished.

11. **NOTES**

- You should be able to partially adjust the desired positions while the Simulink model is running.
- You really only need to rebuild the model if you make major changes to it which will affect the code generation.
- After you stop the model, you can restart the model, but you have to connect to the target again.
- You will be prompted with errors if you are not physically connected when you try to connect to the target.



Figure C.5: Run the Simulink model.



Figure C.6: Observe the Quanser AERO.



Figure C.7: Stop the Simulink model.

C.2 Initialization Code (MATLAB)

```
1 % Andrew Fandel
_{2} % – FIND THE CRITIC NEURAL NEIWORK WEIGHTS BEFORE RUNNING THE SIMULATION
3 % USING RANDOM ERROR DATA
4 close all; clear; clc;
5
6 % Sampling time [s]
7 tau = 0.01;
8 % Update time [s]
 T = 0.2; 
10
11 % COST FUNCTION
12 % Q and R matrices used in the cost function
<sup>13</sup> Q_Mat_ADP = diag([270 \ 100 \ 1 \ 1]);
<sup>14</sup> R_Mat_ADP = 0.005 * \text{diag}([1 \ 1]);
15
16 % CRITIC WEIGHT TUNING NEURAL NEIWORK
17 % This function is specific to the helicopter because of the number of
18 % weights and error model
19 % REFERENCE DR. MIAH'S PAPER FOR EQUATION NUMBERS
20
21 % SYSTEM PARAMETERS SPECIFIC TO THE 2-DOF QUANSER AERO
^{22} A = [0 0 1 0; 0 0 0 1; -1.7442 0 -0.3307 0; 0 0 0 -0.9283];
B = \begin{bmatrix} 0 & 0; & 0 & 0; & -0.0149 & 0.0414; & -0.0751 & -0.1295 \end{bmatrix};
24 % UNCOMMENT TO USE A STATE-SPACE MODEL THAT IS NOT THE CORRECT ONE DERIVED
_{25} % A = [0 0 1 0; 0 0 0 1; 0 0 -0.3307 0; 0 0 0 -0.9283];
{}_{26} \% B = \begin{bmatrix} 0 & 0; & 0 & 0; & 0.1 & 0.1; & -0.3 & -0.3 \end{bmatrix};
27 % System dimensions specific for our model
[n, ~] = size(B);
29
30 % CREATE RANDOM ERROR DATA TO TRAIN THE CRITIC NEURAL NETWORK
31 % Number of equations needed for training, number of sub-intervals
32 % Number of training samples
_{33} nbar = 100;
34 % Create random error in the range of -pi to pi
_{35} e_vec = (2*(pi)) .*rand(n, nbar) - (pi);
36
37 % ERROR MODEL OF THE HELICOPTER
38 % fbar and gbar --- EQ 8
39 % We do not have an xd term because we won't know the desired state vector
_{40} % when we execute this code
41 % This code can also use anonymous functions because this function is
42 % executed by MATLAB and not Simulink
43 fbar = @(e) A*e;
44 gbar = -B;
45
46 % DISCRETE-TIME ERROR MODEL FOR TIME TAU
_{47} % f and g --- EQ 10
48 f = @(e) fbar(e) *tau + e;
49 g = gbar \startau;
50
51 % COST FUNCTION PARAMETERS
```

```
52 % State penalizing function in the continuous cost function
53 Qbar = @(e) e'*Q_Mat_ADP*e;
54 % Control penalizing matrix in the continuous cost function
55 Rbar = R_Mat_ADP;
56 % The discrete-time cost function will have terms:
57 % Right after EQ 11 in paper
58 % State penalizing function in the discretized cost function
59 Q = @(e) \quad Qbar(e) *tau;
60 % Control penalizing matrix in the discretized cost function
_{61} R = Rbar*tau;
62
63 % NEURAL NETWORK FUNCTIONS
64 % Critic neural network activation functions
65 rho = @(e) [e(1); e(2); e(3); e(4);...
                e(1)^{2}; e(1) * e(2); e(1) * e(3); e(1) * e(4); \dots
66
                e(2)^{2}; e(2) * e(3); e(2) * e(4); e(3)^{2}; e(3) * e(4); e(4)^{2};
67
  % Partial derivative of rho with respect to e
68
  drhode = @(e) [1, 0, 0;
69
                  0, 1, 0, 0;
70
                  0, 0, 1, 0;
71
                  0, 0, 0, 1;
                  2*e(1), 0, 0, 0;
73
                  e(2), e(1), 0, 0;
74
                  e(3), 0, e(1), 0;
                  e(4), 0, 0, e(1);
                  0, 2*e(2), 0, 0;
                  0, e(3), e(2), 0;
78
                  0, e(4), 0, e(2);
79
                  0, 0, 2*e(3), 0;
80
                  0, 0, e(4), e(3);
81
                  0, 0, 0, 2*e(4)];
82
83
84 % TOLERANCES
85 % Convergence tolerance for control policy
86 EpsilonPolicy = 0.1;
87 % Convergence tolerance for critic neural network
88 EpsilonWcritic = 0.1;
89
90 % TRAINING PARAMETERS
91 % Number of outer loop iterations
_{92} outerLoopMax = 700;
93 % Number of inner loop iterations
_{94} innerLoopMax = 100;
95 % Number of equations needed for training, number of sub-intervals
96 % Number of training samples
97 [, nbar] = size (e_vec);
98
99 % WEIGHT INITIALIZATION
100 % Initialize the weights of the critic neural network to zero
WcLast = \operatorname{zeros}(\operatorname{length}(\operatorname{rho}(\operatorname{e_vec}(:,1))),1);
102
103 % LEAST-SQUARES COMPUTATION INITIALIZATION
104 % Matrices required for computing least squares weights of the critic
105 % neural networks --- EQ 20
```

```
106 V = zeros(nbar, 1);
Lambda = zeros(nbar, length(rho(e_vec(:,1))));
108
109 % Matrix to hold the derivative of the error model during policy
110 % updating
e_k_plus_1 = zeros(n, nbar);
112
113 % Product of the least squares matrices must be invertible
114 % Logic flag indicating if the critic weights are unsolvable
115 % The weights are unsolvable because the least squares matrices have no
116 % solution — not invertible
117 diverged = 0;
118
119 % OUTER LOOP
120 for i = 1:(outerLoopMax - 1)
121 % Determine if the least squares matrices are invertible
   if diverged == 0
       % For each of the data collection (discrete time index)
123
       for k = 1:nbar
124
           % Initialize the optimal inputs to zero
125
           uNew = [0; 0];
126
           % INNER LOOP
            for j = 1:(innerLoopMax - 1)
128
               % Get the updated input value
129
                uLast = uNew;
130
                % Update the error model
                e_k_plus_1(:,k) = f(e_vec(:,k)) + g*uLast;
                % Compute the new optimal inputs
                uNew = -0.5*(R^{(-1)})*g'*drhode(e_k_plus_1(:,k))'*WcLast;
134
                % Check convergence of the optimal inputs
136
                if norm(uNew - uLast) < EpsilonPolicy
                     break;
138
                end
           end
140
141
           % Update the values for the least-squares computation
142
           V(k,:) = Q(e_vec(:,k)) + uNew' * R * uNew + WcLast' * rho(e_k_plus_1(:,k))
143
           Lambda(k,:) = rho(e_vec(:,k))';
144
       end
145
146 end
147
148 % Verify the least square solution exists for the critic's weights
149 % If the error data is consistent or there is no error, this will not
150 % hold, so set the weights to zero
  if det (Lambda' * Lambda) == 0
       weights = \operatorname{zeros}(\operatorname{length}(\operatorname{rho}(e_{\operatorname{vec}}(:,1))),1);
       break;
154 end;
155
156 % Calculate least squares solution of critic's weights --- EQ 20
157 WcNew = (Lambda' * Lambda)^{(-1)} * Lambda' * V;
158 % Make sure the weights did not diverge
```
```
159 % If the weights are diverging, just set them to a large number
160 if isnan (WcNew)
        weights = 1000 * \text{ones}(\text{length}(\text{rho}(e_{\text{vec}}(:,1))), 1);
161
        break;
163 end;
164
165 % Check for convergence of the critic weights
  if norm(WcNew - WcLast) < EpsilonWcritic
166
        weights = WcNew;
167
        break;
168
169 end
170 % If the weights did not converge, repeat the loop
171 WcLast = WcNew;
172
173 % If we reached the last iteration of the loop, just use the last
174 % weights found
175 if (i == (outerLoopMax - 1))
        weights = WcNew;
176
177 end
178 end
179
180 % Use the weights to determine the P matrix
  P_Mat = [weights(5) weights(6) weights(7) weights(8);
181
              weights (6) weights (9) weights (10) weights (11);
182
              weights (7) weights (10) weights (12) weights (13);
183
              weights (8) weights (11) weights (13) weights (14)];
184
185
186 % Find the state-feedback gain EQ
_{187} \text{ K} = 0.5 * (\text{R}_{\text{Mat}} \text{ADP}^{-} - 1) * \text{B}' * \text{P}_{\text{Mat}};
188
189 % Save the initial weights
190 wcInit = weights;
191
192 % Keep only the ADP gain
193 clearvars -except K tau T wcInit;
```

C.3 Update ADP Gain (MATLAB)

```
1 % CRITIC WEIGHT TUNING NEURAL NETWORK
_{2} % This function is specific to the helicopter because of the number of
<sup>3</sup> % weights and error model
4 % THIS FUNCTION CANNOT HAVE ANY ANONYMOUS FUNCTIONS DUE TO THE C CODE
5 % GENERATION UNSUPPORTING IT
6 function K = quanserAEROCriticTuning(xd, pitchData, yawData, pitchDotData,
     yawDotData, wcInit)
     % CREATE THE ERROR VECTOR MATRIX
7
     % Use the data from the tapped delay blocks to create the error state
8
     % vector matrix
9
      e_vec = [pitchData'; yawData'; pitchDotData'; yawDotData'];
     % REFERENCE DR. MIAH'S PAPER FOR EQUATION NUMBERS
      % Sampling time [s]
      tau = 0.01;
14
```

```
% COST FUNCTION
16
       % Q and R matrices used in the cost function
       Q_{-Mat} = diag([270 \ 100 \ 1 \ 1]);
18
       R_Mat = 0.005 * diag( \begin{bmatrix} 1 & 1 \end{bmatrix});
19
20
       % SYSTEM PARAMETERS SPECIFIC TO THE 2–DOF QUANSER AERO
       A = \begin{bmatrix} 0 & 0 & 1 & 0; & 0 & 0 & 0 & 1; & -1.7442 & 0 & -0.3307 & 0; & 0 & 0 & 0 & -0.9283 \end{bmatrix};
       B = \begin{bmatrix} 0 & 0; & 0 & 0; & -0.0149 & 0.0414; & -0.0751 & -0.1295 \end{bmatrix};
23
      % UNCOMMENT TO USE A STATE-SPACE MODEL THAT IS NOT THE CORRECT ONE
      DERIVED
      \% A = [1 1 1 1; 1 1 1 1; 1 1 1; 1 1 1; 1 1 1; ];
25
      \% B = [0 \ 0; \ 0 \ 0; \ 0.1 \ 0.1; \ -0.3 \ -0.3];
26
       % System dimensions specific for our model
       [n, \tilde{}] = size(B);
28
       % ERROR MODEL OF THE HELICOPTER
30
       % gbar and hbar --- EQ 8
       % An anonymous function cannot be used for fbar
32
       gbar = -B;
33
       hbar = -A*xd;
34
       % DISCRETE-TIME ERROR MODEL FOR TIME TAU
36
       % g and h — EQ 10
37
       % An anonymous function cannot be used for f
38
       g = gbar * tau;
       h = hbar*tau;
40
41
       % COST FUNCTION PARAMETERS
42
       \% The Q matrix cannot be treated as an anonymous function as in the
      % MATLAB simulations
44
       % Control penalizing matrix in the continuous cost function
45
       Rbar = R_Mat;
       % The discrete-time cost function will have terms:
47
       % Right after EQ 11 in paper
48
       % Control penalizing matrix in the discretized cost function
49
       R = Rbar*tau;
50
      % NEURAL NEIWORK FUNCTIONS
       \% These functions cannot be written as anonymous functions as in the
53
       % MATLAB simulations
54
       % TOLERANCES
56
       % Convergence tolerance for control policy
       EpsilonPolicy = 0.1;
58
       % Convergence tolerance for critic neural network
59
       EpsilonWcritic = 0.1;
60
61
       % TRAINING PARAMETERS
62
       % Number of outer loop iterations
63
       outerLoopMax = 700;
64
       % Number of inner loop iterations
65
       innerLoopMax = 100;
66
       \% Number of equations needed for training, number of sub-intervals
67
```

```
% Number of training samples
68
       [, nbar] = size(e_vec);
69
       % WEIGHT INITIALIZATION
71
       % Initialize the weights of the critic neural network to zero
72
       % Number of rows hard-coded for the helicopter
73
       WcLast = zeros(14, 1);
74
       % LEAST-SQUARES COMPUTATION INITIALIZATION
76
       \% Matrices required for computing least squares weights of the critic
       % neural networks --- EQ 20
78
       V = zeros(nbar, 1);
79
       Lambda = zeros(nbar, 14);
80
       % Matrix to hold the derivative of the error model during policy
82
      % updating
83
       e_k_plus_1 = zeros(n, nbar);
84
       % Product of the least squares matrices must be invertible
86
       % Logic flag indicating if the critic weights are unsolvable
87
       \% The weights are unsolvable because the least squares matrices have no
       % solution -- not invertible
89
       diverged = 0;
90
91
       % OUTER LOOP
92
       for i = 1:(outerLoopMax - 1)
93
       % Determine if the least squares matrices are invertible
94
       if diverged = 0
95
           \% For each of the data collection (discrete time index)
96
           for k = 1:nbar
97
               % Initialize the optimal inputs to zero
98
               uNew = [0; 0];
99
               % INNER LOOP
100
                for j = 1:(innerLoopMax - 1)
                   % Get the updated input value
                    uLast = uNew;
                    % Update the error model
104
                    % Because of no anonymous functions we have modified f
                   % fbar = @(e) A*e;
106
                    \% f = @(e) fbar(e) *tau + e;
107
                    e_k_plus_1(:,k) = (A*e_vec(:,k)*tau) + e_vec(:,k) + g*uLast
108
      + h;
                    % Compute the new optimal inputs
                    % Partial derivative of rho with respect to e
110
                    drhode = [1, 0, 0, 0;
                               0, 1, 0, 0;
                               0, 0, 1, 0;
                               0, 0, 0, 1;
114
                               2 * e_v ec(1,k), 0, 0, 0;
115
                               e_vec(2,k), e_vec(1,k), 0, 0;
116
                               e_vec(3,k), 0, e_vec(1,k), 0;
117
                               e_vec(4,k), 0, 0, e_vec(1,k);
118
                               0, 2*e_vec(2,k), 0, 0;
119
                               0, e_{-}vec(3,k), e_{-}vec(2,k), 0;
120
```

```
0, e_{vec}(4,k), 0, e_{vec}(2,k);
                               0, 0, 2*e_vec(3,k), 0;
                               0, 0, e_{vec}(4, k), e_{vec}(3, k);
123
                               0, 0, 0, 2*e_vec(4,k)];
124
                    uNew = -0.5*(R^{(-1)})*g'*drhode'*WcLast;
126
                    % Check convergence of the optimal inputs
                    if norm(uNew - uLast) < EpsilonPolicy
128
                         break;
129
                    end
130
                end
               % Update the values for the least-squares computation
               % Critic neural network activation functions
134
               rhoE = [e_vec(1,k); e_vec(2,k); e_vec(3,k); e_vec(4,k); ...
                         e_vec(1,k)^2; e_vec(1,k) * e_vec(2,k); \dots
136
                        e_vec(1,k) * e_vec(3,k); e_vec(1,k) * e_vec(4,k); \dots
                        e_vec(2,k) 2; e_vec(2,k) * e_vec(3,k);...
138
                        e_vec(2,k) * e_vec(4,k); e_vec(3,k)^2; \dots
139
                        e_vec(3,k) * e_vec(4,k); e_vec(4,k)^2;
140
              rhoEK = [e_k_plus_1(1,k); e_k_plus_1(2,k); e_k_plus_1(3,k); ...
141
                         e_k_plus_1(4,k); e_k_plus_1(1,k)^2;...
                         e_k_plus_1(1,k) * e_k_plus_1(2,k);...
143
                         e_k_plus_1(1,k) * e_k_plus_1(3,k);...
144
                         e_k_plus_1(1,k) * e_k_plus_1(4,k);...
145
                         e_k_plus_1(2,k)^2; e_k_plus_1(2,k) * e_k_plus_1(3,k); ...
146
                         e_k_plus_1(2,k) * e_k_plus_1(4,k); e_k_plus_1(3,k)^2;...
147
                         e_k_plus_1(3,k) * e_k_plus_1(4,k); e_k_plus_1(4,k)^2];
148
               % State penalizing function in the continuous cost function
149
               \% Qbar = @(e) e'*Q_Mat*e;
               % State penalizing function in the discretized cost function
               \% Q = @(e) Qbar(e) *tau;
               V(k,:) = (e_vec(:,k)'*Q_Mat*e_vec(:,k)*tau) + uNew'*R*uNew +
      WcLast '* rhoEK:
               Lambda(k,:) = rhoE';
154
           end
       end
       \% Verify the least square solution exists for the critic's weights
158
       \% If the error data is consistent or there is no error, this will not
159
       \% hold, so set the weights to what they were initially before the
160
       % simulation
       if det (Lambda' \ast Lambda) == 0
           weights = wcInit;
           break;
164
       end;
165
       \% Calulcate least squares solution of critic's weights — EQ 20
167
       WcNew = (Lambda' * Lambda)^{(-1)} * Lambda' * V;
       % Make sure the weights did not diverge
169
       \% If the weights diverged, set the weights to what they were initially
       % before the simulation
       if isnan (WcNew)
173
           weights = wcInit;
```

```
174
           break;
       % Check for convergence of the critic weights
       elseif norm(WcNew - WcLast) < EpsilonWcritic</pre>
176
           weights = WcNew;
177
           break:
178
       % If we reach the last iteration of the loop, just use the last weights
179
       elseif (i == (outerLoopMax-1))
180
            weights = WcNew;
181
       \% If all else fails, just set the weights to the initial weights
182
       else
183
           weights = wcInit;
184
       end
185
       \% If the weights did not converge, do another iteration of the loop
186
       WcLast = WcNew;
187
       end
188
189
       % Use the weights to determine the P matrix
190
       P_Mat = [weights(5) weights(6) weights(7) weights(8);
                 weights (6) weights (9) weights (10) weights (11);
192
                 weights (7) weights (10) weights (12) weights (13);
                 weights (8) weights (11) weights (13) weights (14) ];
194
       % Find the state-feedback gain EQ
196
       K = 0.5 * (R_Mat^{-1}) *B' * P_Mat;
197
198 end
```

C.4 Base Color (MATLAB)

```
1 % THIS FUNCTION DETERMINES THE PROPER COLOR MATRIX FOR THE QUANSER AERO
<sup>2</sup> function colorSelect = colorSelector(u)
3
4 % If the magnitude is less than 1, output the color green
_{5} if (u < 1)
        colorSelect = \begin{bmatrix} 0 & 1 & 0 \end{bmatrix};
6
_{7} % If the magnitude is between 3 and 10, output the color yellow
colorSelect = \begin{bmatrix} 1 & 1 & 0 \end{bmatrix};
9
10 % If the magnitude is greater than 10, output the color blue
11 else
12
        colorSelect = \begin{bmatrix} 0 & 0 & 1 \end{bmatrix};
13 end
```

Appendix D Raspberry Pi 3

D.1 Raspberry Pi 3 (Tutorial)

1. Verify that all of the proper software is installed as mentioned in Section 6.3.

2. Install the QFLEX 2 Embedded panel to the Quanser AERO as seen in Figure D.1.



Figure D.1: QFLEX 2 Embedded panel installed on the Quanser AERO.

3. Connect your laptop to the Raspberry Pi and power it on. Use an Ethernet cable to connect to the Raspberry Pi. See Figure D.2. You can connect the Raspberry Pi to the Quanser AERO, but I prefer to do that in a later step.



Figure D.2: Connect the Raspberry Pi to your laptop and power it on.

4. Wait a few moments and then connect to the Raspberry Pi using PuTTY. The IP address of our particular Raspberry Pi is 169.254.0.2. See Figure D.3.



Figure D.3: Connect to the Raspberry Pi using PuTTY.

5. Once PuTTY is connected to the Raspberry Pi, login to it. Login as "pi" with the password "raspberry." See Figure D.4.



Figure D.4: Login to the Raspberry Pi.

- 6. We can see that this is the Linux version we installed on the SD card when configuring the support package. See Figure D.5. We can see where the code that will be generated by Simulink will be located.
- 7. Open the Simulink model that you would like to push to the Raspberry Pi.
- 8. Deploy the model to hardware. This will generate the C-code on the Raspberry Pi. See Figure D.6. This build may take a while.
- 9. **NOTE** The Simulink model provided uses an initialization function to enable the SPI communication pins. If you just powered on the Raspberry Pi, you need to uncomment the line that disables SPI. If have already pushed to the Raspberry Pi while it has been on and want to do it again, you need to comment out the line that disables SPI. If you don't do this, there will be errors aborting the build.
- 10. Once the model is pushed to the Raspberry Pi, the model will actually begin running as a process.
- 11. Use the ps -A Linux command to see all the processes running. One should have a similar name of the model; find its process number.
- 12. Once you have the process number, use the sudo kill -9 1234 Linux command to kill the process. 1234 is the process number you currently found.

Ē





Figure D.5: Version of Linux installed on SD card by MATLAB.



Figure D.6: Deploy model to hardware.

13. You now have the generated code on the Raspberry Pi, but it is not currently running.14. Connect the Raspberry Pi to the Quanser AERO. See Figure D.7. The SPI wires

should already be connected to the Raspberry Pi. If not, you will need to consult the MATLAB initialization code and the Raspberry Pi data sheet.



Figure D.7: Connect the Raspberry Pi to the Quanser AERO.

- 15. Execute the .elf file to run the generated code. Use the Linux command sudo ./file.elf. See Figure D.8.
- 16. The generated C-code should now be running on the Raspberry Pi.
- 17. When you are done running the code, repeat the steps used to kill the process. Sometimes you can also use Control-C to stop the running model.



Figure D.8: Execute the code on the Raspberry Pi.

D.2 Initialization Code (MATLAB)

```
1 % Andrew Fandel
2 % – FIND THE CRITIC NEURAL NEIWORK WEIGHTS BEFORE RUNNING THE SIMULATION
3 % USING RANDOM ERROR DATA
4 close all; clear all; clc;
5
6 % Sampling time [s]
7 tau = 0.05;
8 % Update time [s]
9 T = 1;
10
11 % COST FUNCTION
_{12} \ensuremath{\,^{12}} Q and R matrices used in the cost function
<sup>13</sup> Q_Mat_ADP = diag([270 \ 100 \ 1 \ 1]);
<sup>14</sup> R_Mat_ADP = 0.005 * \text{diag}([1 \ 1]);
15
16 % CRITIC WEIGHT TUNING NEURAL NEIWORK
_{17} % This function is specific to the helicopter because of the number of
18 % weights and error model
19 % REFERENCE DR. MIAH'S PAPER FOR EQUATION NUMBERS
20
21 % SYSTEM PARAMETERS SPECIFIC TO THE 2-DOF QUANSER AERO
22 A = \begin{bmatrix} 0 & 0 & 1 & 0; & 0 & 0 & 0 & 1; & -1.7442 & 0 & -0.3307 & 0; & 0 & 0 & 0 & -0.9283 \end{bmatrix};
^{23} B = \begin{bmatrix} 0 & 0; & 0 & 0; & -0.0149 & 0.0414; & -0.0751 & -0.1295 \end{bmatrix};
24 % System dimensions specific for our model
_{25} [n, \tilde{}] = size (B);
26
```

```
27 % CREATE RANDOM ERROR DATA TO TRAIN THE CRITIC NEURAL NETWORK
28 % Number of equations needed for training, number of sub-intervals
29 % Number of training samples
_{30} nbar = 100;
31 % Create random error in the range of -pi to pi
_{32} e_vec = (2*(pi)) .* rand(n, nbar) - (pi);
33
34 % ERROR MODEL OF THE HELICOPTER
35 % fbar and gbar --- EQ 8
_{36} % We do not have an xd term because we won't know the desired state vector
_{37} % when we execute this code
_{38} % This code can also use anonymous functions because this function is
39 % executed by MATLAB and not Simulink
40 fbar = @(e) A*e;
41 gbar = -B;
42
43 % DISCRETE-TIME ERROR MODEL FOR TIME TAU
44 % f and g --- EQ 10
45 f = @(e) fbar(e) *tau + e;
_{46} g = gbar*tau;
47
48 % COST FUNCTION PARAMETERS
49 % State penalizing function in the continuous cost function
50 Qbar = @(e) e'*Q_Mat_ADP*e;
51 % Control penalizing matrix in the continuous cost function
52 Rbar = R_Mat_ADP;
53 % The discrete-time cost function will have terms:
54 % Right after EQ 11 in paper
55 % State penalizing function in the discretized cost function
_{56} Q = @(e) Qbar(e) *tau;
57 % Control penalizing matrix in the discretized cost function
_{58} R = Rbar*tau;
59
60 % NEURAL NEIWORK FUNCTIONS
61 % Critic neural network activation functions
e_{2} rho = @(e) [e(1); e(2); e(3); e(4);...
               e(1)^{2}; e(1) * e(2); e(1) * e(3); e(1) * e(4); \dots
63
               e(2)^{2}; e(2) * e(3); e(2) * e(4); e(3)^{2}; e(3) * e(4); e(4)^{2};
64
65 % Partial derivative of rho with respect to e
  drhode = @(e) [1, 0, 0; 0;
66
                 0, 1, 0, 0;
67
                 0, 0, 1, 0;
68
                 0, 0, 0, 1;
69
                 2 * e(1), 0, 0, 0;
                 e(2), e(1), 0, 0;
                 e(3), 0, e(1), 0;
72
                 e(4), 0, 0, e(1);
73
                 0, 2*e(2), 0, 0;
74
                 0, e(3), e(2), 0;
                 0, e(4), 0, e(2);
76
                 0, 0, 2*e(3), 0;
77
                 0, 0, e(4), e(3);
78
                 0, 0, 0, 2*e(4)];
79
80
```

```
81 % TOLERANCES
82 % Convergence tolerance for control policy
83 EpsilonPolicy = 0.1;
84 % Convergence tolerance for critic neural network
85 EpsilonWcritic = 0.1;
86
87 % TRAINING PARAMETERS
88 % Number of outer loop iterations
so outerLoopMax = 700;
90 % Number of inner loop iterations
91 innerLoopMax = 100;
92 % Number of equations needed for training, number of sub-intervals
93 % Number of training samples
94 [, nbar] = size(e_vec);
95
96 % WEIGHT INITIALIZATION
97 % Initialize the weights of the critic neural network to zero
98 WcLast = zeros(length(rho(e_vec(:,1))),1);
99
100 % LEAST-SQUARES COMPUTATION INITIALIZATION
101 % Matrices required for computing least squares weights of the critic
102 % neural networks --- EQ 20
V = zeros(nbar, 1);
Lambda = \operatorname{zeros}(\operatorname{nbar}, \operatorname{length}(\operatorname{rho}(\operatorname{e_vec}(:,1))));
106 % Matrix to hold the derivative of the error model during policy
107 % updating
e_k_plus_1 = zeros(n, nbar);
109
110 % Product of the least squares matrices must be invertible
111 % Logic flag indicating if the critic weights are unsolvable
112 % The weights are unsolvable because the least squares matrices have no
113 % solution --- not invertible
114 diverged = 0;
115
116 % OUTER LOOP
117 for i = 1:(outerLoopMax - 1)
118 % Determine if the least squares matrices are invertible
  if diverged == 0
119
       % For each of the data collection (discrete time index)
120
       for k = 1:nbar
           % Initialize the optimal inputs to zero
           uNew = [0; 0];
123
           % INNER LOOP
124
            for j = 1:(innerLoopMax - 1)
               % Get the updated input value
126
                uLast = uNew;
               % Update the error model
128
                e_k_plus_1(:,k) = f(e_vec(:,k)) + g*uLast;
               % Compute the new optimal inputs
130
                uNew = -0.5*(R^{(-1)})*g'*drhode(e_k_plus_1(:,k))'*WcLast;
131
               % Check convergence of the optimal inputs
133
                if norm(uNew - uLast) < EpsilonPolicy
134
```

```
break;
135
                 end
136
            end
138
            % Update the values for the least-squares computation
139
            V(k,:) = Q(e_vec(:,k)) + uNew' * R * uNew + WcLast' * rho(e_k_plus_1(:,k))
140
            Lambda(k,:) = rho(e_vec(:,k))';
141
        end
142
143 end
144
145 % Verify the least square solution exists for the critic's weights
146 % If the error data is consistent or there is no error, this will not
147 % hold, so set the weights to zero
   if det (Lambda' * Lambda) == 0
148
        weights = \operatorname{zeros}(\operatorname{length}(\operatorname{rho}(\operatorname{e_vec}(:,1))),1);
149
        break;
  end
151
153 % Calculate least squares solution of critic's weights --- EQ 20
154 WcNew = (Lambda' * Lambda)^{(-1)} * Lambda' * V;
155 % Make sure the weights did not diverge
156 % If the weights are diverging, just set them to a large number
   if isnan (WcNew)
157
        weights = 1000 * \text{ones}(\text{length}(\text{rho}(e_vec(:,1))), 1);
158
        break;
159
160 end
161
162 % Check for convergence of the critic weights
   if norm(WcNew - WcLast) < EpsilonWcritic
163
        weights = WcNew;
164
        break;
165
166 end
167 % If the weights did not converge, repeat the loop
168 WcLast = WcNew;
169
170 % If we reached the last iteration of the loop, just use the last
171 % weights found
if (i = (outerLoopMax -1))
        weights = WcNew;
173
174 end
175 end
176
177 % Use the weights to determine the P matrix
  P_{Mat} = [weights(5) weights(6) weights(7) weights(8);
178
              weights (6) weights (9) weights (10) weights (11);
179
              weights (7) weights (10) weights (12) weights (13);
180
              weights (8) weights (11) weights (13) weights (14) ];
181
182
183 % Find the state-feedback gain EQ
184 \text{ K} = 0.5 * (\text{R}_{\text{Mat}} \text{ADP}^{-} - 1) * \text{B}' * \text{P}_{\text{Mat}};
185
186 % Save the initial weights
wcInit = weights;
```

188

189 % After much trial and error, it seems that when SPI is initiated, the 190 % channel remains high and there is no clock signal 191 % I will create a clock signal and issue the slave select ¹⁹² % Create a connection between MATLAB and the Raspberry Pi $_{193}$ aeroPi = raspi; 194 % Make sure that SPI is disable on the Raspberry Pi so we can use those 195 % same GPIO pins 196 % You may have to comment out this line if you are rebuilding a model and 197 % the SPI is already disabled 198 %disableSPI(aeroPi); 199 % Configure the pins for either inputs or outputs 200 % MOSI – output 201 configurePin(aeroPi,10, 'DigitalOutput'); 202 % MISO – input 203 configurePin(aeroPi,9,'DigitalInput'); 204 % SPI Clock – output 205 configurePin(aeroPi,11, 'DigitalOutput'); $_{206}$ % SS - output 207 configurePin (aeroPi, 8, 'DigitalOutput'); 208 % Specify the period of the SPI clockoutput $_{209}$ % 500 kHz = 2 us $_{210}$ spiPeriod = 0.0001; 211 212 % Keep only the variables pertaining to what we will need in the workspace 213 clearvars -except K tau T wcInit aeroPi spiPeriod;

D.3 SPI Communication (MATLAB)

```
1 % THIS FUNCTION CONTROLS THE SPI INTERFACING
2 % THIS FUNCTION DETERMINES WHAT DATA IS SENT TO THE QUANSER AERO AND WHAT
3 % DATA IS RETRIEVED FROM THE QUANSER AERO
4 function [MOSI, SS, pitchEncoder, yawEncoder, byteNumber, bitNumber,
     encoder2_23_16, encoder2_15_8, encoder2_7_0, encoder3_23_16, encoder3_15_8,
     encoder3_7_0 = fcn (MISO, pitchVolt, yawVolt, redValue, greenValue, blueValue
     , byteNumber, bitNumber, encoder2_23_16, encoder2_15_8, encoder2_7_0,
    encoder3_23_16, encoder3_15_8, encoder3_7_0)
5 % Variables used in the function
6 complement = '000000000000000';
10 \text{ tempVoltBin} = '00000000';
_{12} % Output the value of the pitch encoder using the latest byte values
13 % Combine all of the encoder bytes
14 pitchEncoderIn = [encoder2_23_16 encoder2_15_8 encoder2_7_0];
15 % Convert the encoder binary vector into a character array
 for i = 1:24
16
     if (pitchEncoderIn(i) == 1)
17
         pitchEncoderBin(i) = '1';
18
     else
         pitchEncoderBin(i) = '0';
```

```
21
      end
22 end
23 % Calculate the 2's complement value of the encoder
  if (pitchEncoderBin (1) = (1, 2)
24
       for i = 1:32
25
           if (i < 9)
26
                tempBin(i) = '1';
27
           else
28
                tempBin(i) = pitchEncoderBin(i - 8);
           end
30
31
      end
  else
32
       for i = 1:32
33
           if (i < 9)
34
                tempBin(i) = '0';
35
36
           else
                tempBin(i) = pitchEncoderBin(i - 8);
37
           end
38
       end
39
40 end
41 pitchEncoderTemp = typecast(uint32(bin2dec(tempBin)), 'int32');
42 pitchEncoderOut = cast (pitchEncoderTemp, 'double');
43 pitchEncoder = pitchEncoderOut(1);
  % pitchEncoder = cast (pitchEncoderTemp, 'double');
44
45
46 % Output the value of the yaw encoder using the latest byte values
47 % Combine all of the encoder bytes
_{48} yawEncoderIn = [encoder3_23_16 encoder3_15_8 encoder3_7_0];
  % Convert the encoder binary vector into a character array
49
  for i = 1:24
50
       if (yawEncoderIn(i) == 1)
           yawEncoderBin(i) = '1';
       else
           yawEncoderBin(i) = '0';
54
       end
55
56 end
  % Calculate the 2's complement value of the encoder
57
  if(yawEncoderBin(1) = '1')
58
       for i = 1:32
59
           if (i < 9)
60
                \operatorname{tempBin}(i) = '1';
61
           else
62
                tempBin(i) = yawEncoderBin(i - 8);
63
64
           end
      end
65
  else
66
       for i = 1:32
67
           if (i < 9)
68
                tempBin(i) = '0';
69
           else
70
                tempBin(i) = yawEncoderBin(i - 8);
71
           end
72
       end
73
74 end
```

```
<sup>75</sup> yawEncoderTemp = typecast(uint32(bin2dec(tempBin)), 'int32');
76 yawEncoderOut = cast(yawEncoderTemp, 'double');
yawEncoder = yawEncoderOut(1);
78 % yawEncoder = cast (yawEncoderTemp, 'double ');
79
80 % Determine which byte we are currently sending
81
  % Specify the slave select value
_{82} % – 0 to activate the SPI on the Quanser AERO
_{83} % – 1 to de-activate the SPI on the Quanser AERO
84 % Determine the what the byte to be transmitted should be
  switch byteNumber
85
       case 0
86
       % ** START OF BASE PACKET **
87
       \% BYTE 0
88
       \% MOSI DATA – BASE MODE (0X01)
89
       % MISO DATA – BASE ID MSB
90
       % Beginning of the transmission, so SS goes low
91
       SS = 0;
       readyMOSI = dec2bin(1,8);
93
94
       case 1
95
       \% BYTE 1
96
       \% MOSI DATA – PADDING BYTE (0X00)
97
       % MISO DATA – BASE ID LSB
98
       SS = 0;
90
       readyMOSI = dec2bin(0,8);
       case 2
       \% BYTE 2
       % MOSI DATA – BASE WRITE MASK
104
       % MISO DATA – ENCODER 2 (23-16)
       SS = 0;
       \% Enable the overwriting of encoders 2 and 3 and the LED colors
       % Bit 4 - Set encoder 3 enable
108
       \% Bit 3 – Set encoder 2 enable
       \% Bit 2 – Write blue LED
       \% Bit 1 – Write green LED
111
       \% Bit 0 – Write red LED
      % Don't want to overwrite the encoder values
113
       readyMOSI = dec2bin(7,8);
114
       \% Receive the MISO byte bit by bit and update the vector holding the
      % byte
       switch bitNumber
           case 1
118
                   (MISO = true)
                i f
119
                    encoder 2_2 3_1 6(1) = 1;
120
                else
                    encoder 2_2 3_1 6(1) = 0;
                end
           case 2
124
                   (MISO = true)
                i f
                    encoder 2_2 3_1 6(2) = 1;
126
                else
128
                    encoder2_2_3_16(2) = 0;
```

```
129
                 end
            case 3
130
                  i f
                     (MISO = true)
                      encoder 2_2 3_1 6 (3) = 1;
                 else
133
                      encoder2_2_3_16(3) = 0;
134
                 end
             case 4
136
                     (MISO = true)
                 i f
                      encoder 2_2 3_1 6 (4) = 1;
138
                 else
                      encoder 2_2 3_1 6(4) = 0;
140
                 end
141
             case 5
142
                     (MISO = true)
                 i f
143
                      encoder 2_2 3_1 6(5) = 1;
144
                 else
145
                      encoder 2_2 3_1 6(5) = 0;
146
                 end
147
             case 6
148
                     (MISO = true)
                 i f
149
                      encoder 2_2 3_1 6(6) = 1;
                 else
                      encoder 2_2 3_1 6(6) = 0;
                 end
153
            case 7
154
                     (MISO = true)
                  i f
                      encoder 2_2 3_1 6(7) = 1;
156
                 else
                      encoder 2_2 3_1 6(7) = 0;
                 end
159
            case 8
                  i f
                     (MISO = true)
161
                      encoder2_2_3_16(8) = 1;
162
                 else
163
                      encoder 2_2 3_1 6(8) = 0;
164
                 end
165
        end
166
167
        case 3
168
        \% BYTE 3
       \% MOSI DATA – RED LED MSB
170
       % MISO DATA – ENCODER 2 (15-8)
        SS = 0;
        readyMOSI = dec2bin(redValue(1), 8);
       % Receive the MISO byte bit by bit and update the vector holding the
174
       % byte
175
        switch bitNumber
176
            case 1
177
                 if (MISO = true)
178
                      encoder 2_{15} (1) = 1;
179
                 else
180
                      encoder 2_{-}15_{-}8(1) = 0;
181
182
                 end
```

```
case 2
183
                   i f
                      (MISO = true)
184
                        encoder 2_{-1} 5_{-8} (2) = 1;
185
                   else
186
                        encoder 2_{15} (2) = 0;
187
188
                  end
             case 3
                   i f
                      (MISO = true)
190
                        encoder 2_{15_{8}}(3) = 1;
191
                   else
                        encoder 2_{-}15_{-}8(3) = 0;
193
                  end
194
             case 4
195
                      (MISO = true)
                   i f
196
                        encoder 2_{15_{8}}(4) = 1;
198
                   else
                        encoder 2_{1} 5_{8} (4) = 0;
199
                  end
200
             case 5
201
                   i f
                      (MISO = true)
202
                        encoder 2_{15} (5) = 1;
203
204
                   else
                        encoder 2_{1} 5_{8} (5) = 0;
205
                  end
206
             case 6
207
                      (MISO = true)
                   i f
208
                        encoder 2_1 5_8 (6) = 1;
209
                   else
210
                        encoder 2_{-}15_{-}8(6) = 0;
211
                  end
212
             case 7
213
                      (MISO = true)
                   i f
214
                        encoder 2_{15_{8}}(7) = 1;
215
                   else
216
                        encoder 2_{-}15_{-}8(7) = 0;
217
                  end
218
             case 8
219
                   i f
                      (MISO = true)
220
                        encoder 2_{1} 5_{8} (8) = 1;
221
                   else
222
                        encoder 2_{1} 5_{8} (8) = 0;
223
                  end
224
        end
225
226
        case 4
227
        \% BYTE 4
228
        % MOSI DATA – RED LED LSB
229
        % MISO DATA – ENCODER 2 (7-0)
230
        SS = 0;
231
        readyMOSI = dec2bin(redValue(2), 8);
232
        \% Receive the MISO byte bit by bit and update the vector holding the
233
        % byte
234
        switch bitNumber
236
             case 1
```

237	if (MISO == true)
238	$encoder2_7_0(1) = 1;$
239	else
240	$encoder2_7_0(1) = 0;$
241	end
242	case 2
243	if (MISO == true)
244	$encoder 2_7_0(2) = 1;$
245	else
246	$encoder2_7_0(2) = 0;$
247	end
248	$\begin{array}{c} \text{case } 3 \\ \text{if } (\text{MICO} + \text{true}) \end{array}$
249	$\prod_{i=1}^{n} (\text{MISO} = \text{true})$
250	$elicodel 2_7_0(3) = 1;$
251	$encoder 2 \ 7 \ 0 \ (3) = 0$
202	end $(12120(3) - 0)$
254	case 4
255	if (MISO = true)
256	$encoder2_7_0(4) = 1;$
257	else
258	$encoder2_7_0(4) = 0;$
259	end
260	case 5
261	if (MISO == true)
262	$encoder 2_7_0(5) = 1;$
263	else
264	$encoder2_7_0(5) = 0;$
265	end
266	case b $f_{\rm case}$ (MICO true)
267	(MISO = true)
200	else
209	encoder 2, 7, 0(6) = 0
271	end
272	case 7
273	if (MISO == true)
274	$encoder2_7_0(7) = 1;$
275	else
276	$encoder2_7_0(7) = 0;$
277	end
278	case 8
279	11 (MISO $=$ true)
280	$encoder2_7_0(8) = 1;$
281	else
282	$encoder2_{-}(0) = 0;$
283	end
204 285	Circ
286	case 5
287	% BYTE 5
288	% MOSI DATA – GREEN LED MSB
289	% MISO DATA – ENCODER 3 $(23-16)$
290	SS = 0;

```
readyMOSI = dec2bin(greenValue(1), 8);
291
        \% Receive the MISO byte bit by bit and update the vector holding the
292
        % byte
293
        switch bitNumber
294
            case 1
295
                 if (MISO = true)
296
                      encoder3_{2}3_{1}6(1) = 1;
297
                 else
298
                      encoder3_23_16(1) = 0;
299
                 end
300
            case 2
301
                 i f
                     (MISO = true)
302
                      encoder3_23_16(2) = 1;
303
                 else
304
                      encoder3_23_16(2) = 0;
305
306
                 end
            case 3
307
                 i f
                     (MISO = true)
308
                      encoder3_23_16(3) = 1;
309
                 else
310
                      encoder3_23_16(3) = 0;
311
                 end
            case 4
313
                 i f
                     (MISO = true)
314
                      encoder3_23_16(4) = 1;
315
                 else
                      encoder3_23_16(4) = 0;
317
                 end
318
             case 5
319
                     (MISO = true)
                 i f
                      encoder3_23_16(5) = 1;
321
                 else
                      encoder3_23_16(5) = 0;
323
                 end
324
             case 6
325
                 i f
                     (MISO = true)
326
                      encoder3_23_16(6) = 1;
327
                 else
328
                      encoder3_23_16(6) = 0;
329
                 end
330
             case 7
331
                     (MISO = true)
                 i f
332
                      encoder3_{2}3_{1}6(7) = 1;
                 else
334
                      encoder3_23_16(7) = 0;
                 end
336
            case 8
337
                     (MISO = true)
                 i f
338
                      encoder3_23_16(8) = 1;
                 else
340
                      encoder3_23_16(8) = 0;
341
                 end
342
        end
343
344
```

```
345
        case 6
        \% BYTE 6
346
        % MOSI DATA – GREEN LED LSB
347
        % MISO DATA – ENCODER 3 (15-8)
348
        SS = 0;
349
        readyMOSI = dec2bin(greenValue(2), 8);
350
        \% Receive the MISO byte bit by bit and update the vector holding the
351
        % byte
352
        switch bitNumber
353
             case 1
354
                  if (MISO = true)
355
                       encoder3_{1}5_{8}(1) = 1;
356
                  else
357
                       encoder3_{1}5_{8}(1) = 0;
358
                  end
359
             case 2
360
                  i f
                      (MISO = true)
361
                       encoder_{3}15_{8}(2) = 1;
362
                  else
363
                       encoder3_{1}5_{8}(2) = 0;
364
                  end
365
             case 3
366
                  i f
                      (MISO = true)
367
                       encoder_{3_15_8}(3) = 1;
368
                  else
369
                       encoder_{3_15_8}(3) = 0;
370
371
                  end
             case 4
372
                      (MISO = true)
                  i f
373
                       encoder3_{1}5_{8}(4) = 1;
374
                  else
375
                       encoder3_{15_{8}}(4) = 0;
                  end
377
             case 5
378
                  i f
                      (MISO = true)
379
                       encoder3_{1}5_{8}(5) = 1;
380
                  else
381
                       encoder3_{15_{8}}(5) = 0;
382
                  end
383
             case 6
384
                  i f
                      (MISO = true)
385
                       encoder3_{15_{8}}(6) = 1;
386
                  else
387
                       encoder3_{15_{8}}(6) = 0;
388
                  end
389
             case 7
390
                  i f
                      (MISO = true)
391
                       encoder3_{1}5_{8}(7) = 1;
392
                  else
393
                       encoder_{3_15_8}(7) = 0;
394
                  end
395
             case 8
396
                  i f
                      (MISO = true)
397
398
                       encoder3_{1}5_{8}(8) = 1;
```

```
399
                 else
                      encoder3_{1}5_{8}(8) = 0;
400
                 end
401
        end
402
403
404
        case 7
        % BYTE 7
405
       \% MOSI DATA – BLUE LED MSB
406
       % MISO DATA – ENCODER 3 (7-0)
407
        SS = 0;
408
        readyMOSI = dec2bin(blueValue(1), 8);
409
       % Receive the MISO byte bit by bit and update the vector holding the
410
       % byte
411
        switch bitNumber
412
            case 1
413
                     (MISO = true)
414
                  i f
                      encoder3_7_0(1) = 1;
415
                 else
416
                      encoder3_7_0(1) = 0;
417
                 end
418
             case 2
419
                     (MISO = true)
                 i f
420
                      encoder3_7_0(2) = 1;
421
                 else
422
                      encoder3_7_0(2) = 0;
423
                 end
424
425
             case 3
                 i f
                     (MISO = true)
426
                      encoder_{3_7_0}(3) = 1;
427
                 else
428
                      encoder3_7_0(3) = 0;
429
                 end
430
             case 4
431
                 i f
                     (MISO = true)
432
                      encoder_{3_7_0}(4) = 1;
433
                 else
434
                      encoder3_7_0(4) = 0;
435
                 end
436
             case 5
437
                 i f
                     (MISO = true)
438
                      encoder3_7_0(5) = 1;
439
                 else
440
                      encoder3_7_0(5) = 0;
441
                 end
442
             case 6
443
                  i f
                     (MISO = true)
444
                      encoder3_7_0(6) = 1;
445
446
                 else
                      encoder3_7_0(6) = 0;
447
                 end
448
            case 7
449
                  i f
                     (MISO = true)
450
                      encoder3_7_0(7) = 1;
451
452
                 else
```

```
encoder3_7_0(7) = 0;
453
                end
454
            case 8
455
                    (MISO = true)
                 i f
456
                     encoder_{3_7_0}(8) = 1;
457
                 else
458
                     encoder3_7_0(8) = 0;
459
                 end
460
       end
461
462
       case 8
463
       % BYTE 8
464
       % MOSI DATA – BLUE LED LSB
465
       % MISO DATA – TACHOMETER 2 (23-16)
466
       SS = 0;
467
       readyMOSI = dec2bin(blueValue(2), 8);
468
469
       case 9
470
       % BYTE 9
471
       % MOSI DATA – SET ENCODER 2 (23-16)
472
       % MISO DATA - TACHOMETER 2 (15-8)
473
       SS = 0;
474
       readyMOSI = dec2bin(0,8);
475
476
       case 10
477
       % BYTE 10
478
       % MOSI DATA – SET ENCODER 2 (15-8)
479
       % MISO DATA – TACHOMETER 2 (7-0)
480
       SS = 0;
481
       readyMOSI = dec2bin(0,8);
482
483
       case 11
484
       % BYTE 11
485
       % MOSI DATA – SET ENCODER 2 (7-0)
486
       % MISO DATA – TACHOMETER 3 (23-16)
487
       SS = 0;
488
       readyMOSI = dec2bin(0,8);
489
490
       case 12
491
       % BYTE 12
492
       % MOSI DATA – SET ENCODER 3 (23-16)
493
       % MISO DATA – TACHOMETER 3 (15-8)
494
       SS = 0;
495
       readyMOSI = dec2bin(0,8);
496
497
       case 13
498
       % BYTE 13
499
       % MOSI DATA – SET ENCODER 3 (15-8)
500
       % MISO DATA – TACHOMETER 3 (7-0)
501
       SS = 0;
502
       readyMOSI = dec2bin(0,8);
503
504
        case 14
505
       % BYTE 14
506
```

```
% MOSI DATA – SET ENCODER 3 (7-0)
507
       \% MISO DATA – RESERVED (0X00)
508
       SS = 0;
509
       readyMOSI = dec2bin(0,8);
510
512
       case 15
       % ** START OF CORE PACKET **
513
       % BYTE 15
514
       \% MOSI DATA – CORE MODE (0X01)
       % MISO DATA – CORE ID MSB
       SS = 0:
       readyMOSI = dec2bin(1,8);
518
519
       case 16
       % BYTE 16
       \% MOSI DATA – PADDING BYTE (0X00)
       % MISO DATA – CORE ID LSB
       SS = 0;
524
       readyMOSI = dec2bin(0,8);
       case 17
       % BYTE 17
528
529
       % MOSI DATA – CORE WRITE MASK
       % MISO DATA – CURRENT SENSE 0 (15-8)
530
       SS = 0;
       \% Do not enable the overwriting of encoders 0 and 1, but do enable the
       % overwriting of the motor voltages
       \% Bit 5 – Set encoder 1 enable
534
       \% Bit 4 - Set encoder 0 enable
       \% Bit 3 – Write motor 1 voltage
536
       \% Bit 2 – Write motor 1 enable
       \% Bit 1 – Write motor 0 voltage
       \% Bit 0 - Write motor 0 enable
       readyMOSI = dec2bin(15,8);
540
541
       case 18
542
       % BYTE 18
543
       % MOSI DATA – MOTOR 0 COMMAND (15-8)
544
       % MISO DATA – CURRENT SENSE 0 (7-0)
545
       SS = 0:
546
       \% Convert the desired voltage to a value between -999 and 999
547
       \% 24 is the saturation level in the model
548
       pitchVoltTemp = ceil((999*pitchVolt)/24);
549
       \% If the desired voltage is positive, concatenate a '1' to the front of
       \% the voltage value
       % Pass the MSB of the new value
       if (sign(pitchVoltTemp) == 1)
553
           temp = dec2bin(pitchVoltTemp, 15);
            for i = 1:8
                if (i = 1)
556
                    tempVoltBin(i) = '1';
                else
558
                    tempVoltBin(i) = temp(i - 1);
559
560
                end
```

```
561
           end
           readyMOSI = tempVoltBin;
562
       \% If the desired voltage is negative, find the 2's complement value of
563
       % the voltage
564
       elseif (sign(pitchVoltTemp) == -1)
565
           temp = dec2bin(-1*pitchVoltTemp, 15);
566
           % Find the complement
567
            for i = 1:15
568
                if (temp(i) = '1')
569
                     complement(i) = '0';
                else
                     complement(i) = '1';
                end
573
           end
574
           temp1 = bin2dec(complement) + 1;
           temp2 = dec2bin(temp1, 15);
576
            for i = 1:8
                if (i = 1)
578
                     tempVoltBin(i) = '1';
579
                else
580
                     tempVoltBin(i) = temp2(i - 1);
581
                end
582
           end
583
           readyMOSI = tempVoltBin;
584
       % If the desired voltage is zero, don't activate the motor
585
       else
           readyMOSI = dec2bin(0,8);
587
       end
588
589
       case 19
590
       % BYTE 19
       % MOSI DATA – MOTOR 0 COMMAND (7-0)
       % MISO DATA – CURRENT SENSE 1 (15-8)
       SS = 0:
594
       \% Convert the desired voltage to a value between -999 and 999
595
       \% 24 is the saturation level in the model
596
       pitchVoltTemp = ceil((999*pitchVolt)/24);
       % If the voltage is positive, pass the LSB
       if (sign(pitchVoltTemp) == 1)
599
           temp = dec2bin(pitchVoltTemp, 15);
600
           readyMOSI = temp(8:15);
601
       \% If the voltage is negative, find the 2's complement value and then
602
       % pass the LSB
603
       elseif (sign(pitchVoltTemp) == -1)
604
           temp = dec2bin(-1*pitchVoltTemp, 15);
605
           % Find the complement
606
            for i = 1:15
607
                if (temp(i) = '1')
608
                     complement(i) = '0';
609
                else
610
                     complement(i) = '1';
611
                end
612
           end
613
            temp1 = bin2dec(complement) + 1;
614
```

```
temp2 = dec2bin(temp1, 15);
615
            readyMOSI = temp2(8:15);
616
       \% If the desired voltage is zero, do not activate the motor
617
       else
618
            readyMOSI = dec2bin(0,8);
619
       end
620
       case 20
622
       % BYTE 20
623
       % MOSI DATA – MOTOR 1 COMMAND (15-8)
624
       % MISO DATA – CURRENT SENSE 1 (7-0)
625
       SS = 0;
626
        \% Convert the desired voltage to a value between -999 and 999
627
       \% 24 is the saturation level in the model
628
       yawVoltTemp = ceil((999*yawVolt)/24);
       \% If the desired voltage is positive, concatenate a '1' to the front of
630
       % the voltage value
631
       % Pass the MSB of the new value
632
       if (sign(yawVoltTemp) = 1)
633
            temp = dec2bin(yawVoltTemp, 15);
634
            for i = 1:8
635
                if (i = 1)
636
                     tempVoltBin(i) = '1';
637
                else
638
                     tempVoltBin(i) = temp(i - 1);
639
                end
            end
641
            readyMOSI = tempVoltBin;
642
       \% If the desired voltage is negative, find the 2's complement value of
643
       % the voltage
644
       elseif (sign(yawVoltTemp) = -1)
645
            temp = dec2bin(-1*yawVoltTemp, 15);
646
           % Find the complement
647
            for i = 1:15
648
                if (temp(i) = '1')
649
                     complement(i) = '0';
650
                else
                     complement(i) = '1';
652
                end
653
            end
654
            temp1 = bin2dec(complement) + 1;
655
            temp2 = dec2bin(temp1, 15);
656
            for i = 1:8
657
                if (i = 1)
658
                     tempVoltBin(i) = '1';
659
                else
660
                     tempVoltBin(i) = temp2(i - 1);
661
                end
662
            end
663
            readyMOSI = tempVoltBin;
664
       % If the desired voltage is zero, don't activate the motor
665
       else
666
            readyMOSI = dec2bin(0,8);
667
668
       end
```

```
669
       case 21
670
       % BYTE 21
671
       % MOSI DATA – MOTOR 1 COMMAND (7-0)
672
       % MISO DATA – TACHOMETER 0 (23-16)
673
       SS = 0;
674
       \% Convert the desired voltage to a value between -999 and 999
675
       \% 24 is the saturation level in the model
676
       yawVoltTemp = ceil((999*yawVolt)/24);
677
       % If the voltage is positive, pass the LSB
678
       if (sign(yawVoltTemp) = 1)
679
            temp = dec2bin(yawVoltTemp, 15);
680
            readyMOSI = temp(8:15);
       \% If the voltage is negative, find the 2's complement value and then
682
       % pass the LSB
683
       elseif (sign(yawVoltTemp) = -1)
684
            temp = dec2bin(-1*yawVoltTemp, 15);
685
            for i = 1:15
686
                if (temp(i) = '1')
687
                     complement(i) = '0';
688
                else
                     complement(i) = '1';
690
                end
691
            end
692
            temp1 = bin2dec(complement) + 1;
693
            temp2 = dec2bin(temp1, 15);
694
            readyMOSI = temp2(8:15);
695
       \% If the desired voltage is zero, do not activate the motor
696
       else
697
            readyMOSI = dec2bin(0,8);
698
       end
699
700
       case 22
701
       % BYTE 22
702
       % MOSI DATA – SET ENCODER 0 (23-16)
703
       % MISO DATA – TACHOMETER 0 (15-8)
704
       SS = 0;
705
       readyMOSI = dec2bin(0,8);
706
707
       case 23
708
       % BYTE 23
709
       % MOSI DATA – SET ENCODER (7-0)
710
       % MISO DATA – TACHOMETER 0 (7-0)
711
       SS = 0;
       readyMOSI = dec2bin(0,8);
713
714
       case 24
715
       % BYTE 24
716
       % MOSI DATA – SET ENCODER 0 (7-0)
       % MISO DATA – TACHOMETER 1 (23-16)
718
       SS = 0;
719
       readyMOSI = dec2bin(0,8);
720
721
722
       case 25
```

```
% BYTE 25
723
       \% MOSI DATA – SET ENCODER 1 (23–16)
724
       % MISO DATA – TACHOMETER 1 (15-8)
725
       SS = 0;
726
       readyMOSI = dec2bin(0,8);
727
728
       case 26
       % BYTE 26
730
       % MOSI DATA – SET ENCODER 1 (15-8)
731
       % MISO DATA – TACHOMETER 1 (7-0)
       SS = 0:
       readyMOSI = dec2bin(0,8);
734
735
       case 27
736
       % BYTE 27
       % MOSI DATA – SET ENCODER 1 (7-0)
738
       % MISO DATA – STATUS
739
       SS = 0;
740
       readyMOSI = dec2bin(0,8);
741
742
       case 28
743
       % BYTE 28
744
       \% MOSI DATA – PADDING BYTE (0X00)
745
       % MISO DATA – ENCODER 0 (23-16)
746
       SS = 0;
747
       readyMOSI = dec2bin(0,8);
748
749
       case 29
750
       % BYTE 29
751
       \% MOSI DATA – PADDING BYTE (0X00)
752
       % MISO DATA – ENCODER 0 (15-8)
753
       SS = 0;
754
       readyMOSI = dec2bin(0,8);
755
756
       case 30
757
       % BYTE 30
758
       \% MOSI DATA – PADDING BYTE (0X00)
759
       % MISO DATA – ENCODER 0 (7-0)
760
       SS = 0:
761
       readyMOSI = dec2bin(0,8);
762
763
       case 31
764
       % BYTE 31
765
       \% MOSI DATA – PADDING BYTE (0X00)
766
       % MISO DATA – ENCODER 1 (23-16)
767
       SS = 0;
768
       readyMOSI = dec2bin(0,8);
769
770
       case 32
771
       % BYTE 32
772
       \% MOSI DATA – PADDING BYTE (0X00)
773
       % MISO DATA – ENCODER 1 (15-8)
774
       SS = 0;
775
       readyMOSI = dec2bin(0,8);
776
```

777

```
case 33
778
       % BYTE 33
779
       \% MOSI DATA – PADDING BYTE (0X00)
780
       % MISO DATA – ENCODER 1 (7-0)
781
       SS = 0;
782
       readyMOSI = dec2bin(0,8);
783
784
       case 34
785
       % BYTE 34
786
       \% MOSI DATA – PADDING BYTE (0X00)
787
       % MISO DATA – ACCELEROMETER X (15-8)
788
       SS = 0;
789
       readyMOSI = dec2bin(0,8);
790
       case 35
792
       % BYTE 35
793
       \% MOSI DATA – PADDING BYTE (0X00)
794
       % MISO DATA – ACCELEROMETER X (7-0)
795
       SS = 0;
796
       readyMOSI = dec2bin(0,8);
798
       case 36
799
       % BYTE 36
800
       \% MOSI DATA – PADDING BYTE (0X00)
801
       % MISO DATA – ACCELEROMETER Y (15-8)
802
       SS = 0;
803
       readyMOSI = dec2bin(0,8);
804
805
       case 37
806
       % BYTE 37
807
       \% MOSI DATA – PADDING BYTE (0X00)
808
       % MISO DATA – ACCELEROMETER Y (7-0)
809
       SS = 0;
810
       readyMOSI = dec2bin(0,8);
811
812
       case 38
813
       % BYTE 38
814
       \% MOSI DATA – PADDING BYTE (0X00)
815
       % MISO DATA – ACCELEROMETER Z (15-8)
816
       SS = 0;
817
       readyMOSI = dec2bin(0,8);
818
819
       case 39
820
       % BYTE 39
821
       \% MOSI DATA – PADDING BYTE (0X00)
822
       % MISO DATA – ACCELEROMETER Z (7-0)
823
       SS = 0;
824
       readyMOSI = dec2bin(0,8);
825
826
       case 40
827
       % BYTE 40
828
       \% MOSI DATA – PADDING BYTE (0X00)
829
       % MISO DATA – GYROSCOPE X (15-8)
830
```

```
SS = 0;
831
       readyMOSI = dec2bin(0,8);
832
833
       case 41
834
       % BYTE 41
835
       \% MOSI DATA – PADDING BYTE (0X00)
836
       % MISO DATA – GYROSCOPE X (7-0)
837
       SS = 0;
838
       readyMOSI = dec2bin(0,8);
839
840
       case 42
841
       % BYTE 42
842
       \% MOSI DATA – PADDING BYTE (0X00)
843
       % MISO DATA – GYROSCOPE Y (15-8)
844
       SS = 0;
845
       readyMOSI = dec2bin(0,8);
846
847
       case 43
848
       % BYTE 43
849
       \% MOSI DATA – PADDING BYTE (0X00)
850
       % MISO DATA – GYROSCOPE Y (7-0)
851
       SS = 0;
852
       readyMOSI = dec2bin(0,8);
853
854
       case 44
855
       % BYTE 44
856
       \% MOSI DATA – PADDING BYTE (0X00)
857
       % MISO DATA – GYROSCOPE Z (15-8)
858
       SS = 0;
859
       readyMOSI = dec2bin(0,8);
860
861
       case 45
862
       % BYTE 45
863
       \% MOSI DATA – PADDING BYTE (0X00)
864
       % MISO DATA – GYROSCOPE Z (7-0)
865
       SS = 0;
866
       readyMOSI = dec2bin(0,8);
867
868
       case 46
869
       % BYTE 46
870
       \% MOSI DATA – PADDING BYTE (0X00)
871
       \% MISO DATA – RESERVED (0X00)
872
       SS = 0;
873
       readyMOSI = dec2bin(0,8);
874
875
       case 47
876
       % BYTE 47
877
       \% MOSI DATA – PADDING BYTE (0X00)
878
       \% MISO DATA – RESERVED (0X00)
879
       SS = 0;
880
       readyMOSI = dec2bin(0,8);
881
882
       case 48
883
       % BYTE 48
884
```

```
\% MOSI DATA – PADDING BYTE (0X00)
885
       \% MISO DATA – RESERVED (0X00)
886
       SS = 0;
887
       readyMOSI = dec2bin(0,8);
888
889
        case 49
890
       % BYTE 49
891
       \% MOSI DATA – PADDING BYTE (0X00)
892
       \% MISO DATA – RESERVED (0X00)
893
       SS = 0;
894
       readyMOSI = dec2bin(0,8);
895
896
        case 50
897
       % BYTE 50
898
       \% MOSI DATA – PADDING BYTE (0X00)
899
       \% MISO DATA – RESERVED (0X00)
900
       readyMOSI = dec2bin(0,8);
901
       \% Reset the the slave select to begin the process again
902
       % Maybe not needed
903
       SS = 1;
904
905
        otherwise
906
       % Just in case
907
       readyMOSI = dec2bin(0,8);
908
       SS = 0;
909
910 end
911 % Pass the byte we have derived bit by bit
912 % Use the character array to determine the bit value
  switch bitNumber
913
        case 1
914
            if (readyMOSI(1) = '1')
915
                MOSI = 1;
916
            else
917
                 MOSI = 0;
918
            end
919
        case 2
920
                (readyMOSI(2) = '1')
            i f
921
                MOSI = 1;
922
            else
923
                MOSI = 0;
924
            end
925
        case 3
926
            if (readyMOSI(3) = '1')
927
                MOSI = 1;
928
            else
929
                 MOSI = 0;
930
            end
931
        case 4
932
                (readyMOSI(4) = '1')
            i f
933
                 MOSI = 1;
934
            else
935
                 MOSI = 0;
936
            end
937
938
        case 5
```

```
if (readyMOSI(5) = '1')
939
                MOSI = 1;
940
            else
941
                MOSI = 0;
942
            end
943
        case 6
944
            i f
               (readyMOSI(6) = '1')
945
                MOSI = 1;
946
            else
947
                MOSI = 0;
948
            end
949
950
        case 7
            if (readyMOSI(7) = '1')
951
                MOSI = 1;
952
            else
953
                 MOSI = 0;
954
            end
955
        case 8
956
            i f
                (readyMOSI(8) = '1')
957
                 MOSI = 1;
958
            else
959
                 MOSI = 0;
960
            end
961
            \% If we have just sent the last bit, update the byte number
962
            byteNumber = byteNumber + 1;
963
            % Reset back to zero if we have finished one transfer
964
            if (byteNumber = 51)
965
                 byteNumber = 0;
966
            end
967
        otherwise
968
            MOSI = 0;
969
970 end
971 % Update the bit number that we are sending
_{972} bitNumber = bitNumber + 1;
973 % Reset back to 1 if we have finished one byte
_{974} if (bitNumber = 9)
        bitNumber = 1;
975
976 end
977 end
```

Appendix E

Android Application

E.1 Android Application (Tutorial)

- 1. Verify that all of the proper software is installed as mentioned in Section 6.4.
- 2. Repeat the steps used in Appendix D.1 to push the Simulink model onto the Raspberry Pi. The Simulink model is specific for Android communication.
- 3. Open the Simulink model for the Android smart phone application.
- 4. Connect your Android smart phone to your laptop using a USB cable. Mine was provided with the phone. See Figure E.1.
- 5. Deploy the model to the hardware. This is the same as that in Appendix D.1, but this time the hardware is your smart phone.
- 6. You need to have your phone connected to internet in order for the build to complete.
- 7. Once the build is complete, you should see an application with the name of your model. See Figure E.2.
- 8. Connect your smart phone to the local network you set up with static IPs. This was discussed in Section 6.4.
- 9. Open the application and the model should begin to run. See Figure E.3.
- 10. Execute the code on the Raspberry Pi for its corresponding model. This is the same as in Appendix D.1.
- 11. If both models are setup correctly and the IP addresses are consistent with the models, there should be communication working between the two devices.



Figure E.1: Connect your Android smart phone to your laptop.



Figure E.2: Android smart phone application.


Figure E.3: Android smart phone application.

Appendix F Quanser AERO Data Sheets

F.1 Quanser AERO Product Specifications

QUANSER INNOVATE · EDUCATE

QUANSER AERO

Dual-rotor aerospace experiment with reconfigurable dynamic components for mechatronics exploration and controls.

FLEXIBLE PLATFORM FOR MECHATRONICS AND CONTROLS

The Quanser AERO is a fully integrated lab experiment, designed for teaching mechatronics and control concepts at the undergraduate level, as well as for advanced aerospace research applications.

The experiment is reconfigurable for various aerospace systems, from 1 DOF attitude control and 2 DOF helicopter to half-quadrotor. Integrating Quanser-developed QFLEX 2 computing interface technology, the Quanser AERO also offers flexibility in lab configurations, using a PC, or microcontrollers, such as NI myRIO, Arduino and Raspberry Pi. With the comprehensive course materials included, you can build a state-ofthe-art teaching lab for your mechatronics or control courses, engage students in various design and capstone projects, and validate your research concepts on a high-quality, robust, and precise platform.

HOW IT WORKS

The Quanser AERO consists of two propellers, powered by DC motors. Combined with the light-weight design of the experiment, this makes the system capable of highly responsive movements. The Quanser AERO's compact base includes a built-in amplifier with an integrated current sensor, built-in data acquisition device, and an interchangeable QFLEX 2 interface panel. The experiment comes with additional propellers to illustrate the efficiency of different propeller designs and effects of cross-coupling.

The propeller motors are equipped with optical encoders. The motor current and voltage sensors can be used to monitor the power consumption of the experiment.

The slip ring mechanism allows for continuous 360° yaw rotation. The angles of the pitch and yaw axes are measured using high-resolution optical encoders. The pitch and yaw axis can be independently locked, and the angle of the propeller assemblies can be adjusted between horizontal and vertical positions. This allows users to reconfigure the Quanser AERO for various aerospace systems (1 DOF attitude control, 2 DOF helicopter, half-quadrotor) and experiments (e.g. pitch-only system modeling).

The Inertial Measurement Unit (IMU) board includes accelerometer and gyroscope sensors, which can be used for attitude and yaw estimation and verification against the direct position measurements from the encoders.

The Quanser AERO also has a user-controllable tri-color LED strip. It can be programmed to indicate state, power, or other control performance characteristics of the Quanser AERO.



See system specifications on reverse.

QUANSER AERO SOLUTION COMPONENTS

- Quanser AERO with QFLEX 2 interface panel of your choice (USB or Embedded)
- Optional: Additional QFLEX 2 interface panel
- Quanser Control Software (required for Quanser AERO USB experiment): QUARC for MATLAB/Simulink or QRCP for LabVIEW*
- Complete dynamic model and pre-build controllers
- ABET¹-aligned, flexible digital media courseware (for Quanser AERO USB experiment)
- Arduino examples and interfacing datasheet (for Quanser AERO Embedded experiment)

*MATLAB/Simulink and LabVIEW licenses not included

QUANSER AERO INTERFACE OPTIONS

The Quanser AERO is available with two different, easily interchangeable interface panels:

- Quanser AERO USB experiment (with QFLEX 2 USB panel) interfaces to Quanser's control software running on your lab's PC via a standard USB 2.0 connection. The Quanser AERO USB can be used with MATLAB[®]/ Simulink[®] and Quanser QUARC software, or with LabVIEW[™] using the Quanser RCP software. With the USB version of the experiment, you can take full advantage of the comprehensive course materials and lab experiments for your controls-based courses and projects.
- Quanser AERO Embedded experiment (with QFLEX 2 Embedded panel) interfaces to your microcontroller (not included with the experiment) via SPI connection. The Quanser AERO Embedded does not require any additional software. This option is ideal to expose students to various microcontroller techniques, as well as for final (capstone) projects in mechatronics, control, or other similar programs.

Note: The Quanser AERO experiment includes one interface panel of your choice. Additional interface panel(s) can be purchased separately.

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SYSTEM SPECIFICATIONS Quanser AERO



FEATURES

- Compact and integrated system
- High-efficiency coreless DC motors
- High resolution optical encoders
- Pitch & yaw axes and DC motors/rotors speed measurements through digital tachometer
- Built-in voltage amplifier with integrated current sensor
- Integrated data acquisition(DAQ) device
- User-controllable tri-color LED
- Easy-connect cables and connectors

COURSEWARE TOPICS COVERED

The Quanser AERO USB courseware includes ABET¹-aligned Instructor and Student Workbooks with complete lab exercises, covering topics:

- Hardware integration
- Single propeller speed control

1 DOF attitude control configuration:

- PID control
- Introduction to IMU
- Modeling using transfer function
- System identification
- Gain scheduling

DEVICE SPECIFICATION

- Flexible QFLEX 2 computing interface for USB and SPI connections
- Open architecture design, allowing users to design their own controller
- Fully compatible with MATLAB[®]/Simulink[®] and LabVIEW[™]
- Fully documented system models and parameters provided for MATLAB[®]/Simulink[®] and LabVIEW[™]
- ABET-aligned, modular, digital media courseware provided for the Quanser AERO USB
- Microcontroller examples and interfacing datasheet provided for the Quanser AERO Embedded
- Additional community-created resources available on www.QuanserShare.com

Laboratory Guides with modeling and control design examples:

- 2 DOF helicopter configuration
- Modeling
- Linear state-space representation
- State-feedback control
- Coupled dynamics

Half-quadrotor configuration

- Modeling
- Simple yaw control
- Kalman filter

Base dimensions (W x H x D)	17.8 cm x 17.8 cm x 7 cm
Device height (with propeller in horizontal position)	35.6 cm
Device length	51 cm
Device mass	3.6 kg
Propeller diameter	12.7 cm
Yaw angle range	360°
Elevation angle range	124° (± 62° from horizontal) half-quadrotor configuration
Pitch encoder resolution (in quadrature)	512 counts/revolution
Yaw/travel encoder resolution (in quadrature)	1024 counts/revolution
Motor current torque constant	57.7 mN.m/A
Tri-axis gyroscope range	± 245 dps
Tri-axis accelerometer range	± 2g
Interfaces available: QFLEX 2 USB	USB 2.0
QFLEX 2 Embedded	SPI

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Quanser is the world leader in education and research for real-time control design and implementation. We specialize in outfitting engineering control laboratories to help universities captivate the brightest minds, motivate them to success and produce graduates with industry-relevant skills. Universities worldwide implement Quanser's open architecture control solutions, industry-relevant curriculum and cutting-edge work stations to teach Introductory, Intermediate or Advanced controls to students in Electrical, Mechanical, Mechatronics, Robotics, Aerospace, Civil, and various other engineering disciplines.

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F.2 Quanser AERO User Manual



USER MANUAL

Quanser AERO Experiment

Set Up and Configuration



CAPTIVATE. MOTIVATE. GRADUATE.

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1 PRESENTATION

The Quanser Quanser Aero, pictured in Figure 1.1, is a compact dual-rotor two degree-of-freedom aerospace system that can be used to perform a variety of actuator and flight control based experiments. The Quanser Aero can be configured with either the QFLEX 2 USB or QFLEX 2 Embedded interface modules. The QFLEX 2 USB allows control by a computer via USB connection. The QFLEX 2 Embedded allows for control by a microcontroller device such as an Arduino via a 4-wire SPI interface.

For all versions, the system is driven using two direct-drive 18V brushed DC motors. The motors are powered by a built-in PWM amplifier with built-in current sense. Single-ended rotary encoders are used to measure the angular position of the DC motors, and the speed of the motors can be measured with a software tachometer.

Main Quanser Aero features:

- · Compact and complete aerospace control system
- 18V direct-drive brushed DC motors
- · Encoders mounted on DC motors and support yolk
- DC motor tachometer output
- · Built-in PWM amplifier with integrated current sense
- Built-in data acquisition (DAQ) device
- · Low and High-efficiency propellers
- · Lockable pitch and yaw axes
- Tri-color LED indicator lights



Figure 1.1: Quanser Aero

Caution: This equipment is designed to be used for educational and research purposes and is not intended for use by the general public. The user is responsible to ensure that the equipment will be used by technically qualified personnel only.

2 SYSTEM HARDWARE

2.1 System Schematic

The Quanser Aero can be configured with one of two different I/O interfaces: the QFLEX 2 USB, and the QFLEX 2 Embedded. The QFLEX 2 USB provides a USB interface for use with a computer. The QFLEX 2 Embedded provides a 4-wire SPI interface for use with an external microcontroller board.

The interaction between the different system components on the Quanser Aero is illustrated in Figure 2.1. On the data acquisition (DAQ) device block, the motor position encoders are connected to Encoder Input (EI) channels #0 and #1. El2 reads the pitch angle of the Aero body, and El3 reads the yaw angle of the yoke. The Analog Output (AO) channels are connected to the power amplifier command, which then drives the DC fan motors. The DAQ Analog Input (AI) channels are connected to the PWM amplifier current sense circuitry. The DAQ also controls the integrated tri-colour LEDs via an internal serial data bus. The DAQ can be interfaced to the PC or laptop via USB link in the QFLEX 2 USB, or to an external microcontroller via SPI in the QFLEX 2 Embedded.



Figure 2.1: Interaction between Quanser Aero components.

The schematic given in Figure **??** illustrates the main Quanser Aero components and how they interact with a data acquisition (DAQ) device.

2.2 Hardware Components

The main Quanser Aero components - for the USB and SPI embedded interfaces - are listed in Table 2.1. The components on the QFLEX 2 USB are labeled in Figure 2.2c, the components on the QFLEX 2 Embedded are shown in Figure 2.2d.

ESD Warning: Quanser Aero internal components are sensitive to electrostatic discharge. Before handling the Quanser Aero ensure that you have been properly grounded.



ID	Component	ID	Component
1	Aero base	11	Propeller guard screw
2	Yaw pivot	12	Propeller attachment hub
3	Support yolk	13	IMU (Internal to aero body)
4	Aero body	14	System Power LED
5	Pitch pivot	15	Interface Power LED
6	Pitch lock screws	16	SPI Data Connector*
7	Status LED strip	17	USB connector [†]
8	Thruster rotation locks	18	Quanser Aero internal data bus
9	Thruster 0	19	Quanser Aero DAQ/amplifier board
10	Thruster 1	20	Power Connector

Table 2.1: Quanser Aero Components

† only on QFLEX 2 USB *only on QFLEX 2 Embedded



(a) Quanser Aero Top View



(b) Quanser Aero Interior





(d) Quanser Aero with QFLEX 2 Embedded

(c) Quanser Aero with QFLEX 2 USB (Figure 2.2: Quanser Aero components

2.2.1 DC Motor

The Quanser Aero includes two direct-drive 18V brushed DC motors. The motor specifications are given in Table 2.2.

The Quanser Aero incorporates the Allied Motion CL40 Series Coreless DC Motor model 16705. The complete specification sheet of the motor is included at: http://alliedmotion.com/Products/Series.aspx?s=29.





Caution: Holding the motor in a stalled position for a prolonged period of time at applied voltages of over 5V can result in permanent damage.

2.2.2 Thruster Assemblies

The Quanser Aero has two identical thruster assemblies which are attached to the Aero body. Thruster 0 can be identified by locating the pitch lock screws or the yaw lock magnets. Both of these items are on the side of the yolk facing thruster 0.

2.2.3 High-efficiency Propeller

The Quanser Aero ships with two counter-rotating APC 5.0x4.6 propellers, models LP05046E/EP. More information on the propellers can be found on the Advanced Precision Composites website (www.apcprop.com).

2.2.4 Low-efficiency Propeller

The Quanser Aero ships with two eight-vane counter-rotating 3D-printed propellers.

2.2.5 Propeller Hub

The Quanser Aero propellers are connected to the DC motors with aluminum prop adapters with collets. The propeller adaptors are E-flite part number EFLM1922.

2.2.6 Pitch and Motor Position Encoders

The encoders used to measure the pitch of the Aero body and the angular position of the DC motorson the Quanser Aero is a single-ended optical shaft encoder. It outputs 2048 counts per revolution in quadrature mode (512 lines per revolution).

The encoders used to measure the pitch of the Aero body, and angular position of the DC motors on the Quanser Aero is the US Digital E8P-512-118 single-ended optical shaft encoder. The complete specification sheet of the E8P optical shaft encoder is given in E8P Data Sheet.

2.2.7 Yaw Encoder

The encoders used to measure the yaw of the support yolk on the Quanser Aero is an optical encoder. It outputs 4096 counts per revolution in quadrature mode (1024 lines per revolution).

The encoders used to measure the yaw of the support yolk on the Quanser Aero is the US Digital E3-1024-984 optical encoder. The complete specification sheet of the E3 optical shaft encoder is given in E3 Data Sheet.

2.2.8 Inertial Measurement Unit

The Quanser Aero includes an integrated IMU mounted within the Aero body. This module allows for real-time measurement of the angular position and velocity along all three of the primary axes of the Aero body.

The IMU incorporated into the Quanser Aero is the STMicroelectronics LSM6DS0 iNEMO intertial module. Further information on the module can be found in the LSM6DS0 data sheet.

2.2.9 Data Acquisition (DAQ) Device

The Quanser Aero includes an integrated data acquisition device with four 16-bit encoder channels with quadrature decoding and two PWM analog output channels. The DAQ also incorporates a 12-bit ADC which provides current sense feedback for the motors. The current feedback is used to detect motor stalls and will disable the amplifier if a prolonged stall is detected.

2.2.10 Power Amplifier

The Quanser Aero circuit board includes a PWM voltage-controlled power amplifier capable to providing 2 A peak current and 0.5 A continuous current (based on the thermal current rating of the motor). The output voltage range to the load is between \pm 24 V.

2.2.11 Power Supply

The Quanser Aero is equipped with an external DC power supply which provides power for the sensors and motors. This supply is intended for use with 100-240 VAC at 50-60 Hz.

Only the provided power supply and AC cord should be used with the Quanser Aero. The included supply is an Adapter Technology Co Ltd model ATS065-P241.

2.2.12 Embedded System Connector

The SPI data connector pictured on the QFLEX 2 Embedded in Figure 2.2d allows an external microcontroller to set motor voltage and LED brightnesses, read and set encoder counters, and read motor speed and current flow. See the QFLEX 2 Embedded data sheet for information on connecting the SPI interface.

2.3 Environmental

The Quanser Aero is designed to function under the following environmental conditions:

- Standard rating
- Indoor use only
- Temperature 5°C to 40°C
- Altitude up to 2000 m
- Maximum relative humidity of 80% up to 31°C decreasing linearly to 50% relative humidity at 40°C
- Pollution Degree 2
- Mains supply voltage fluctuations up to \pm 10% of nominal voltage
- Maximum transient overvoltage 2500 V

• Marked degree of protection to IEC 60529: Ordinary Equipment (IPX0)

2.4 System Parameters

Table 2.2 lists and characterizes the main parameters associated with the Quanser Aero.

Symbol	Description	Value
DC Moto	r	
Vnom	Nominal input voltage	18.0 V
$ au_{nom}$	Nominal torque	22.0 mN-m
$\omega_{\sf nom}$	Nominal speed	3050 RPM
Inom	Nominal current	0.540 A
R_m	Terminal resistance	8.4 Ω
k_t	Torque constant	0.042 N-m/A
k_m	Motor back-emf constant	0.042 V/(rad/s)
J_m	Rotor inertia	$4.0 imes 10^{-6} \text{ kg-m}^2$
L_m	Rotor inductance	1.16 mH
Aero Boo	dy	
M_b	Mass of body	1.075 kg
D_m	Center of mass	-7.59 mm
J_p	Pitch inertia	$2.15 imes 10^{-2} \text{ kg-m}^2$
J_y	Yaw inertia	$2.37 imes 10^{-2} \text{ kg-m}^2$
D_t	Thrust displacement	15.8 cm
Motor an	d Pitch Encoders	
	Encoder line count	512 lines/rev
	Encoder line count in quadrature	2048 lines/rev
	Encoder resolution (in quadrature)	0.176 deg/count
Yaw Enc	oder	
	Encoder line count	1024 lines/rev
	Encoder line count in quadrature	4096 lines/rev
	Encoder resolution (in quadrature)	0.088 deg/count
Amplifie	<u></u>	
	Amplifier type	PWM
	Peak Current	2 A
	Continuous Current	0.5 A
	Output voltage range (recommended)	±18 V
	Output voltage range (maximum)	±24 V

Table 2.2: Quanser Aero System Parameters



3 SYSTEM SETUP



Caution: If the equipment is used in a manner not specified by the manufacturer, the protection provided by the equipment may be impaired.

3.1 Components

To setup the Quanser Aero system, you need the following components:

- 1. Quanser Aero (USB or Embedded version)
- 2. High and low-efficiency propellers
- 3. Propeller adapters
- 4. Power supply with the following ratings:
 - Input Rating: 100-240 V AC, 50-60 Hz, 1.4 A
 - Output Rating: 24 V DC, 2.71 A

Note: Only the power supply provided (AC-DC adapter by Adapter Technology Co Ltd, model ATS065-P241) should be used with the Quanser Aero

5. Power cable

Note: Only the power cable provided should be used with the Quanser Aero

Note: Make sure that the power cable is accessible for disconnection in case of emergency.

Caution: Precaution must be taken during the connection of this equipment to the AC outlet to make sure the grounding (earthing) is in place, and that the ground wire is not disconnected

6. USB 2.0 A/B cable (for QFLEX 2 USB) or jumper wires (for QFLEX 2 Embedded)

3.2 QFLEX 2 USB Hardware Setup

To setup the QFLEX 2 USB follow these steps:

- 1. The Quanser Aero should have one of the included sets of propellers installed. If the other propellers are required, follow the procedure for exchanging propellers in Section **?? before** connecting power.
- Connect USB 2.0 cable from back cover of Quanser Aero to an enabled USB 2.0 port on your desktop PC or laptop.
- 3. Connect the **Power** connector on the Quanser Aero to the power supply. Ensure the power supply is connected to a wall outlet using the appropriate power cable.
- 4. The QFLEX 2 USB driver should install automatically. If not, then you may not have installed all the required software to support the device including either QUARC[®] or Quanser Rapid Control Prototyping Toolkit[®].

3.3 **QFLEX 2 Embedded Hardware Setup**

This section describes how to connect the QFLEX 2 Embedded to an external microcontroller board. The connection procedure is given below, and summarized in Table 3.1. The wires required to connect the QFLEX 2 Embeddedare not included with the unit, connections may be made with jumper wires or a custom wiring solution dependent on the external controller being used.

Follow these steps to connect the QFLEX 2 Embedded to your microcontroller device:

- 1. Before proceeding make sure your microcontroller device has been setup and successfully tested. Refer to the documentation supplied with your control system for set up and testing instructions.
- 2. Make sure the everything is powered off before making any of these connections. This includes turning off the external microcontroller board.
- 3. Connect the GND pin on the QFLEX 2 Embedded to a digital ground connection on the microcontroller board.
- 4. Connect the MOSI, MISO, and CLK pins on the QFLEX 2 Embedded to the microcontroller board as outlined in the SPI interface documentation for your controller.
- 5. Connect the CS pin on the QFLEX 2 Embedded to a digital output on the microcontroller board.
- 6. Connect the 1.8V-5V pin on the QFLEX 2 Embedded to a signal level power pin on the microcontroller board in the 1.8V to 5V range.



Caution: Applying voltages in excess of 5V to the 1.8V-5V input on the QFLEX 2 Embedded may result in damage to the QFLEX 2 Embedded.

Cable #	From Microcontroller	To QFLEX 2 Embedded	Signal
1	VCC/VDD(1.8V-5V)	1.8V-5V	QFLEX 2 Embedded interface power.
2	MOSI/SDO/SO	MOSI	SPI master out, slave in data line.
3	MISO/SDI/SI	MISO	SPI master in, slave out data line.
4	SCLK/SCK	CLK	SPI clock line.
5	Digital output line	CS	SPI slave select line.
6	GND/DGND	GND	SPI digital signal ground.

Table 3.1: QFLEX 2 Embedded wiring summary

3.4 Exchanging **QFLEX 2** Panels

Follow these steps to install the QFLEX 2 USB or QFLEX 2 Embedded panel in your Quanser Aero.

- 1. Disconnect the 24VDC power input from the Quanser Aero.
- 2. Disconnect any connections between the currently installed QFLEX panel and the computer or microcontroller board.



ESD Warning: The interior of the Quanser Aero contains components which are sensitive to electrostatic discharge. Before opening the Quanser Aero case, ensure that both you and the workspace are properly grounded.

3. Remove the four screws at the corners of the QFLEX panel to release the panel from the Aero chassis. The Quanser Aero is shown in Figure 3.1 below with the screws removed.





Figure 3.1: Quanser Aero with QFLEX panel detached.

- 4. Disconnect the Aero internal data cable from the QFLEX panel by depressing the latching tab.
- 5. Connect the Aero internal data cable to the QFLEX panel to be installed, pressing the connector into the socket until a click is heard and the connector latches in place.
- 6. Anchor the QFLEX panel in place using the four screws removed earlier.

 Δ Caution: Ensure that the Quanser Aero is completely reassembled, with all screws in place before connecting power or attempting operation.

3.5 Changing Thruster Orientation

Follow these steps to change the orientation of the thrusters on the Quanser Aero.

1. Use the included hex key to loosen the thruster rotation lock set-screw on the thruster you wish to rotate by one quarter turn as shown in Figure 3.2a

Caution: Do not loosen the set screws more than one half turn to prevent accidentally detaching the thruster assembly.

2. Rotate the propeller to the desired angle as shown in Figure 3.2b

Note: Each thruster has a 90 degree range motion and will only rotate in one direction from either the vertical or horizontal positions.

3. Tighten the thruster rotation lock set-screw.

3.6 Exchanging Propellers

Follow these steps to change the propellers in the Quanser Aero.



- 1. Disconnect the 24VDC power input from the Quanser Aero.
- 2. Disconnect any connections between the currently installed QFLEX panel and the computer or microcontroller board.
- 3. Unfasten the propeller guard screws and remove the propeller guard.



Figure 3.2: Quanser Aero propeller change steps

- 4. Insert a small hex key or similar object (not provided) through the cap of the propeller hub as shown in Figure 3.3a.
- 5. While holding the propeller still, loosen the collet in the propeller hub slightly by turning the cap counterclockwise.
- 6. Pull gently on the propeller hub to slide the assembly off the motor shaft.
- 7. Disassemble the propeller hub as shown in Figure 3.3b



Figure 3.3: Quanser Aero propeller change steps

- 8. Identify the correct propeller from the counter-rotating pairs for the thruster being swapped. Under positive voltage, viewed from above, thruster 0 rotates counter-clockwise and thruster 1 rotates clockwise. Select the propeller such that positive voltage results in downward thrust. In the case of the high-efficiency propellers, the prop labeled 5x4.6E is intended for thruster 0, and that labeled 5x4.6EP is intended for thruster 1.
- 9. Slide the propeller assembly on to the motor shaft and tighten the cap.
- 10. Place the propeller cover back in position and replace the screws holding it in place.



Caution: Ensure that the Quanser Aero is completely reassembled, with all screws in place before connecting power or attempting operation. The outer propeller guard screw must be fastened with the included lock nut. Improper assembly may result in damage and/or injury.

3.7 Pitch and Yaw Locks

To lock the pitch of the Aero body, use the included hex key to tighten the pitch lock screws as shown in Figure 3.4a

To lock the yaw of the yoke, remove the hex key from its storage location in the bottom of the yoke and reinsert it with the long arm of the key down as shown in Figure 3.4b. Once the key is inserted all the way and protruding from the bottom of the yoke, rotate the yoke clockwise until the key comes in contact with the magnetized stop on the yaw pivot.



(a) Locking Pitch

(b) Locking Yaw

Figure 3.4: Quanser Aero attitude locks

Quanser aerospace and unmanned systems for teaching and research



These systems allow you to study or research traditional and modern controls applications relating to spacecraft, unmanned vehicles, rescue missions and autonomous control. For more information please contact info@quanser.com

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F.3 Quanser AERO Embedded Data Sheet



QFLEX 2 Embedded for Quanser AERO

For additional resources visit www.quanser.com/courseware

Features

Control DC motor voltages Read encoder feedback from DC motors Read encoder value from pitch and yaw pivots Read current sense feedback from PWM amplifier Read acceleration and velocity information from IMU Set encoder reference points Control tri-colour LED output

System Overview

The QFLEX 2 Embedded provides a standard SPI interface for interacting with the Quanser AERO. The QFLEX 2 Embedded panel is powered via the 1.8V-5V and GND input pins and provides signal isolation between the external connector and the internal QUBE data bus. Data signals in the SPI interface will operate at the voltage provided on the 1.8V-5V pin. The Quanser AERO controller receives the data from the SPI interface and sets the motor PWM duty cycle, LED brightness, and encoder reference. The controller collects current sense and encoder read values and feeds this information back over the data bus, back to the SPI interface. The full system diagram is shown in Figure 1.

Connections

The QFLEX 2 Embedded has a 7-pin connector (shown in Figure 2) which mates with a TE Connectivity 104257-6 rectangular connector. Table 1 outlines the IO connection pins

Communication

The QFLEX 2 Embedded operates as a standard SPI slave with SPI mode 2. Note that the complete data



Figure 1: System Diagram

Table 1:	QFLEX 2	Embedded	ю	connections
----------	---------	----------	---	-------------

#	Pin Name	Description
1	1.8V-5V	Controller Vcc
2	MOSI	AERO data IN
3	MISO	AERO data OUT
4	CLK	SPI data clock IN
5	QCLK	QBUS clock OUT
6	CS	AERO slave chip select
7	GND	Controller ground

packet is comprised of two sub-packets, the first intended for the AERO base board, the second intended for the core board. When SPI communication is initiated by pulling the CS line low, the first byte of data received by the QFLEX 2 Embedded, referred to as the mode byte, will determine the communication mode to follow.

Mode = 0 Only read ID (no settings changed)

Mode = 1 A full command packet is transmitted

Concurrently, as the mode byte is being received, the Quanser AERO will respond with the upper byte of the base board ID. The next byte sent to the QFLEX 2 Em-



Figure 2: QFLEX 2 Embedded Panel

bedded is a padding byte with a value of 0, the data returned during this transmission will be the lower byte of the base board ID. At this point if Mode = 0, the next expected byte will be the mode byte for the core board, or byte 15 in a normal packet. If Mode = 1, the command communication packet continues as outlined in Table 2. Similarly, if mode 0 is selected for the core board, a padding byte will be expected and the lower byte of the core board ID will be returned, after which communication is complete and the CS line can be returned high. Otherwise if Mode = 1 for the core board, communication continues as outlined in Table 2.

Transmit/Receive Bytes

The expected values and description for each of the data bytes are as follows.

Base Mode (Transmit byte 0)

Expected value 0x00-0x01. A value of 0 will result in only the Base ID being returned. A value of 1 will initiate the transmission of a full command packet.

Base Write Mask (Transmit byte 2)

Expected value 0x00-0x7F. This byte controls what values will be overwritten on the AERO base board. The mapping of the bits in the mask to written values is shown in Table 3 To zero the pivot encoders, the write mask must be 0b00011xxx. To set the LED values, the mask must be 0b000x111.

LED Values (Transmit bytes 3-8)

Expected value 0x0000-0x03E7. The brightness of the LEDs is controlled on a scale from 0 to 999 (decimal) this value is transmitted over two bytes for each LED with the MSB preceding the LSB.

Set Encoder Values (Transmit bytes 9-14,22-27)

These bytes allow for the encoder counts value to be set as desired. The most likely application for these bytes is to send 0x00 for all bytes to zero the encoder counts.
 Table 2: QFLEX 2 Embedded Command Packet Structure

В	MOSI Data	MISO Data
	Start of Base F	Packet
0	Base Mode (0x01)	Base ID MSB
1	Padding byte (0x00)	Base ID LSB
2	Base Write Mask	Encoder 2 (23-16)
3	Red LED MSB	Encoder 2 (15-8)
4	Red LED LSB	Encoder 2 (7-0)
5	Green LED MSB	Encoder 3 (23-16)
6	Green LED LSB	Encoder 3 (15-8)
7	Blue LED MSB	Encoder $3(7-0)$
8	Blue LED LSB	Tachometer 2 (23-16)
9	Set Encoder 2 (23-16)	Tachometer 2 (15-8)
10	Set Encoder 2 (15-8)	Tachometer 2 $(7-0)$
11	Set Encoder 2 (7-0)	Tachometer 3 (23-16)
12	Set Encoder 3 (23-16)	Tachometer 3 (15-8)
13	Set Encoder 3 (15-8)	Tachometer 3 (7-0)
14	Set Encoder 3 (7-0)	Reserved (0x00)
14	Start of Core F	Packet
15	Core Mode (0x01)	Core ID MSB
16	Padding byte (0x00)	Core ID LSB
17	Core Write Mask	Current Sense 0 (15-8)
18	Motor 0 Command (15-8)	Current Sense 0 (7-0)
19	Motor 0 Command (7-0)	Current Sense 1 (15-8)
20	Motor 1 Command (15-8)	Current Sense 1 (7-0)
21	Motor 1 Command (7-0)	Tachometer 0 (23-16)
22	Set Encoder 0 (23-16)	Tachometer 0 (15-8)
23	Set Encoder 0 $(15-8)$	Tachometer 0 (7-0)
24	Set Encoder 0 (7-0)	Tachometer 1 (23-16)
25	Set Encoder 1 (23-16)	Tachometer 1 (15-8)
26	Set Encoder 1 (15-8)	Tachometer 1 (7-0)
27	Set Encoder 1 (7-0)	Status
28	Padding byte (0x00)	Encoder 0 (23-16)
29	Padding byte (0x00)	Encoder 0 (15-8)
30	Padding byte (0x00)	Encoder 0 $(7-0)$
31	Padding byte (0x00)	Encoder 1 (23-16)
32	Padding byte (0x00)	Encoder 1 $(15-8)$
33	Padding byte (0x00)	Encoder 1 $(7-0)$
34	Padding byte (0x00)	$\Delta ccelerometer X (15-8)$
35	Padding byte (0x00)	Accelerometer $X(7_0)$
36	Padding byte (0x00)	Accelerometer X (15-8)
37	Padding byte (0x00)	Accelerometer $Y(7.0)$
38	Padding byte (0x00)	Accelerometer $7(15.8)$
30	Padding byte (0x00)	Accelerometer $Z(7.0)$
40	Padding byte (0x00)	$Acceleronieter \Sigma (1-0)$
40 //1	Padding byte (0x00)	Gyroscope X (7.0)
41	Padding byte (0x00)	Gyroscope X (1-0)
42 74	Padding byte (0x00)	Gyroscope V (7.0)
40	Padding byte $(0x00)$	$G_{V} = \frac{1}{2} \left(\frac{1}{2} - 0 \right)$
44 15	Padding byte (0x00)	Gyroscope Z (10-0) Gyroscope Z $(7,0)$
40 46	Fauling byte (0x00)	Gyruscope $\angle (7-0)$
40 47	Fadding byte (0x00)	
41 10	Fadding byte (0x00)	Reserved (UXUU)
40 40	Padding byte (0x00)	Reserved (UXUU)
49	Padding byte (UXUU)	Reserved (UXUU)
50	Padding byte (0x00)	Reservea (UXUU)

Table 3:	Base	Write	Mask E	Bit I	Mapping
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b	Action Enabled
7	-
6	-
5	-
4	Set Encoder 3
3	Set Encoder 2
2	Write Blue LED
1	Write Green LED
0	Write Red LED

Core Mode (Transmit byte 15)

Expected value 0x00-0x01. A value of 0 will result in only the Core ID being returned, ending communication after byte 16. A value of 1 will continue the transmission of a full command packet.

Core Write Mask (Transmit byte 2)

Expected value 0x00-0x7F. This byte controls what values will be overwritten on the AERO base board. The mapping of the bits in the mask to written values is shown in Table 3 To set the motor encoders, the write

Table 4: Core Write Mask Bit Mapping

b	Action Enabled
7	-
6	-
5	Set Encoder 1
4	Set Encoder 0
3	Write Motor 1 Voltage
2	Write Motor 1 Enable
1	Write Motor 0 Voltage
0	Write Motor 0 Enable

mask must be 0b0011xxxx. To enable the motors and allow a command value to be written the mask must be 0b00xx1111.

Motor Command (Transmit bytes 18-21)

Bit 15 of the motor commands control whether the PWM amplifier in the Quanser AERO for that motor is activated. Thus the upper byte of the motor command must be 0b1xxxxxx in order for the motor to be enabled. The expected value of the bits 14-0 of the motor command is a value between -999 and 999 (decimal) formatted in 2's complement and represents a value equal to 10 times the desired percentage duty cycle of the PWM amplifier. For example, writing the motor command value 0x81F4 would activate the PWM amplifier and apply a 50% duty cycle, equivalent to approximately 12VDC.

Padding Bytes (Transmit bytes 28-50)

The value of these bytes are ignored by the QFLEX 2 Embedded, serving only to provide clock cycles for additional MISO data. By convention these bytes are usually zero.

Base Module ID (Receive bytes 0-1)

Expected Value of these bytes for the Quanser AERO Base Module is always 0x304 or 772 decimal. Any other value indicates a fault in communication between the QFLEX 2 Embedded and the AERO base board.

Read Encoder Values (Receive bytes 2-7,28-33)

These bytes represent the current value of the encoder counts, represented in 2's complement. Each value consists of three bytes and begins with the most significant byte. Encoders 0 and 1 are the encoders indicating the position of the DC motors. Encoder 2 is the encoder connected to the shoulder "pitch" pivot of the AERO body. Encoder 3 is the encoder connected to the base "yaw" pivot of the support yolk. Note that encoders 0-2 are 2048 quadrature counts per rotation, while encoder 3 is 4096 quadrature counts per rotation.

Tachometer Value (Receive bytes 8-13,21-26)

These bytes represent the current value read from the Quanser AERO internal tachometer, consisting of three bytes and beginning with the most significant byte. Tachometer values are calculated for each encoder. The most significant bit represents the direction of rotation with a value of zero indicating counter-clockwise rotation. Bits 23-0 represent the number of clock cycles (at 40MHz) between rising edges of the encoder *A* signal line. The encoder value can be converted to encoder counts per second with the following equation:

$$\text{Counts/Sec} = \frac{4}{Tach \cdot (25 \cdot 10^{-9})} \tag{1}$$

Note that when the motor/pivot velocity is 0 the tachometer value will indicate the maximum value of 0x7FFFFF indicating a reading of approximately 20 counts/s.

Core Module ID (Receive bytes 15-16)

Expected Value of these bytes for the Quanser AERO Core Module is always 0x307 or 775 decimal. Any other value indicates a fault in communication between the QFLEX 2 Embedded and the AERO core board.

Current Sense (Receive bytes 17-20)

These bytes represent the measured current draw of the DC motors, represented in two's complement. The current can be calculated using the equation:

Current (mA) =
$$\frac{I_{sense} - 8190}{9828}$$
 (2)

Status (Receive byte 27)

The three least significant bits of this byte represent various warning or error states as outlined in Table 5. A status byte value of 0x00 indicates normal operation.

Table 5: Status Bit

b	Indicated Status
7	-
6	-
5	Motor 1 Stall Error
4	Motor 1 Stall Detected
3	Motor 0 Stall Error
2	Motor 0 Stall Detected
1	Amplifier 1 Over-Current Fault
0	Amplifier 0 Over-Current Fault

Accelerometer Readings (Receive bytes 34-39)

There are three accelerometer readings each represented by a 16-bit signed integer in two's complement. The full scale reading is $\pm 2g$ so the acceleration can be calculated from the measured value using the following formula:

Acceleration =
$$\frac{2 \cdot Acc_n \cdot 9.81}{32768}$$
 (3)

Note that only the X axis (extending down the center of the AERO body towards thruster 0) and Z axis (extending vertically when the AERO body is horizontal) will be of interest. Under normal operation the Y-axis acceleration should always be approximately zero.

Gyroscope Readings (Receive bytes 40-45)

There are three accelerometer readings each represented by a 16-bit signed integer in two's complement. The full scale reading is ± 245 degrees/s so the angular velocity can be calculated from the measured value using the following formula:

$$\omega_n = \frac{Gyro_n \cdot 245}{32768} \tag{4}$$

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